

A REAL TIME IONOSPHERIC DRIFT SYSTEM

Description and operating

manual

(4 antennas)

by

C.E. MEEK

and

J. SOIFERMAN

Institute of Space and Atmospheric Studies

Physics Department

University of Saskatchewan

Saskatoon, Canada

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ERRATA (JUNE 1979)

During the second printing of this report, the opportunity has been taken to correct some minor errors in the text. Additional modifications which could not be easily implemented are listed below.

CH.2(Pg.5)

Reference should have been made to G.J.Fraser (PhD thesis, 1965, Univ. of Canterbury, N.Z.) who appears to have been the first to apply the binary correlation method to drift analysis.

CH.3(Pg.8)

The gate separation of $20\mu\text{sec.}$ corresponds to a free space sampling separation of 3Km.

(Pg.11,last paragraph)

P.I.A. is later used as an abbreviation for peripheral interface adaptor.

APP.A(Pg.40)

The first statement in interrupt should have been LDA \$45 since there is some internal processing which stores the accumulator in \$45 before the program jumps to the interrupt address. In the present program this omission is unimportant.

APP.B(Pg.62,line 4)

Insert "when they are not" after "(i.e. equal in)"

APP.D(Pg.82)

The delay parameter (CLKIN) given in the program actually covers the clocking in of 65 height gates.

Acknowledgements

The authors would like to thank Dr. D.G. Stephenson for his helpful comments on Chapter 3, and Mrs. Alice Brown for some of the typing and artwork.

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I. Introduction

The measurement of winds in the mesosphere by partially reflected radio waves consists in estimating the velocity of the ground amplitude pattern, which itself may be varying internally in time and have 'non-uniform' structure. Although there has been some argument in the past over the relation of this pattern velocity to the actual neutral drift, it is usually taken to be twice the drift velocity. The pattern is sampled at at least three non-colinear ground locations (antennas) and at heights (given by the delay from the transmitted pulse) in the range $\sim 60-120$ Km. A simple estimate of velocity, called apparent velocity, can be made by examining fading sequences at different antennas for similar features, and measuring their delay between antennas; or by calculating the cross correlations between fading sequences and looking for the lag at which there is maximum correlation (Fig. 1). This is a good estimate of the

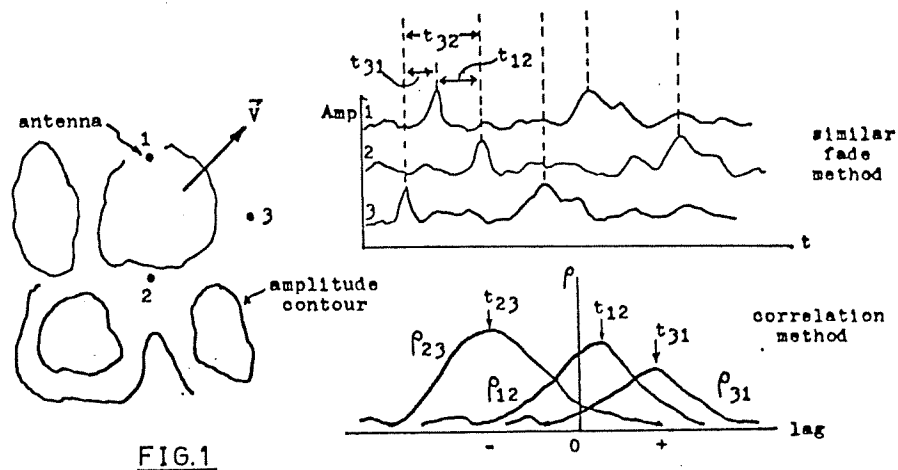


FIG.1

pattern velocity if the pattern, considered on the average over the record length, is statistically the same in all directions, and does not vary internally as it moves.

There is no way of testing for these conditions with just the time delays between antennas found above. It can be shown (Phillips and Spencer, 1955) that if the pattern has a preferred direction of elongation not \perp to the drift velocity, the measured apparent velocity has a tendency to be perpendicular to the direction of elongation and have a magnitude less than the actual drift (as shown in Fig. 2).

An extreme example of this would be a plane wave pattern.

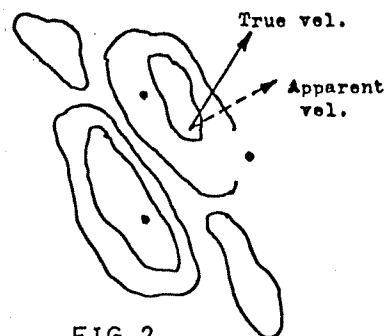


FIG. 2

If the pattern changes internally with time, an additional decrease in correlation with time delay (lag) between antennas, not related to the average pattern structure or drift velocity, shifts the position of maximum correlation towards zero lag - thus reducing the measured delays and increasing the apparent velocity (Fig. 3). This effect

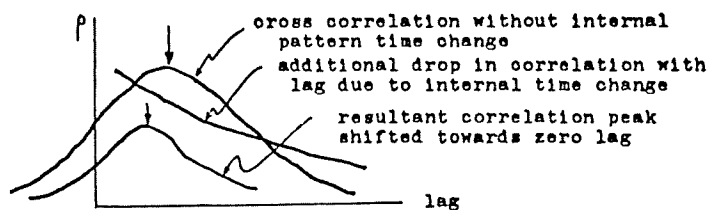


FIG. 3

usually predominates, so that the apparent velocity is usually greater than the 'true' velocity - which has taken account of these possibilities.

The rest of this discussion will be restricted to a three antenna (usually equilateral triangle) receiving array, or a system which simulates such an array (Fig. 4). It is fairly obvious

that for a single fading feature, defined by the time in each of the three fading sequences, the sum of the delays (i.e. time differences) between antennas must be zero. This is also true when correlations are used to find the delays,

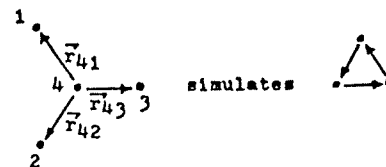


FIG. 4

assuming that the peaks selected are due to a single pattern motion, and can be shown to hold in the more general case of preferred directions of elongation and internal pattern time change (Meek, 1978). This provides a valuable criterion for deciding whether peaks chosen from the cross correlations are the result of a pattern motion to which the theory applies. In general, if t_{ij} represents the

delay between antennas i and j (+ve if the pattern moves from i to j) with separation vector \vec{r}_{ij} ; then if $\sum \vec{r}_{ij} = 0$, $\sum t_{ij} = 0$. (The latter is not restricted by the shape of the array or the number of antennas). The normalized time discrepancy, NTD, will be defined by:

$$\text{NTD} = \frac{|\sum t_{ij}|}{\sum |t_{ij}|}$$

and has a range 0-1.

Distributions of this parameter can be used to compare the data quality at different locations - i.e. how often are 'spurious' peaks the largest in the cross correlations. Since spurious peaks are quite common at our location, careful examination of the cross correlations is required to maximize data output while rejecting most of the spurious values. In the present analysis selection of peaks is based wholly on the NTD, although the widths of the peaks might also be useful (Meek, 1978; Ch. 3).

There are many reasons for having real time analysis of data. Foremost is cost. For example, on the present university computer (IBM370), it takes ~ 1 HR to analyse a day's data (records every 5 min). A high record rate is desirable in order that tidal components, gravity waves, etc. may be examined. Although a more suitable computer would probably be cheaper in terms of execution time, the expense is still prohibitive. Second is storage. At the above rate, a standard 2000⁰ tape (800BPI, recording the amplitude sequences) lasts about $1\frac{1}{2}$ days. Full (or partial) analysis of the data to be stored reduces this drastically - a factor of ~ 50 with the present system; i.e. a 2000⁰ tape should last about 2 months. Third, since the data is being analysed in situ, the results may be fed back into the system control to change operating parameters, specifically receiver gain, so that data are collected under optimum conditions.

II. Considerations for a micro-computer system

There are three aspects to be considered; firstly the control of data gathering, secondly data correlations, and thirdly the calculation of drift velocities.

a) Control

In order that fading sequences may be correlated (without interpolation), pulses must be evenly spaced for each receiving antenna. A sample spacing of ~ 0.5 sec at each antenna is a recommended maximum (~ 150 m antenna spacing) to insure that at least one cross correlation peak be far enough away from zero lag for accurate determination of, for example, NTD, given normally occurring wind speeds. Since timing is important here, this process should be under interrupt control by an external synch pulse which also controls the transmitter.

In addition to pulse timing, the signal must be sampled at an accurate and constant set of time delays. These "height gates" will be required to be at least as closely spaced as the transmitter pulse width (e.g. $20 \mu\text{sec}$ in our case) to get the maximum height resolution (any smaller spacing means that adjacent height samples cannot be considered independent). Since micro-processor instructions take from $\sim 2-7 \mu\text{sec}$, it is rather difficult to put these gates under direct software control (although the delay to the first height gate, ~ 50 Km, may be).

The standard system, in which raw fading sequences are recorded on magnetic tape and processed in a large computer, has the whole record (e.g. 300-400 amplitudes per height per antenna) available for analysis. This is impossible in a micro-computer due to storage limitations, and data must be dealt with in smaller portions, called blocks here. A block consists of (n amps.)x(m ants.)x(p hts.) amplitudes. It is easiest to use one byte (0-255) per amplitude.

b) Correlations

These must be done on a 'partial' basis, i.e. only one or two blocks available at a time. It is possible to do standard correlations by accumulating sums, squares, and cross products of amplitudes from two sequential small

blocks whose minimum length is the maximum lag required; however, multiplication is very time consuming. For example, a test of this process on a mini-computer (BASIC) indicated that processing time (for 24 heights) was a factor of ~ 36 greater than data collection time.

One solution, and that which is used presently, is to convert each block to binary with respect to the mean of the particular sequences. Two blocks of binary (which occupy $\frac{1}{4}$ the storage of an original block) are saved so that 'partial' correlations can be accumulated. This is very fast, since a single 'AND' instruction can be used to correlate 8 points. Another advantage of this process is that short-lived jumps in amplitude (Fig. 1) whose effects are often seen in the range ~ 85 -100 Km, and are probably due to meteor trails, only affect a small part of the whole record. A disadvantage is that if there are only a few scatterers distributed throughout the record (Fig. 2), the blocks recorded in between, essentially noise, will have equal effect on the correlations as the blocks with signal.



FIG.1

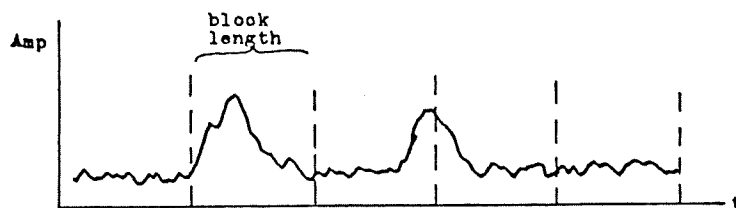


FIG.2

Another solution ('average cross correlation') tested was to correlate each block separately and average (sum) the results. The test gave approximately the same number of wind values as the method above. Conceptually it seems that this would give more accurate values if the data were not stationary between blocks as shown in Fig. 3 (i.e. the mean block amplitude varies significantly). It would also be possible to reject blocks in which there was insufficient signal (i.e. noise). However, this would be no help at night when the

signal to noise level is low - also variable rejection criteria would probably be required in the subsequent analysis depending on the number of blocks finally used at a given height.

The latter comment raises an interesting point- viz. do the magnitudes of the correlation peaks depend on the length of the record.

If the peaks are due to a constant pattern motion, then it appears that they should not be affected. However, spurious high correlation values are more likely, for statistical reasons, as the number of points correlated is reduced. Another interesting question involves the effect of block length on the results. It appears that the block method used constitutes a form of high pass filter on the raw amplitudes. Further investigation of these points would be worthwhile.

Because storage is limited, blocks must be converted to binary as soon as they are taken. In order that there not be a gap in the time sequences (although this is not **important** for the average cross correlation method), the conversion must be done between the handling of pulses. The length of time required depends on the block size. The available time depends on the pulse rate. A large block size is important for 'stationarity' reasons. As mentioned previously, choice of pulse rate is restricted by the required sample spacing.

Apart from binary conversion, there is a lot of free time left during a record which can be used for correlations and other tasks.

c) Calculation of wind

The determination of drift velocity is a 'numerical' rather than 'logical' process, and so is easiest to do in one of the higher computer languages.

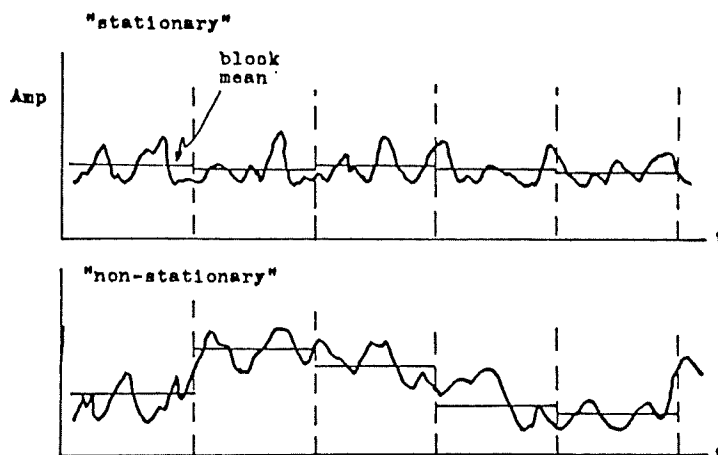


FIG.3

This could be done in the same micro-computer, but would require a lot of time at the end of a record - and a good knowledge of the particular compiler or interpreter (e.g. BASIC) so that memory locations do not conflict. The easiest way is to use a second micro-computer to do the numerical computations; this also will allow a faster record rate, since the first computer will not have to pause for numerical calculations. The method of analysis must be fast, while still allowing 'quality' testing, and rejection of poor data. The latter task essentially consists in picking out the 'proper' peaks in the cross correlations. After the peaks have been found, there are several methods of analysis available. One is graphical (Briggs,1968), requiring the determination of crossing points of correlation curves. Another is a least squares fit to a Gaussian correlation function (Fedor,1967), which requires the inversion of a 6x6 matrix plus several other matrix operations. The method finally chosen was derived by Meek(1978), and is called the Poor man's full correlation analysis; poor not in the sense of giving poor results, but in that it requires the minimum amount of input data to produce a wind value. The graphical analysis is not suitable for our location because of our often irregularly shaped correlation peaks, and the least squares fit seems to require too much storage and execution time.

III. Brief description of present system and operation

a) Present system

Table 1 lists the present operating parameters. Fig. 1 is a plan of the transmitting and receiving arrays. Parallel dipoles in the outer squares are connected in parallel. The linear N-S mode is used for transmission and reception.

Table 1.

Transmitter:	power- 40 Kw pulse width- 20 μ sec (variable) pulse rate- 7.5 Hz frequency- 2.22 MHz antenna $\frac{1}{2}$ power beamwidth- $\sim 22^\circ$
Receiver:	bandwidth- ~ 75 KHz Gain- voltage controlled, range ~ 60 dB
System:	Rx gain control- any of 8 manually set voltages # blocks/record- 8 block length- 34 sec #height gates- 64 (32 used) gate width- 20 m gate separation- 20 μ sec (programmable) record rate- every 5 min sampled height range- ~ 49 -142 Km

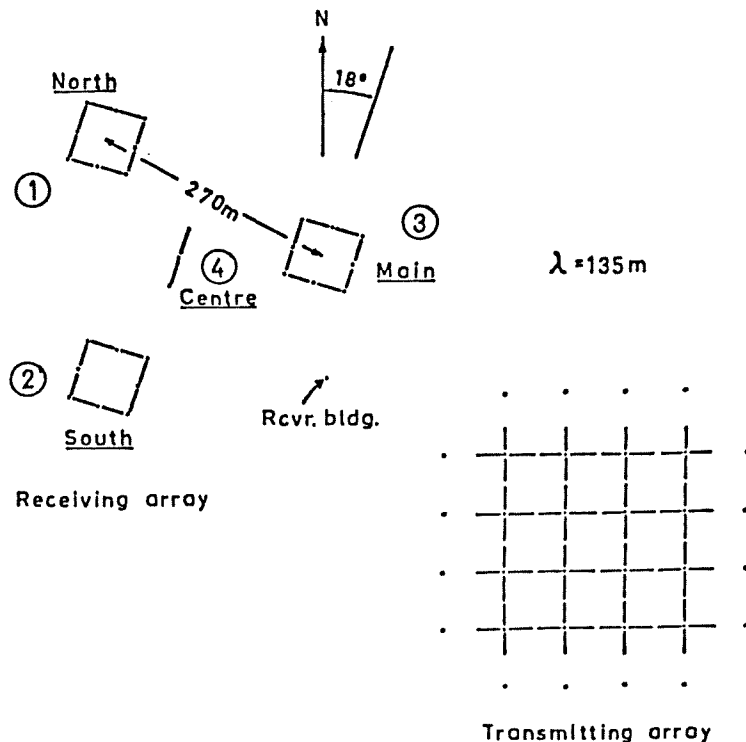


Fig. 1

Fig. 2 shows a block diagram of the receiving/analysing/recording system.

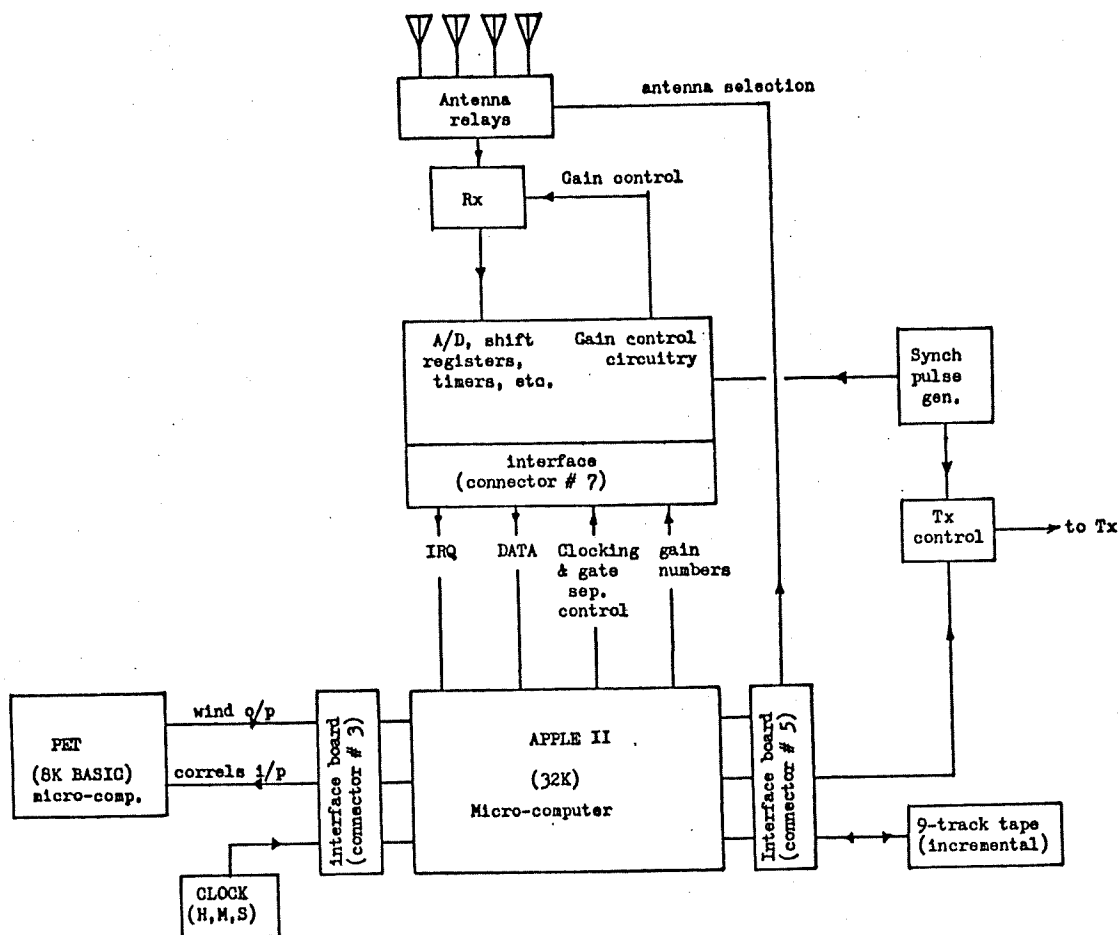


Fig. 2

b) Interfaces to Apple computer

Three peripheral interface boards are used by the Apple II. These are the "analogue interface board" (pg. 14) which sets the receiver gain for each height gate, and samples the receiver output; the " tape interface board" (pg.15) which controls the 9-track tape recorder and selects antenna relays through the relay board (pg.16); and the interface board (not shown, but essentially the same as the tape interface board) connecting the Apple to the external clock and the Pet computer. These are plugged into the C700, C500, and C300 peripheral

interface connectors respectively, on the Apple. The analogue interface is the most important, and will be described first.

(1) Analogue interface

The circuit diagram is shown on page 14 . The Tx synch pulse (which also interrupts the Apple through this interface) resets the logic and initializes a free running counter (IC7, pin 11) which puts out a 20 μ sec square wave. This counter, like all others on the board, is driven by the Apple clock (ϕ_0 , 1 MHz) or its sub-multiples. The Apple generates a software delay, when interrupted, ($\sim 50\text{Km}$) and then loads a 3-bit binary number (variable GSEP) into the Johnson counter (IC9). This action enables the Johnson counter

to start counting at the next positive transition of IC7-pin 11. It then puts out a string of evenly spaced 1 μ sec pulses (the separation is determined by the 3-bit number;

Table 2 shows available values). These pulses control both the A/D converter and the gain control circuitry. When a pulse is received by

the A/D (IC16), it samples the receiver output (0-10V) through the sample and hold amplifier (IC15), converts it to binary (8 bits), and stores it in the 8x64-bit shift registers (IC1-4). When a pulse is received by the gain circuitry, it reads a 3-bit number from the 3x64-bit shift registers (IC18,19) which have been previously loaded by the Apple program, and uses this to select one of 8 preset voltages through a multiplexer (IC21). This voltage is used to control the gain of the external receiver. Thus the gain is being changed just as the receiver output is sampled; however, the receiver has a relatively long response time to a gain change, and so the output is not affected immediately.

As the amplitudes and gains are being clocked in under hardware control, the Apple program provides a software delay (which has been determined in advance for the particular clocking rate used, and is defined by the variable CLKIN in

Table 2.

Gate separation (μ sec)	loaded number (hex)
5	\$20
10	\$10
15	\$30
20	\$08
25	\$28
30	\$18
35	\$38

the program) and terminates the clocking when exactly 64 amplitudes have been stored. The amplitude shift registers are now full, and the gain registers empty. The program then modifies the input to IC5 so that the amplitudes may be clocked out under program control, and later to IC20 so that the gain shift registers may be reloaded by the program for the next pulse. Figure 3 illustrates the sequence of events.

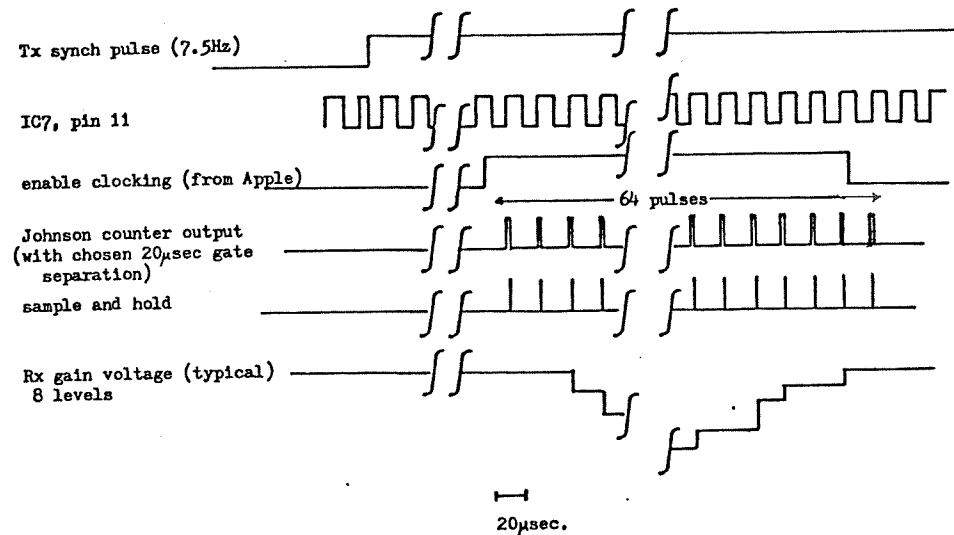


Fig. 3

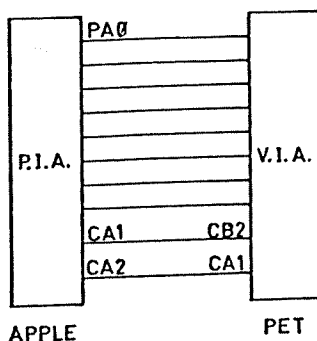
The reason for tying the sampling to the next positive transition of IC7-pin 11 after the Apple has sent the enable command is that some slack is permitted in the Apple software delay without affecting the actual delay of the first height gate from the transmitted pulse. This is necessary because the Apple must complete an instruction before responding to an interrupt request, which may take up to $\sim 7\mu\text{sec}$, depending on the type of instruction (in the present program the actual delay is $\leq 2\mu\text{sec}$). (Note that the disable works in the same way)

(2) Tape interface board

There are two peripheral interface adapters on this board; one is used for data and command output to the 9-track incremental tape recorder, and the other for tape signal lines (input), and antenna relay control (output). A spare line on the board is used for enabling the transmitter.

(3) Pet and clock interface board

Fig. 4 shows the connections between the P.I.A. (Apple) and V.I.A. (Pet).

Fig. 4

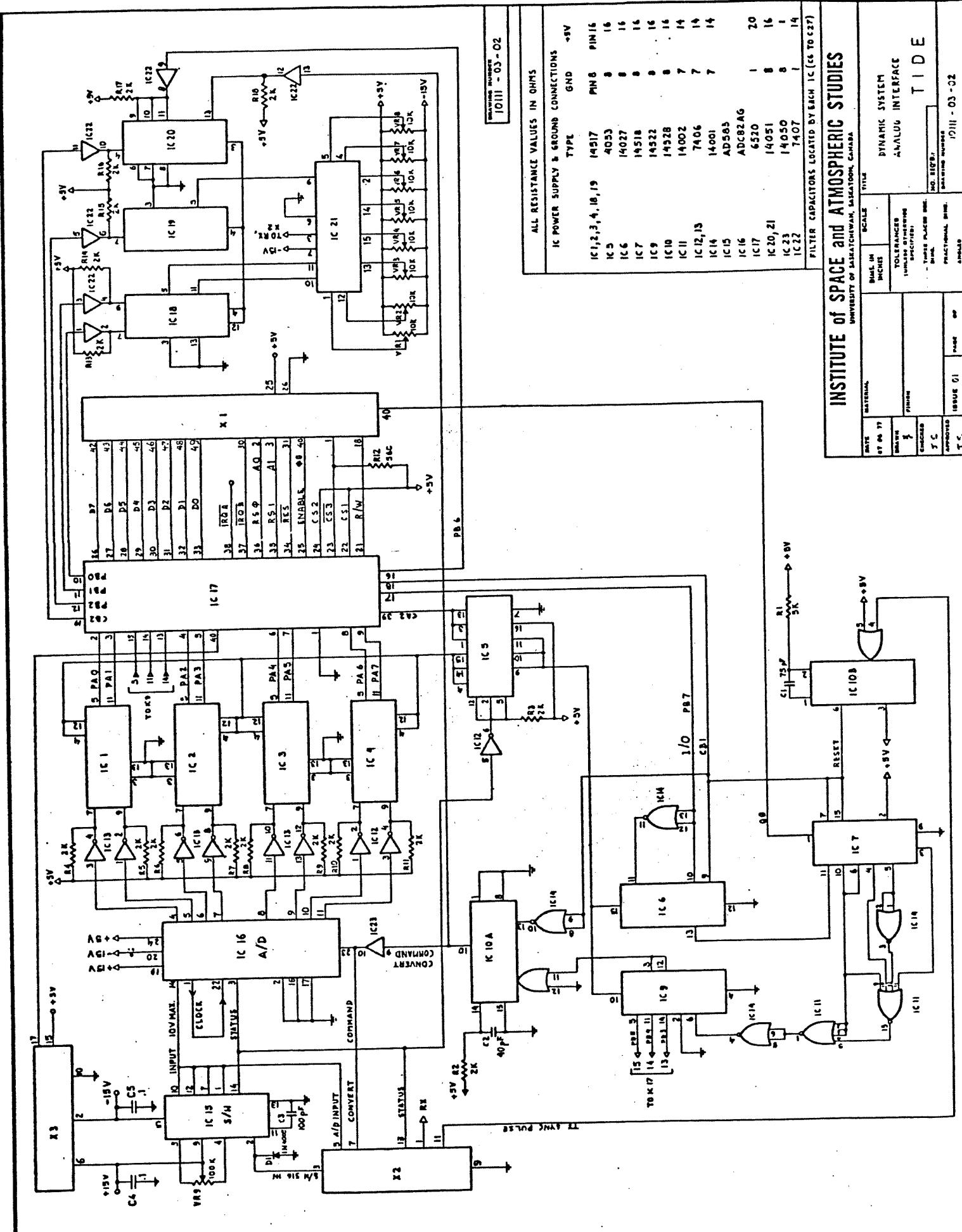
(c) Program operation

The program operation is as follows. After the program has been started it checks for a 0 or 5 min mark (seconds are ignored). When this is found it turns on the transmitter and waits to be interrupted by the synch pulse. The antennas are selected cyclically under program control. When a block has been stored, it is converted to binary between pulses - the mean signal being accumulated for use in setting gains in the next record. During the second and further blocks, correlations are also done between pulses. At the end of a record (which is determined by a block count) the average signal for the whole record is calculated, and the receiver gains set accordingly for the next record. At the beginning of the second record, the correlations calculated in the previous record are transferred to a second storage location from which they are fed to the Pet computer one height at a time. At the same time, new correlations are being calculated, and the results from the Pet calculation are received and stored on 9-track tape. All of these tasks are performed between the handling of incoming amplitudes.

Since the correlations are done as the record progresses, and only 6-7 seconds are required at the end to finish these, the record rate is not dependent on the Apple program. Calculation of wind values is rather time consuming however, and it is possible that the Pet may not finish in time for the next record. Consequently a new record is not initiated unless the Pet is finished

with the last.

A comparison between winds from this real time system and a 'standard' system has been shown by Gregory et al. (1978).



10111 - 03 - 02

ALL RESISTANCE VALUES IN OHMS

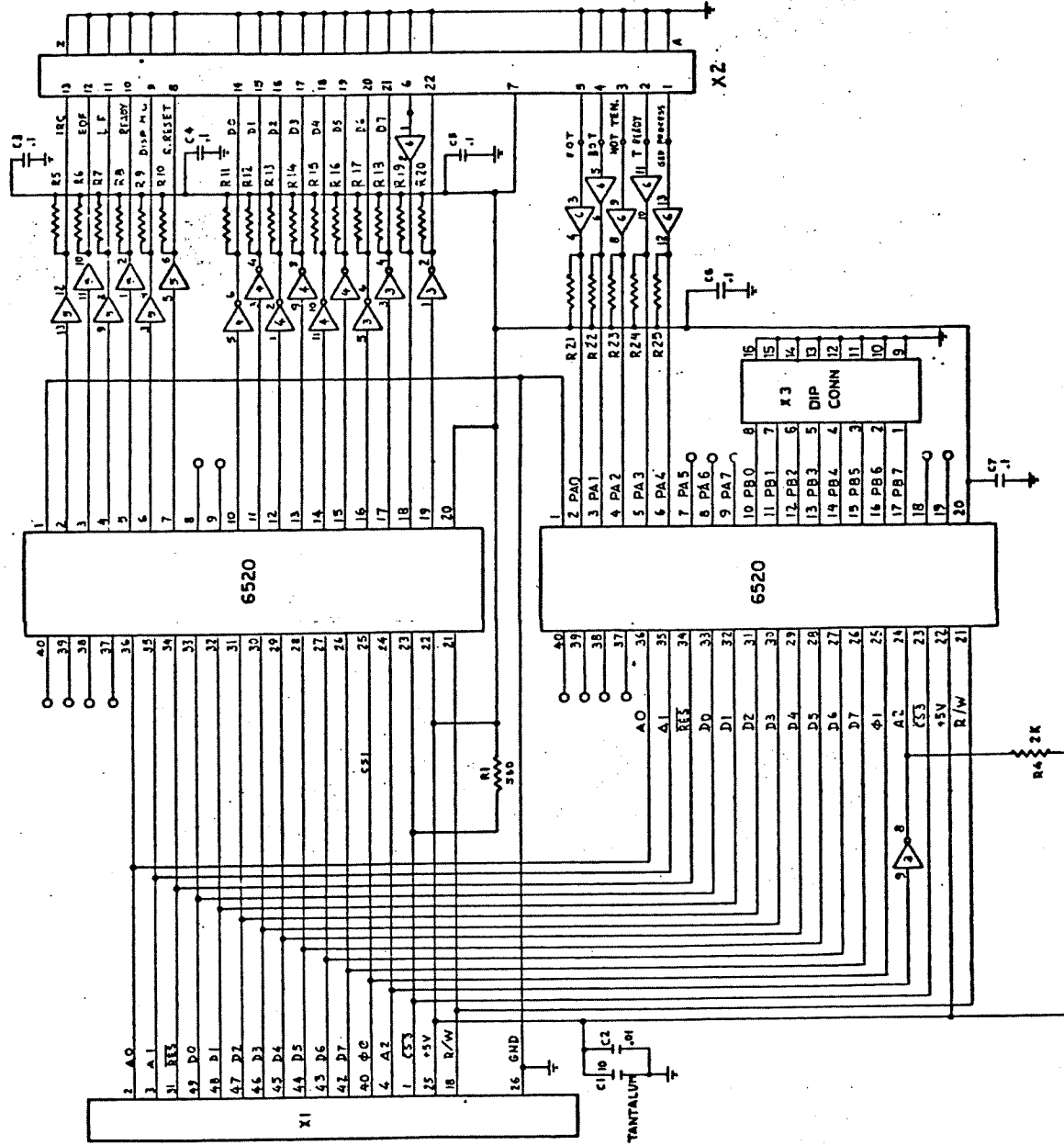
IC	TYPE	GND	+5V
IC 1, 2, 3, 4, 10, 19	14517	8	16
IC 5	4053	8	16
IC 6	14027	8	16
IC 7	14518	8	16
IC 9	14522	8	16
IC 10	14528	8	16
IC 11	14002	7	14
IC 12, 13	7406	7	14
IC 14	14001	7	14
IC 15	AD565		
IC 16	ADC82AG		
IC 17	6520	1	20
IC 20, 21	14051	8	16
IC 23	14050	8	1
IC 22	7407	1	14

FILTER CAPACITORS LOCATED BY EACH IC (C6 TO C27)

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DATE	BY	SCALE	TITLE
07 04 79	J.S.	1/8"	DYNAMIC SYSTEM ANALOG INTERFACE
DRAWN	CHECKED	DESIGNED	APPROVED
J.S.	J.S.	J.S.	J.S.
ISSUE 01	PAGE 02	PROJECT NUMBER	10111 - 03 - 02

DYNAMIC SYSTEM ANALOG INTERFACE
T I D E



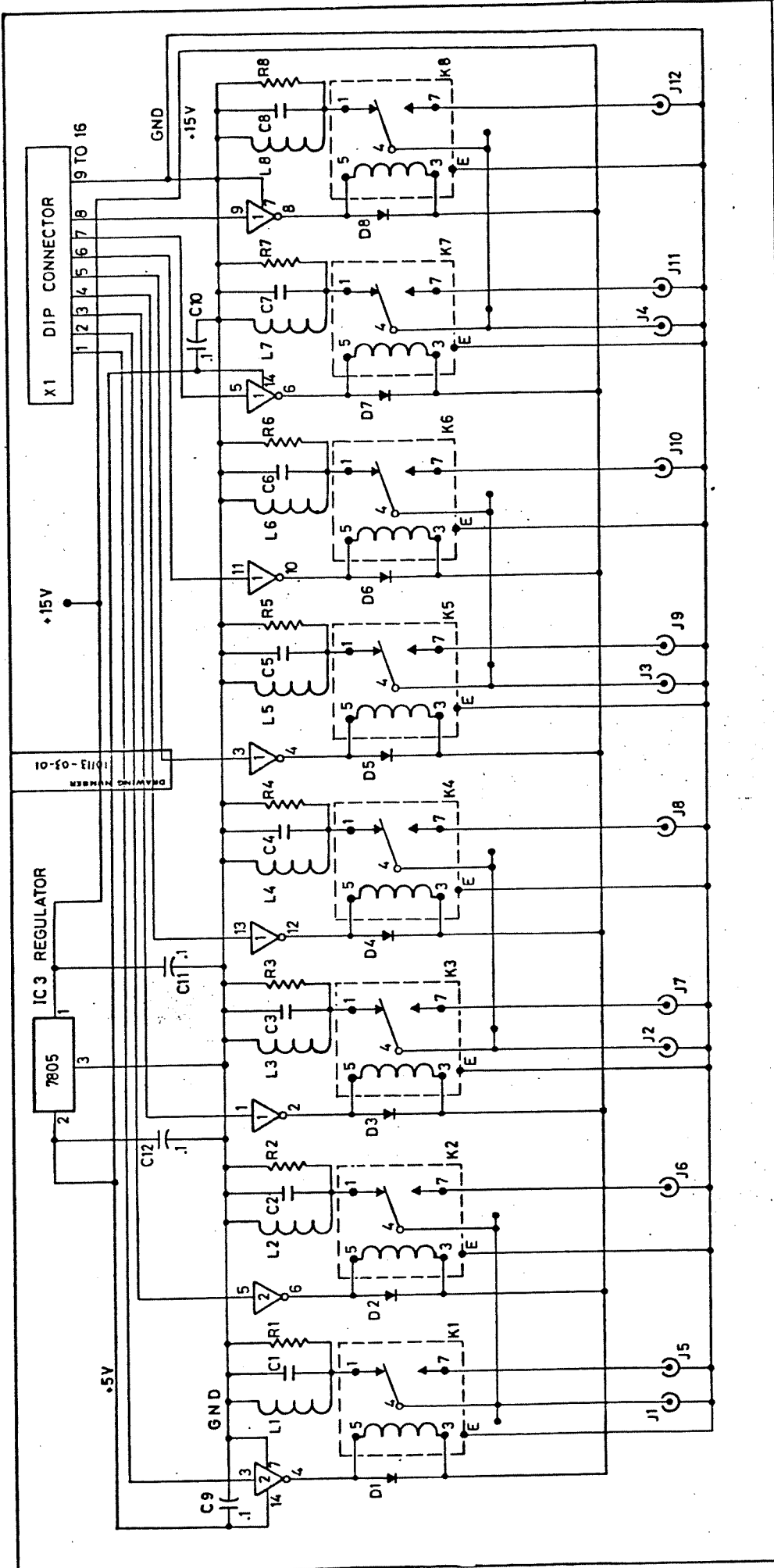
SUPPLY BUSY
W STEP COMM.

R5 THROUGH R20 = 180Ω EXCEPT
R19 = 2.2K OHMS
R21 TO R25 = 2.2K OHMS

REVISING NUMBER
10112-03-01

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DATE	MATERIAL	SCALE	TITLE
10-07-77			TAPE INTERFACE
DRAWN BY	PROJEN	TOLERANCES (UNLESS OTHERWISE SPECIFIED)	
ENGINEER		UNLESS PLACED ON DRAWING	
APPROVED		FUNCTIONAL DATA	
ISSUE		NUMBER	
PAGE			
OF			
			REV. 10112-03-01



RELAYS SHOWN IN NON-ENERGIZED MODE
 DRIVER DEVICES ■ 7406
 ALL DIODES IN4007
 ALL RESISTANCE VALUES IN OHMS
 ALL CAPACITANCE VALUES IN PICOFARADS
 J1 THRU J12 BNC CONNECTORS

DATE		MATERIAL		TITLE	
10-08-77		RELAY BOARD		UNIVERSITY OF SASKATCHEWAN, SASKATOON, CANADA	
DRAWN		FINISH		TOLERANCES (UNLESS OTHERWISE SPECIFIED)	
7			 THREE PLACES DEC. DIMS.	
CHECKED		ISSUE	 FRACTIONAL DIMS.	
		OF	 ANGLES	
APPROVED		PAGE		DRAWING NUMBER	
		10113-03-01		No. Req'd.	
				10113-03-01	

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IV. Operating problems

This chapter outlines some of the problems met in practice and their resolution, or ways in which they may be overcome in future modifications.

a) Receiver gain control

With the present receiver it is not possible to change the gain in 20 μ sec (the usual gate separation) because of the narrow bandwidth. Some re-design is proceeding which will place the gain control after the I.F. stage, but this will probably remain a problem. In the program (Version 17d, Apple) the attempt to set a different gain for each gate depending on the measured mean amplitude failed because the gain could not be set independently for each gate. The result was that the gain (and signal) usually 'locked' into a herringbone pattern, since the gain for one height gate had more effect on the next. A partial solution (version 18, Apple) is to set the gain for groups of four adjacent height gates on the basis of the maximum signal in the last three, since the gain for the first is uncertain. This allows overloading in the first gate of every four, since its signal is not considered, however it has turned out to be only a minor problem.

b) Overloading

Appendix B, section 5 (pg.63) describes the serious effect of overloading on the correlation values. It is not possible to avoid overloading completely because of unpredictable changes in signal strength occurring over periods longer than a record length. This problem can probably be resolved by modifying the Apple program to save separate means of the binary sequences (i.e. separate zero-lag auto correlations) instead of the present practice of using an 'average' over all (three) antennas for the calculation of cross correlations in the Pet.

c) A/D converter

The particular A/D converter used does not tolerate overloading. For minor overloading ($\geq 10V$) the output is zero; for greater overloading it is unpredictable—thus the substitution of \$FF for \$00 in the Apple program. A final solution was to put a 9.9V zener diode across the receiver output.

d) Pet

It was found that the temporary data storage (1 page, see Appendix C, pg. 78) was not safe from the Basic system even though the internally calculated storage requirements seem to leave enough room, and no string operations are used. Consequently, the transfer of all data into Basic program variables was made as soon as possible in the program. It would be more efficient to check the auto correlation for fast fading first and only read the rest if necessary, since the 'PEEKING' of data out of storage is time consuming. However, the present program does finish a record in less than the required time (5 min) so nothing will be done about this at present.

V. Conclusions and future work

The real time system has been operating since Sept. 1978 with the gain control modification described in Chapter 4. The amount of output is $\sim 10-20\%$ less than for the previous system in which raw data (amplitudes) were recorded and analysed at great expense on a large computer, presumably due to the binary method in use; but the quality of the resulting data is essentially the same. Some increase in output is expected when separate means are used in the calculation of cross correlations.

Since incremental tape recorders are becoming obsolete, a future modification of the program will be made to save the output for a full record and dump it onto tape at the end of the record. The present four antenna array is unique, thus a modification will be made to allow operation on the more usual three antenna array (commonly an equilateral or right angle triangle arrangement). A slightly lower pulse rate ($\sim 6\text{Hz}$ instead of 7.5) and longer block length (80 pulses/antenna instead of the present 64) will be employed.

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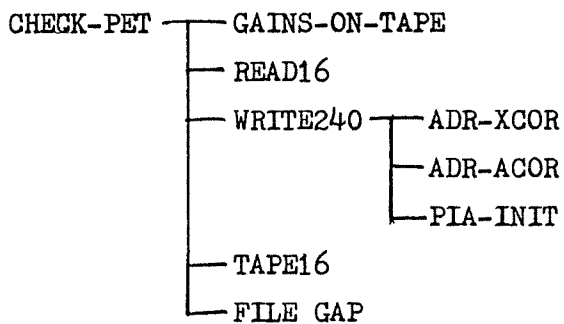
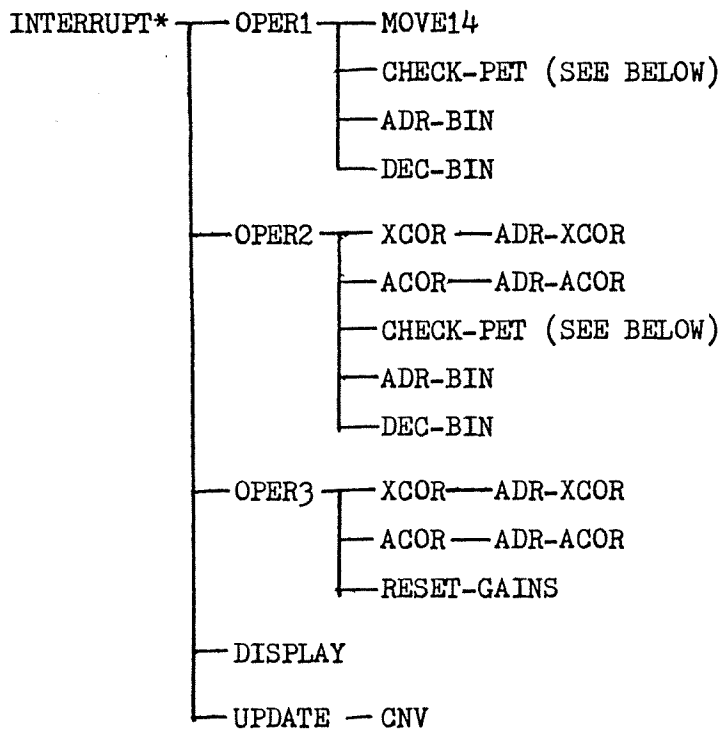
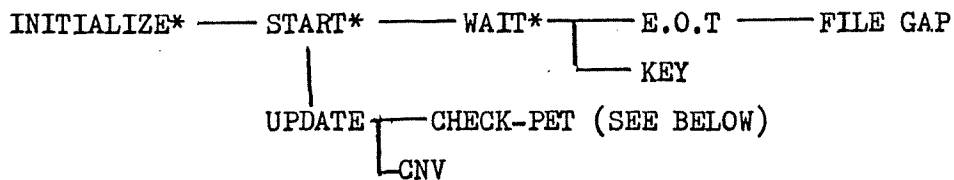
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Saskatoon, Canada
- Phillips, G.J., 1955 Proc. Phys. Soc. , B68, 481
M. Spencer

APPENDIX A

Apple program details

1. Subroutine calling diagram
2. Detailed description of routines
3. Method of calculating correlation
4. Some flow charts
5. Apple bulk storage map
6. Volatile zero page storage
7. Zero page storage (constants, control flags, counters)
8. Program listing (version 17d, 4 antennas)

1. SUBROUTINE CALLING DIAGRAM (4 ANTENNAS)
 (* - NOT A SUBROUTINE)



2. Short description of routines

START : Updates clock store and checks for keyboard commands between records; checks for start time (minute units = 0,5) if the Pet is finished with the previous record. It initiates a record by setting a flag and enabling the interrupt, and then sits in a short-instruction loop waiting to be interrupted by the synch pulse. At the end of a record it disables the interrupt.

INITIALIZATION : Initializes the gain settings for the 32 height gates (low to high) to 7,7,7,7,6,6,6,6, 0,0,0,0. Reads in day#, and # blocks per record; sets bottom delay, gate separation, and clockin delay parameters; initializes the P.I.A.'s; chooses the first antenna relay; runs the gain settings into the P.I.A., and sets cycling parameters and other flags.

UPDATE : Reads H,M,S from external clock and stores in clock location and on monitor; checks for day# change; sets file gap flag when day# is changed; services the Pet (and Tape recorder) through CHECK-PET routine if a record is not in progress.

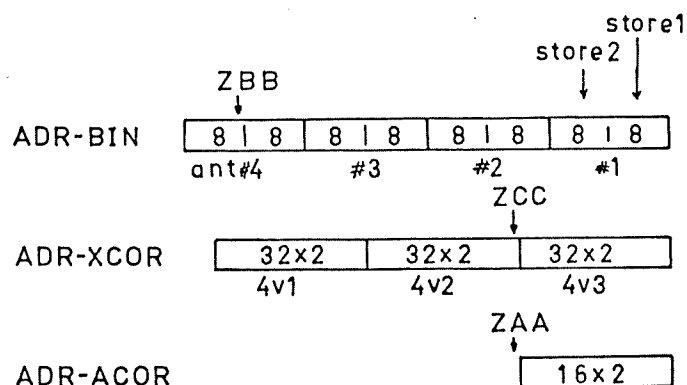
INTERRUPT : Saves accumulator and X register (not needed) on stack; gives software delay before first height gate; turns on clocking, waits for 64 gates, stops clocking; runs in gains for next pulse. If a record is not in progress (this is a faint possibility since the system could come out of interrupt at the end of a record and be interrupted again before the interrupt is disabled) it returns from interrupt at this point. It reads 32 amplitudes from the P.I.A., loads them into block storage and selects the next antenna relay. If the first block of the record is in progress, then OPER1 is called; if any other block, then OPER2; if the record has ended (by block count) then OPER3 is called.

OPER1 : Operations in this routine depend on the pulse count (0-\$FF). After the first three pulses it shifts the auto, cross and date/time store 1 to store 2 (for output to Pet) and zeroes store 1; stores the present date/time in store 1 and zeroes the binary storage (store 1 and 2). After all other pulses, except the last four, it services the Pet and tape recorder (through CHECK-PET). As the block is finished for each antenna, the data are converted to binary through DEC-BIN.

OPER2 : Again the operations depend on the pulse number. After the first 96 pulses it cycles around the cross correlation (XCOR), auto correlation (ACOR), and CHECK-PET routines. The idea is to service the Pet as often as possible since its calculations constitute the limit on the record rate. After all other pulses, except the last four, it calls CHECK-PET. At the end of the block (last four pulses) the data are converted to binary.

OPER3 : This is called after the last block in the record has been converted to binary, and it runs through the cross and auto correlation routines twice to finish up. Then RESET-GAINS is called to calculate the average signal and choose the gain settings for the next record. The record is terminated in this routine by resetting REC.

ADR-BIN, ADR-XCOR, ADR-ACOR : These calculate the address to the appropriate binary sequences and the accumulators for store 1 cross and store 1 auto correlations from the input height number (0-\$1F) as shown below:



Note that the address points to the byte below the 'boundary' for programming reasons.

DEC-BIN : Averages signal ($\div 64$) for a given antenna number (0-3) at 32 heights, and adds this to the $\sum \overline{\text{sig}}$ store (for future use in setting gains and for output to tape): $\text{avge} = \sum \text{sig}/64 + 1$ (truncated). This average is compared with each amplitude in the block to get the binary representation of the sequences (amplitude \geq avge gives a '1' bit) which is put into store 2 binary (see Apple bulk storage map, pg.35). This process takes ~ 0.08 sec for one antenna at 32 heights.

XCOR : Does 'partial' cross correlations ($4v1, 4v2, 4v3$) for one height. The linear (instead of loop) section to add 1's was found to be necessary for saving time. The method is described in detail in Section 3 (pg.27). Required time is ~ 0.06 sec.

ACOR : Does 'partial' auto correlation for one height (0 to 15 lags) over antennas #1,2,3 only (see Section 3, pg.28). Required time is ~ 0.06 sec.

RESET-GAINS : The first half of this routine divides the $\sum \overline{\text{sig}}$ by the (number of blocks times the number of antennas) to get the average signal, $\overline{\text{sig}}$, for the whole record; and stores this and the present gain settings for output to tape. The second half compares $\overline{\text{sig}}$ with 80 and 160 ($\$5\phi$ and $\$A\phi$) at each height. If it is between these limits the gain is left as is at that particular height; if greater then the gain is reduced by one step (if possible); if less then the gain is increased by one step (if possible). Ideally each gain step (adjusted manually) should be equal to the ratio of the two limits used (e.g. 80, 160), which is 6 dB here, otherwise the gain will sometimes bounce back and forth from too high to too low, even when the average signal

strength is constant. The upper limit is restricted by overloading: overloading usually begins when the mean reaches ~140, and may be 5 or 10% at a mean of 160. At present, gain steps of ~8dB are used since 48 dB (=6 dB x 8 gain steps) may not quite cover the range of signals encountered in practice.

CHECK-PET : This routine controls the tape recorder and services the Pet computer. The latter is programmed to output 16 bytes and input 240 bytes each time it is accessed. Operations are governed by flags best described by the flow chart (pg. 34). Tasks are made short enough to fit between pulses (except for the file gap - but this is put on only once per day after the fourth pulse in a record). Note that at the end of a record the Pet is holding the last output and waiting for the next input, which means that the 'true' end of the record as far as the CHECK-PET routine is concerned, is during the beginning of the next.

TAPE16 : Puts 16 bytes on tape (1.2 msec/byte with the present tape recorder) from the temporary Pet output store.

WRITE240 : Feeds 240 bytes to Pet (date/time, cross and auto for one height).

READ16 : Takes 16 bytes from Pet and puts them in a temporary store.

FILE-GAP : Puts a file gap on tape and waits until it is finished (0.5 sec).

It also takes the present day # and puts it back into the manually initialized store - this is just a safety feature in case someone resets the Apple and forgets to re-initialize the day#.

DISPLAY : Displays a character corresponding to the ASCII graphic for the lowest 4 bits for 17 bytes (designated by a pointer to one byte below the desired bytes, which is initialized manually before running the program).

0-9	come out as	0-9
A	comes out as	:
B	" " "	;
C	" " "	<
D	" " "	=
E	" " "	>
F	" " "	?

3. Method of continuous correlation calculation (4 antennas)

Two consecutive blocks of data (n, n+1) are available in binary form as shown in Fig. 1 (at the beginning and end of the calculation, one of them will be all zeroes). Antenna #4 is correlated with the other three at each lag (because of the special antenna array used)- so just one pair will be described. Ten bytes of antenna #4 are transferred to zero page as shown in Fig. 2. This situation will be arbitrarily called a negative lag, and since #4 is shifted 2 bytes (16 bits) to the right, this is lag -16. An 'AND' operation is performed between the 8 pairs of bytes indicated - the result of an 'AND' contains a '1' bit if the corresponding bits in the pair of bytes 'ANDed' were both 1's, and is a '0' bit otherwise. The number of '1' bits in each result is accumulated for the 8 ANDs and is added to the 4v3 lag -16 accumulator (2 bytes in store 1 cross correlations -see Apple bulk storage map, pg. 35) The zero page antenna #4 sequence (12 bytes, although only 10 used) is now shifted left by 1 bit using the 'ROL' instruction. A zero bit is inserted on the RHS for each set of rolls, so zeroes fill the space left by the shifting sequence. Now the same 8 bytes of antenna #3 are ANDed again for lag -15. Note that the leftmost bit of the antenna #4 sequence is now outside the boundary of the 8 bytes, and does not come into the AND results here.

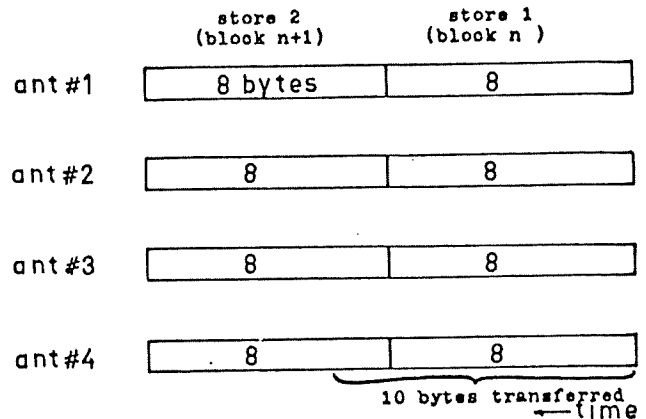


Fig. 1

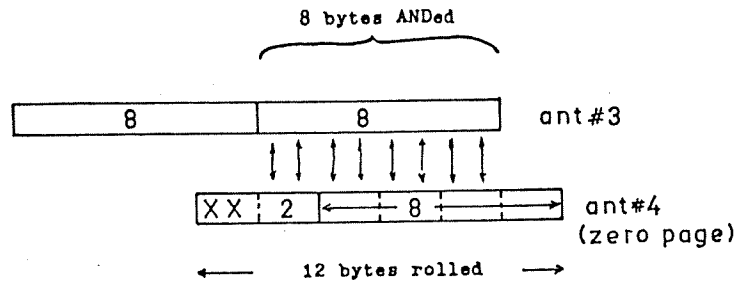


Fig. 2

The procedure continues down to lag -1. Before lag 0 (when antenna #3 and #4 are perfectly aligned), the two extra bytes in antenna #4 (LHS) are zeroed. Now ten bytes are ANDed as shown in Fig. 3 (12 bytes are still rolled for easier programming). For zero and +ve lags, no bytes of antenna #4 will be outside the AND operation, but some of the antenna #3 bits effectively will be since they are being ANDed with zero bits.

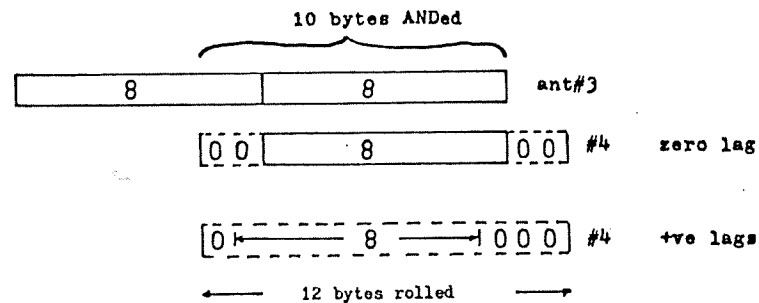
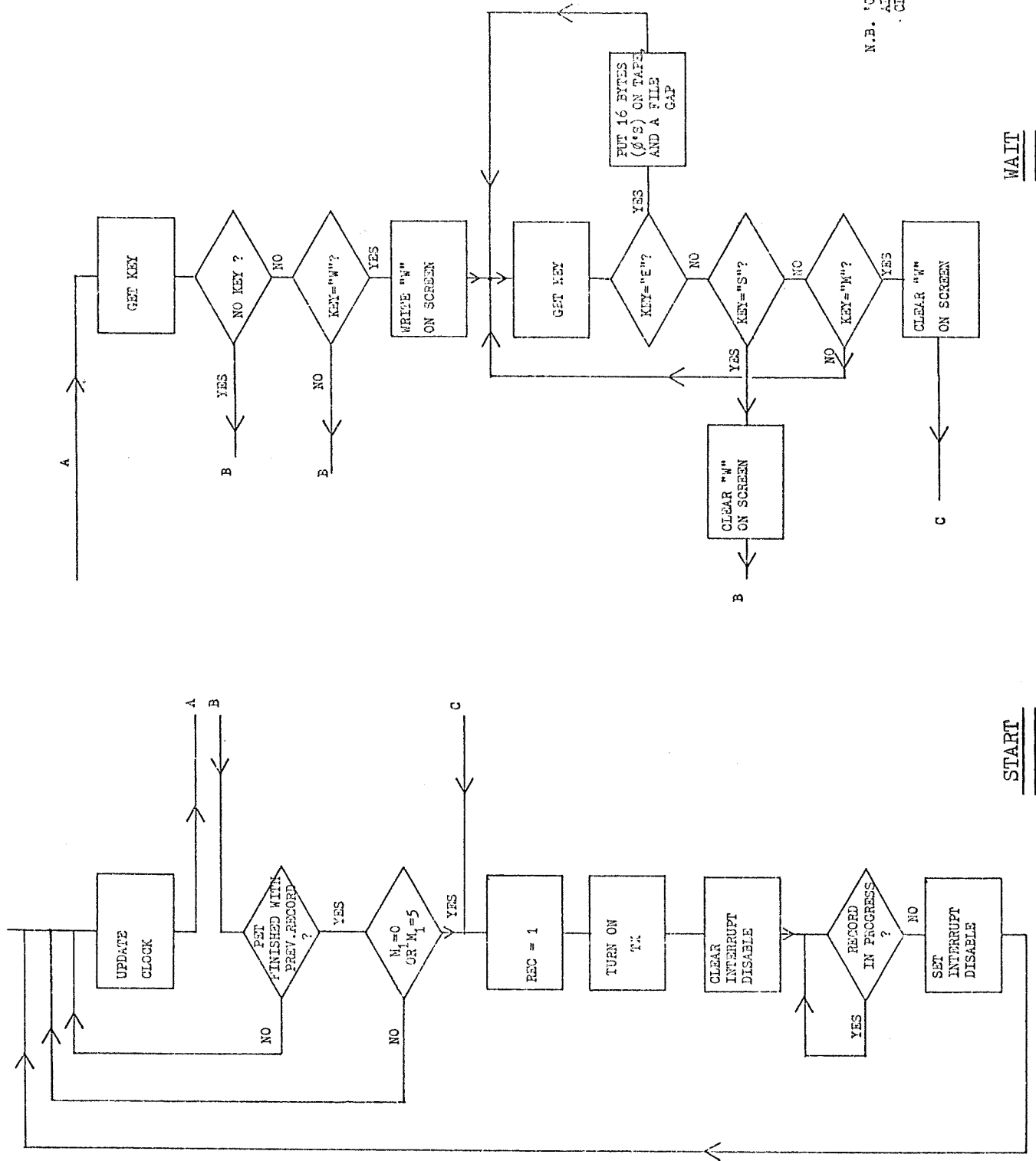


Fig. 3

The auto correlation is almost exactly the same except that only antennas # 1,2,3 are used, and just zero and +ve lags are calculated (the sequences are ANDed with themselves); the results for each lag being added into the same accumulator - effectively giving an 'average' auto correlation at each lag .

At the end, store 2 binary is moved to store 1 and store 2 is zeroed (the latter is actually only required at the end of the record). When the next block has been converted to binary, it is placed in store 2, and the same correlation procedure is followed again. The final result of this process is exactly the same as if the binary sequences for the whole record were correlated at once. The accumulators (auto and cross) contain the number of 1-matches for each lag. The zero lag auto correlation contains the total number of 1's in the three (antenna #1,2,3) sequences - this is used for calculating the actual correlation value from the number of 1-matches (done in the Pet) as described in Appendix B, Section 1 (pg. 57).

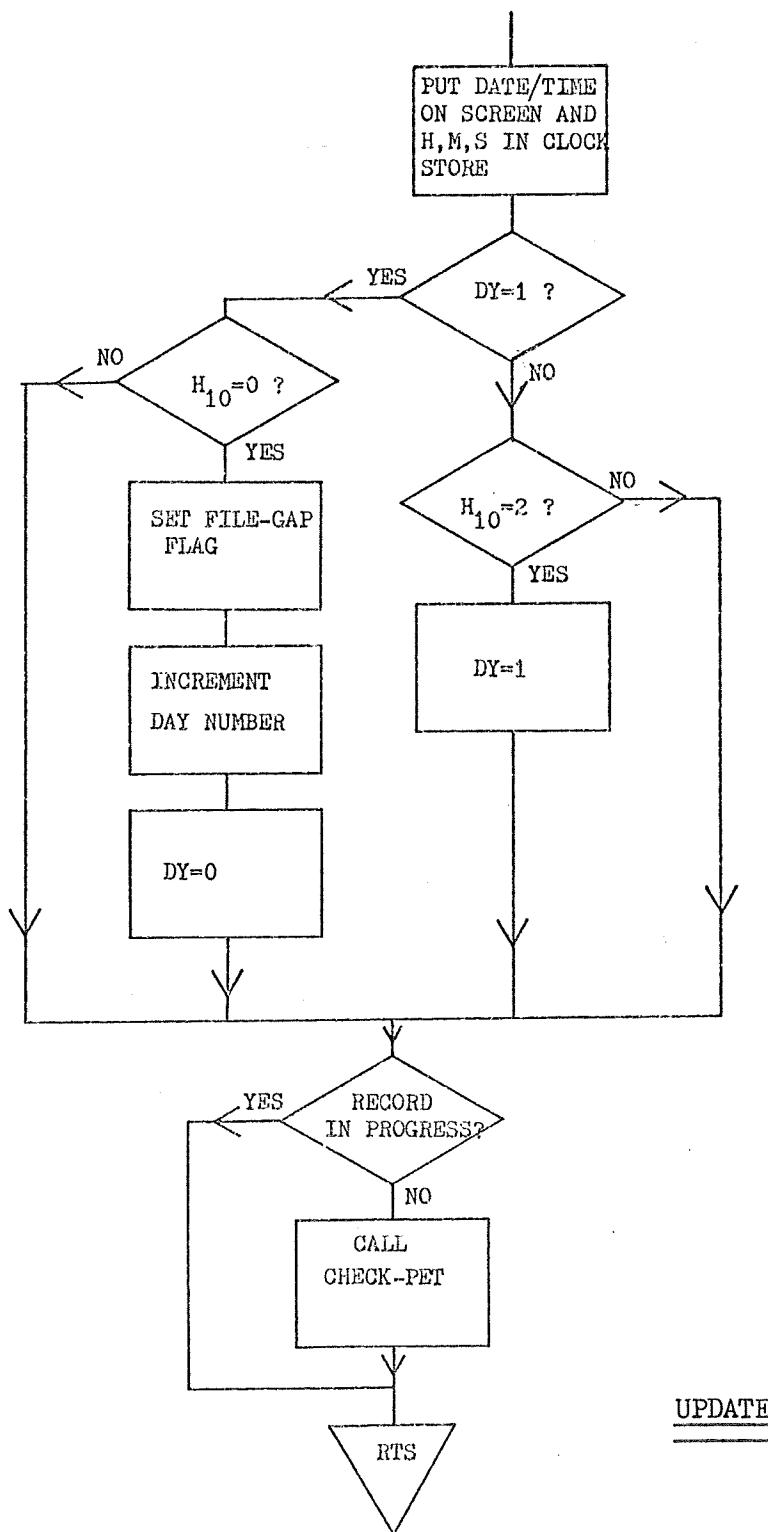
4. Flow charts

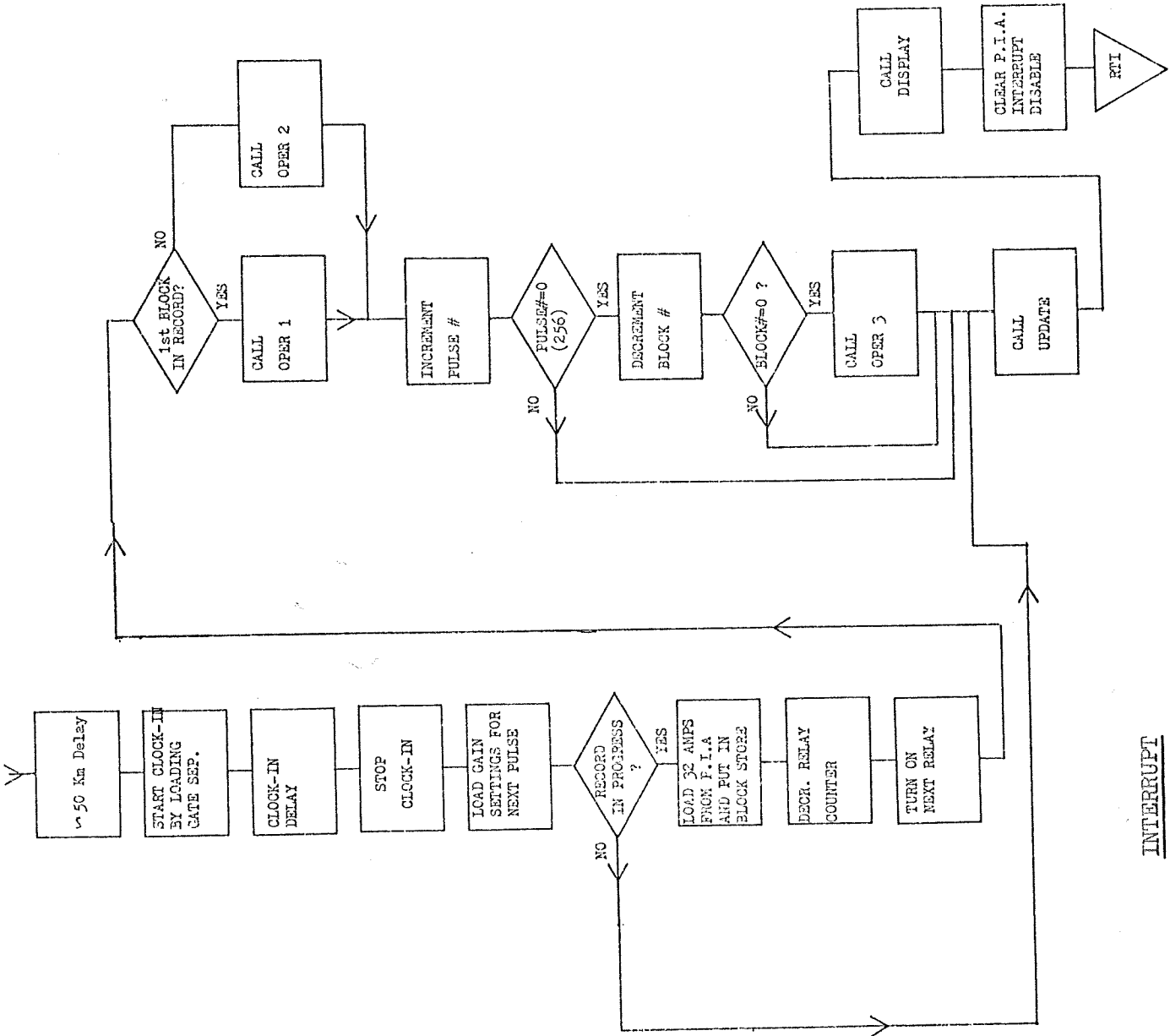


N.B. 'GET KEY' ROUTINE ALSO UPDATES THE CLOCK

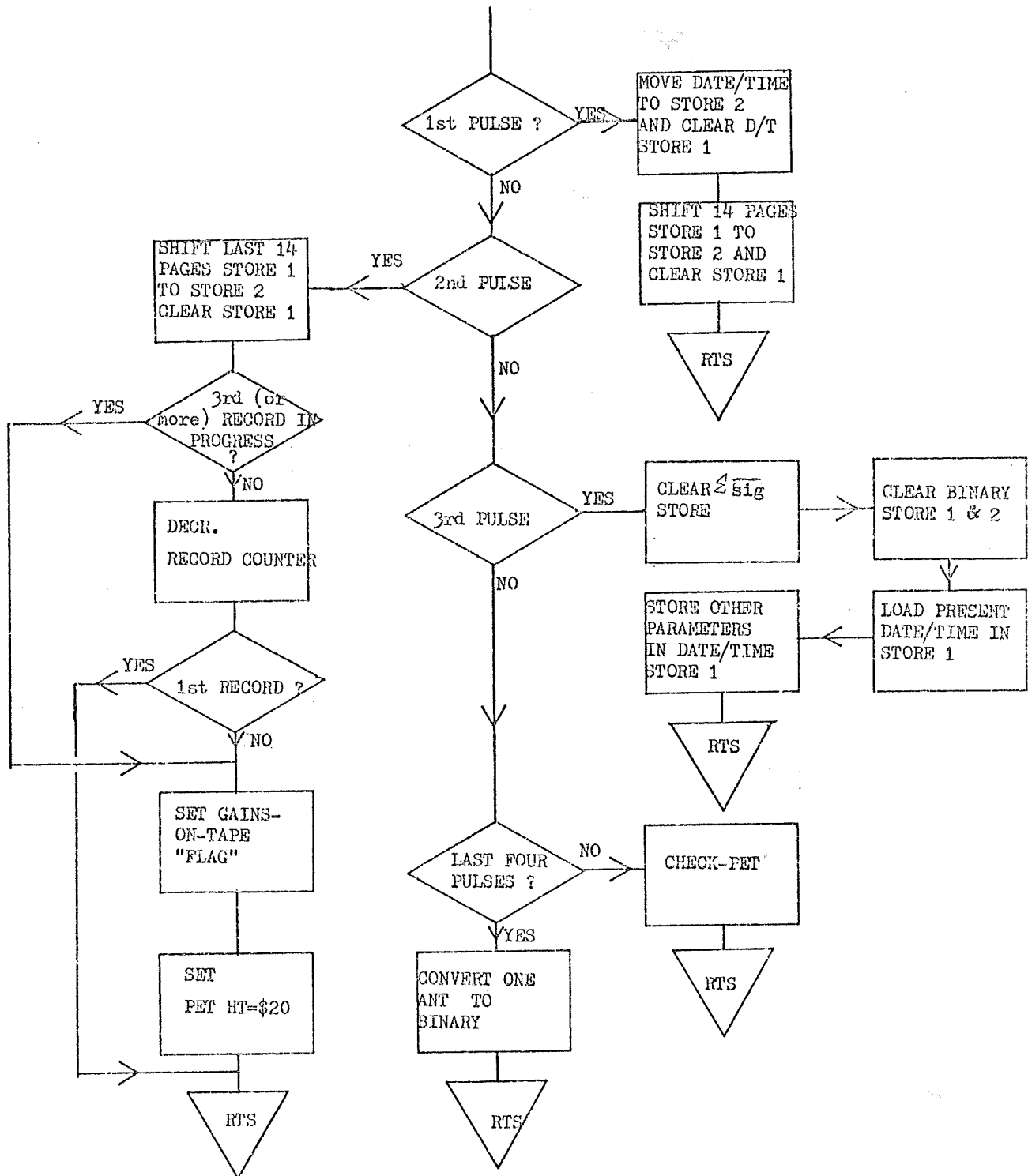
WAIT

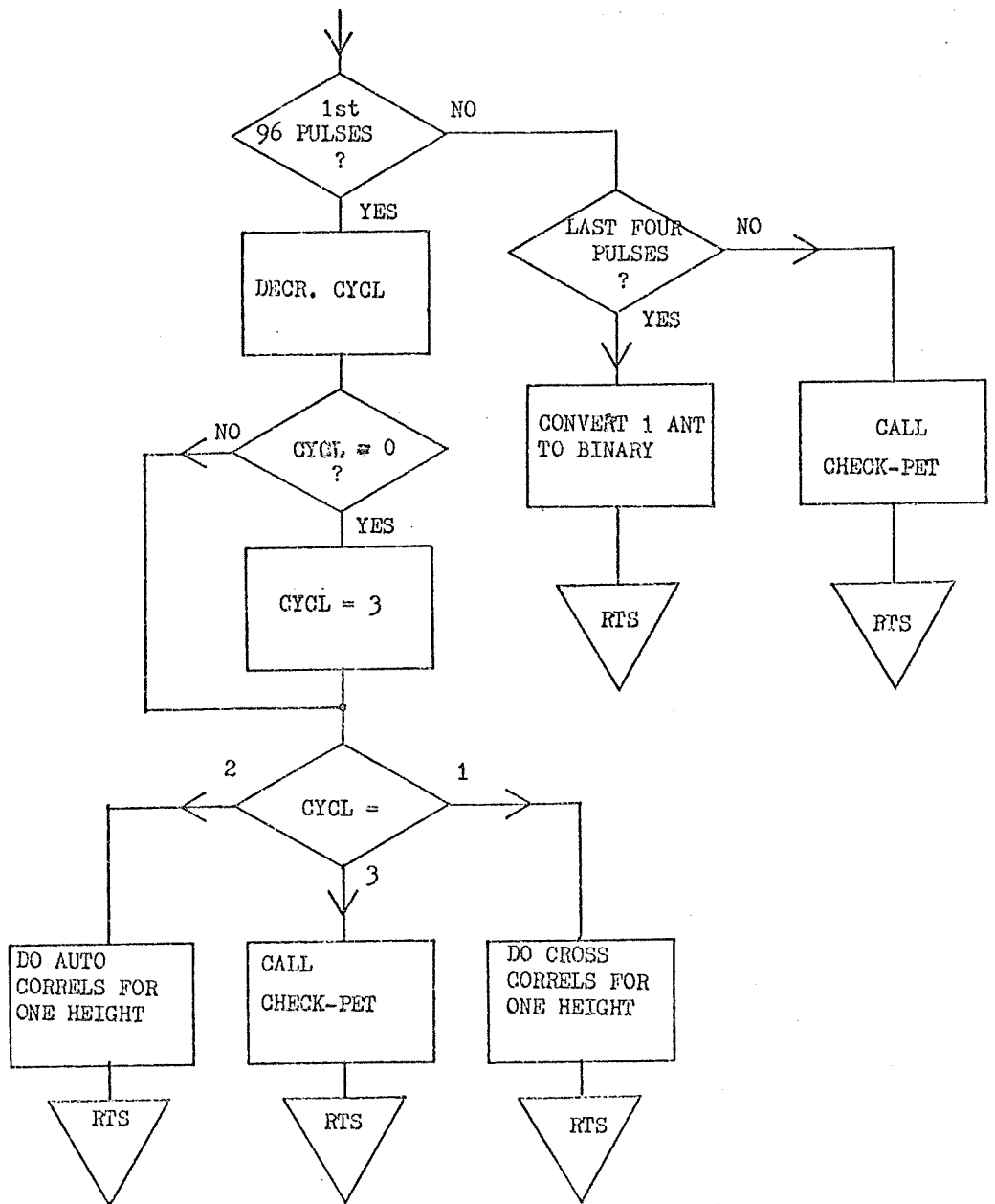
START



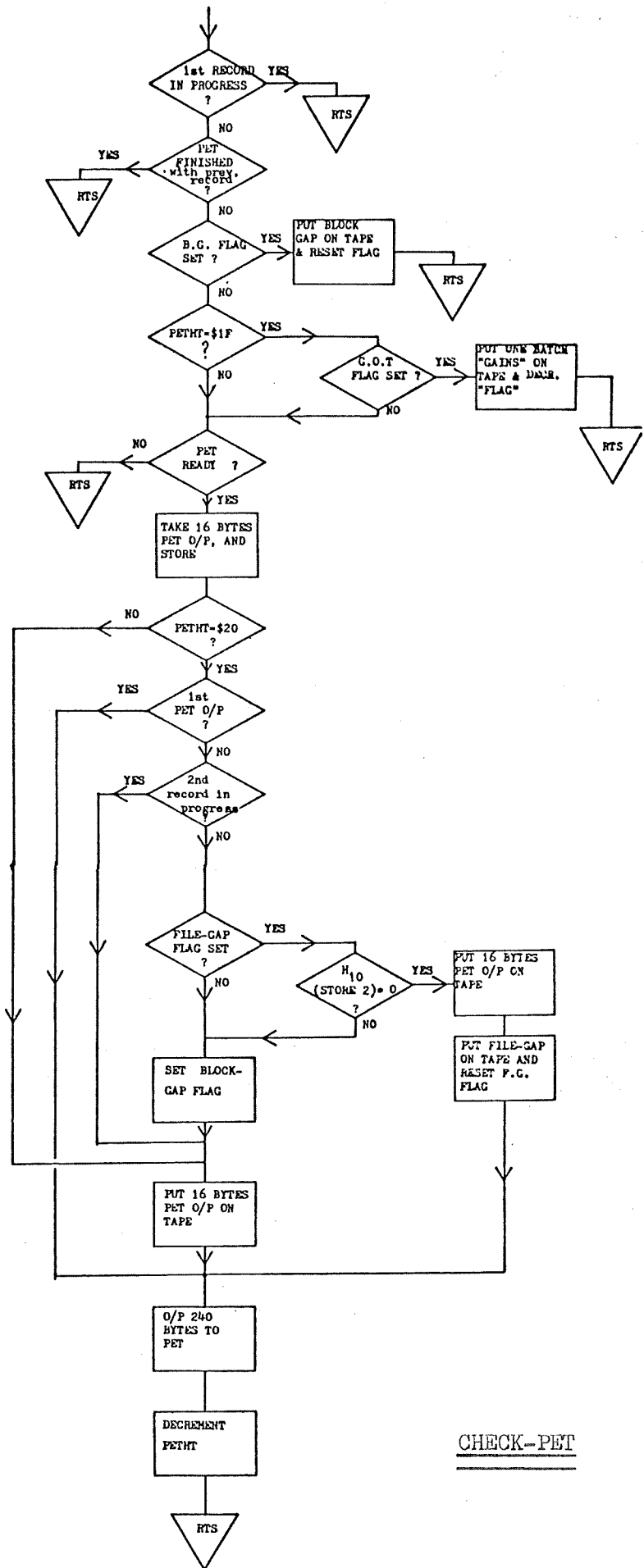


INTERRUPT

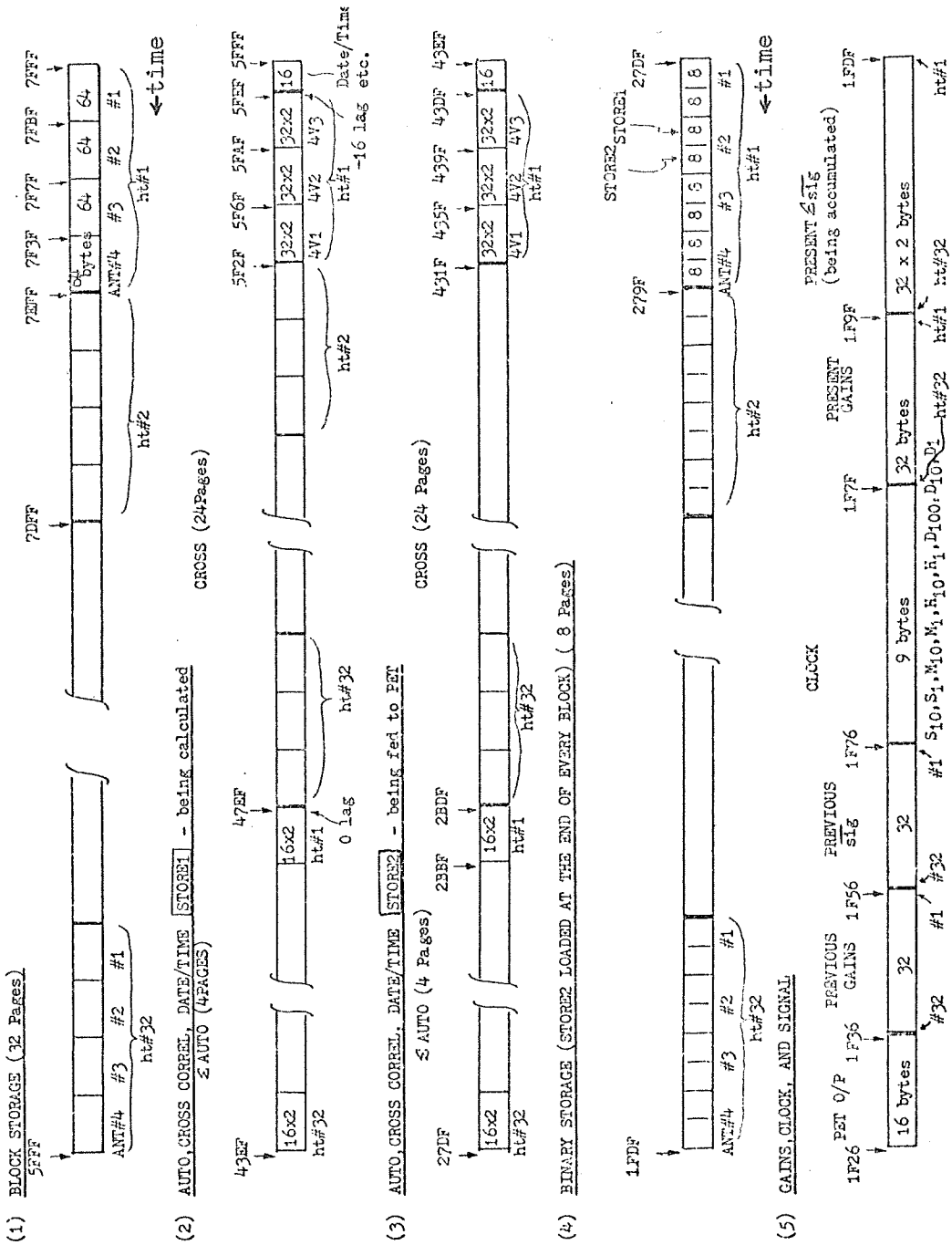




OPER 2



CHECK-PET



5. APPLE BULK STORAGE (the low order byte in two byte accumulators has the higher address)

LOCATION	OPER1	MOVE 14	OPER2	OPER3	ADR-BIN	ADR-XCOR	ADR-ACOR	XCOR	ACOR	DEC-BIN	CHECK-PET	RESET-CAIN	INTER-PUPT
9E	ZANT												
9F	ZD + 1												
A0	ZD + 1		ZANT										
A1	ZD + 1		ZHT										
A2	ZD + 1		ZHT										
A3	ZD + 1		ZHT										
A4	ZD + 1		ZHT										
A5													
A6													
A7													
A8													
A9													
AA													
AB													
AC													
AD													
AE													
AF													
B0													
B1													
B2													
B3													
B4													
B5													
B6													
B7													

6. VOLATILE ZERO-PAGE STORAGE FOR SUBROUTINES

7. ZERO PAGE STORAGE
(CONSTANTS*, CONTROL FLAGS, COUNTERS)

LOCATION	NAME	USE	ROUTINE
00B8	DELAY*	Controls delay before 1st ht. gate	INTERRUPT
B9	GSEP*	Determines gate separation (20 μ s at present)	"
BA	CLOCKIN*	Controls delay for clocking in amplitudes must cover 1st 64 ht. gates exactly	"
BB	REC	= 1 if record in progress	VARIOUS
BC	DY	= 1 if looking for day # update	UPDATE
BD	NBLK	= NBLK1 at record start, decremented to zero, counts # blocks taken	interrupt
BE	NBLK1*	# BLOCKS/RECORD -chosen at system start	"
BF	PLSE	PULSE count (0-255) for each block, also used to calculate addresses, and select operations in OPER1, OPER2	INTERRUPT OPER1, OPER2
C0	NRELAY	decremented with every pulse, used to choose relay #, also used to calculate addresses.	INTERRUPT
C1	NRELAY1*	= 4 (# ANTENNAS USED), used to reload NRELAY	INTERRUPT
C2	PET HT	= \$20 when STORE2(AUTO & CROSS) filled, decremented to zero as PET processes these.	CHECK-PET
C3	B.G. FLAG	= 1 if block gap required	CHECK-PET
C4	RECORD#	indicates whether 1st or 2nd record in progress	CHECK-PET & others
C5	1stPET O/P FLAG	= 1 until 1st PET O/P after system start	
C6	G.O.T. FLAG	= 2,1,0 - controls writing date/time, sig, gains on tape	GAINS-ON-TAPE CHECK-PET
C7	F.G.FLAG	= 1 if file gap required	UPDATE, CHECK-PET
C8	CYCL	controls subroutine choice in OPER2	OPER2
3FE	L	} INTERRUPT VECTOR - points to interrupt routine	} INITIALIZE
3FF	H		
0	#BLOCKS/RECORD	} INITIALIZED MANUALLY AT SYSTEM START. This day # is updated everytime a file gap is put on tape.	} INITIALIZE
1	D ₁₀₀		
2	D ₁₀		
3	D ₁		
A	L	} pointer to one byte below chosen display bytes - INITIALIZED MANUALLY AT START	} DISPLAY
B	H		
C	L	} USED IN <u>DISPLAY</u> TO POINT TO SCREEN LOCATION	}
D	H		

Address	Hex	Op	Hex	Comment
08FC-	20 20 09	JSR	\$0920	<u>START</u>
08FF-	4C 75 12	JMP	\$1275	
0902-	A5 C2	LDA	\$C2	
0904-	D0 F6	BNE	\$08FC	
0906-	AD 7A 1F	LDA	\$1F7A	start record if M ₁ = 0 or 5
0909-	F0 04	BEQ	\$090F	
090B-	C9 05	CMP	#\$05	
090D-	D0 ED	BNE	\$08FC	
090F-	E6 BB	INC	\$BB	
0911-	A9 3C	LDA	#\$3C	TURN ON Tx
0913-	8D 01 C5	STA	\$C501	
0916-	58	CLI		
0917-	A5 BB	LDA	\$BB	tight loop when record in progress
0919-	D0 FC	BNE	\$0917	
091B-	78	SEI		
091C-	4C FC 08	JMP	\$08FC	
091F-	00	BRK		
0920-	A9 00	LDA	#\$00	<u>UPDATE</u>
0922-	20 D0 09	JSR	\$09D0	
0925-	8D 04 05	STA	\$0504	H ₁₀
0928-	29 0F	AND	#\$0F	
092A-	8D 7B 1F	STA	\$1F7B	
092D-	A9 10	LDA	#\$10	
092F-	20 D0 09	JSR	\$09D0	
0932-	8D 05 05	STA	\$0505	H ₁
0935-	29 0F	AND	#\$0F	
0937-	8D 7C 1F	STA	\$1F7C	
093A-	A9 BA	LDA	#\$BA	
093C-	8D 06 05	STA	\$0506	
093F-	A9 20	LDA	#\$20	
0941-	20 D0 09	JSR	\$09D0	
0944-	8D 07 05	STA	\$0507	M ₁₀
0947-	29 0F	AND	#\$0F	
0949-	8D 79 1F	STA	\$1F79	
094C-	A9 60	LDA	#\$60	
094E-	20 D0 09	JSR	\$09D0	
0951-	8D 08 05	STA	\$0508	M ₁
0954-	29 0F	AND	#\$0F	
0956-	8D 7A 1F	STA	\$1F7A	
0959-	A9 BA	LDA	#\$BA	
095B-	8D 09 05	STA	\$0509	
095E-	A9 50	LDA	#\$50	
0960-	20 D0 09	JSR	\$09D0	
0963-	8D 0A 05	STA	\$050A	S ₁₀
0966-	29 0F	AND	#\$0F	
0968-	8D 77 1F	STA	\$1F77	
096B-	A9 40	LDA	#\$40	
096D-	20 D0 09	JSR	\$09D0	
0970-	8D 0B 05	STA	\$050B	S ₁
0973-	29 0F	AND	#\$0F	
0975-	8D 78 1F	STA	\$1F78	
0978-	AD 7D 1F	LDA	\$1F7D	
097B-	09 B0	ORA	#\$B0	D ₁₀₀
097D-	8D 00 05	STA	\$0500	
0980-	AD 7E 1F	LDA	\$1F7E	
0983-	09 B0	ORA	#\$B0	D ₁₀
0985-	8D 01 05	STA	\$0501	
0988-	AD 7F 1F	LDA	\$1F7F	
098B-	09 B0	ORA	#\$B0	D ₁
098D-	8D 02 05	STA	\$0502	
0990-	A9 BA	LDA	#\$BA	
0992-	8D 03 05	STA	\$0503	
0995-	EA	NOP		
0996-	EA	NOP		
0997-	EA	NOP		
0998-	A5 BC	LDA	\$BC	
099A-	D0 09	BNE	\$09A5	check for Day # update, set file gap flag when Day # changes
099C-	AD 7B 1F	LDA	\$1F7B	
099F-	C9 02	CMP	#\$02	
09A1-	D0 22	BNE	\$09C5	
09A3-	E6 BC	INC	\$BC	
09A5-	AD 7B 1F	LDA	\$1F7B	
09A8-	D0 1B	BNE	\$09C5	
09AA-	E6 C7	INC	\$C7	
09AC-	A2 03	LDX	#\$03	
09AE-	BD 7C 1F	LDA	\$1F7C,X	

09B1-	18		CLC	
09B2-	69 01		ADC	##01
09B4-	9D 7C 1F		STA	\$1F7C,X
09B7-	C9 0A		CMP	##0A
09B9-	D0 08		BNE	\$09C3
09BB-	A9 00		LDA	##00
09BD-	9D 7C 1F		STA	\$1F7C,X
09C0-	CA		DEX	
09C1-	D0 EB		BNE	\$09AE
09C3-	C6 BC		DEC	\$BC
09C5-	A5 BB		LDA	\$BB
09C7-	F0 01		BEQ	\$09CA
09C9-	60		RTS	
09CA-	20 9E 10		JSR	\$109E
09CD-	60		RTS	
09CE-	00		BRK	
09CF-	00		BRK	
09D0-	8D 02 C3		STA	\$C302
09D3-	AD 02 C3		LDA	\$C302
09D6-	29 0F		AND	##0F
09D8-	8D 02 C3		STA	\$C302
09DB-	09 B0		ORA	##B0
09DD-	60		RTS	
09DE-	00		BRK	
09DF-	00		BRK	
09E0-	00		BRK	
09E1-	00		BRK	
09E2-	00		BRK	
09E3-	00		BRK	
09E4-	00		BRK	
09E5-	00		BRK	
09E6-	00		BRK	
09E7-	00		BRK	
09E8-	00		BRK	
09E9-	00		BRK	
09EA-	78		SEI	
09EB-	D8		CLD	
09EC-	A9 07		LDA	##07
09EE-	A2 20		LDX	##20
09F0-	A0 04		LDY	##04
09F2-	9D 7F 1F		STA	\$1F7F,X
09F5-	CA		DEX	
09F6-	88		DEY	
09F7-	D0 F9		BNE	\$09F2
09F9-	38		SEC	
09FA-	E9 01		SBC	##01
09FC-	10 F2		BPL	\$09F0
09FE-	A5 00		LDA	\$00
0A00-	85 B0		STA	\$B0
0A02-	85 BE		STA	\$BE
0A04-	A9 39		LDA	##39
0A06-	85 B8		STA	\$B8
0A08-	A9 08		LDA	##08
0A0A-	85 B9		STA	\$B9
0A0C-	A5 01		LDA	\$01
0A0E-	8D 7D 1F		STA	\$1F7D
0A11-	A5 02		LDA	\$02
0A13-	8D 7E 1F		STA	\$1F7E
0A16-	A5 03		LDA	\$03
0A18-	8D 7F 1F		STA	\$1F7F
0A1B-	A9 14		LDA	##14
0A1D-	85 BA		STA	\$BA
0A1F-	A9 00		LDA	##00
0A21-	8D 01 C5		STA	\$C501
0A24-	8D 03 C5		STA	\$C503
0A27-	8D 05 C5		STA	\$C505
0A2A-	8D 07 C5		STA	\$C507
0A2D-	8D 01 C7		STA	\$C701
0A30-	8D 03 C7		STA	\$C703
0A33-	8D 03 C3		STA	\$C303
0A36-	A9 FF		LDA	##FF
0A38-	8D 04 C5		STA	\$C504
0A3B-	8D 06 C5		STA	\$C506
0A3E-	8D 02 C5		STA	\$C502
0A41-	8D 02 C7		STA	\$C702

If Record not in progress then "check-pet"

GNV

INITIALIZE

Set initial gains (arbitrary)
ht#0-3 have gain 7
ht#4-7 have gain 6
etc.

delay to first gate

gate separation

load day #

clock in delay

P.I.A. initialization

0A44-	A9 00	LDA	##00	
0A46-	8D 00 C5	STA	\$C500	
0A49-	8D 00 C7	STA	\$C700	
0A4C-	A9 F0	LDA	##F0	
0A4E-	8D 02 C3	STA	\$C302	
0A51-	A9 2D	LDA	##2D	
0A53-	8D 01 C5	STA	\$C501	
0A56-	8D 05 C5	STA	\$C505	
0A59-	8D 07 C5	STA	\$C507	
0A5C-	8D 01 C7	STA	\$C701	
0A5F-	8D 03 C7	STA	\$C703	
0A62-	8D 03 C3	STA	\$C303	
0A65-	8D 03 C5	STA	\$C503	
0A69-	A0 02	LDY	##02	
0A6A-	A2 20	LDX	##20	
0A6C-	BD 7F 1F	LDA	\$1F7F.X	
0A6F-	09 C0	ORA	##C0	
0A71-	8D 02 C7	STA	\$C702	
0A74-	CA	DEX		
0A75-	D0 F5	BNE	\$0A6C	
0A77-	88	DEY		
0A78-	D0 F0	BNE	\$0A6A	
0A7A-	A0 10	LDY	##10	
0A7C-	A9 00	LDA	##00	
0A7E-	99 EF 5F	STA	\$5FEF.Y	
0A81-	88	DEY		
0A82-	D0 FA	BNE	\$0A7E	
0A84-	85 BB	STA	\$BB	
0A86-	85 C2	STA	\$C2	
0A88-	85 BC	STA	\$BC	
0A8A-	A9 04	LDA	##04	← 4 relays (Antennas)
0A8C-	85 C0	STA	\$C0	
0A8E-	85 C1	STA	\$C1	
0A90-	A9 01	LDA	##01	
0A92-	8D 02 C5	STA	\$C502	
0A95-	A9 C0	LDA	##C0	
0A97-	8D FE 03	STA	\$03FE	
0A9A-	A9 0A	LDA	##0A	
0A9C-	8D FF 03	STA	\$03FF	
0A9F-	A9 00	LDA	##00	
0AA1-	85 BF	STA	\$BF	
0AA3-	A9 03	LDA	##03	
0AA5-	85 C4	STA	\$C4	
0AA7-	A9 01	LDA	##01	
0AA9-	85 C5	STA	\$C5	
0AAB-	A9 00	LDA	##00	
0AAD-	85 C6	STA	\$C6	
0AAF-	85 C3	STA	\$C3	
0AB1-	85 C7	STA	\$C7	
0AB3-	A9 34	LDA	##34	
0AB5-	8D 01 C5	STA	\$C501	
0AB8-	A9 01	LDA	##01	
0ABA-	85 C8	STA	\$C8	
0ABC-	4C FC 08	JMP	\$08FC	
0ABF-	00	BRK		
<hr/>				
0AC0-	48	PHA		
0AC1-	8A	TXA		
0AC2-	48	PHA		
0AC3-	A4 B8	LDY	\$B8	
0AC5-	88	DEY		
0AC6-	D0 FD	BNE	\$0AC5	
0AC8-	A5 B9	LDA	\$B9	
0ACA-	8D 02 C7	STA	\$C702	
0ACD-	A4 BA	LDY	\$BA	
0ACF-	A2 0B	LDX	##0B	
0AD1-	CA	DEX		
0AD2-	D0 FD	BNE	\$0AD1	
0AD4-	EA	NOP		
0AD5-	88	DEY		
0AD6-	D0 F7	BNE	\$0ACF	
0AD8-	A2 03	LDX	##03	
0ADA-	CA	DEX		
0ADB-	D0 FD	BNE	\$0ADA	
0ADD-	A9 80	LDA	##80	
0ADF-	8D 02 C7	STA	\$C702	

Run gain setting into PIA (twice!)

zero date/time store 1.

set interrupt vector

pulse count

initial record counter

turn off Tx (in case it was on)

INTERRUPT

delay

turn on clocking at chosen gate sep.

clock in delay

STOP clocking

0AE2-	A0 02	LDY	#\$02	
0AE4-	A2 20	LDX	##20	
0AE6-	B0 7F IF	LDA	#\$F7F,X	load gain setting for next pulse
0AE9-	09 C0	ORA	##C0	
0AEB-	8D 02 C7	STA	##C702	
0AEE-	CA	DEX		
0AEF-	D0 F5	BNE	##0AE6	
0AF1-	88	DEY		
0AF2-	D0 F0	BNE	##0AE4	
0AF4-	A5 BB	LDA	##BB	if a record is not in progress, skip out
0AF6-	D0 03	BNE	##0AFB	
0AF8-	18	CLC		
0AF9-	90 5C	BCC	##0B57	
0AFB-	D8	CLD		
0AFC-	A9 7F	LDA	##7F	
0AFE-	85 A6	STA	##A6	
0B00-	A5 BF	LDA	##BF	
0B02-	4A	LSR		calculate data address
0B03-	4A	LSR		
0B04-	85 A5	STA	##A5	
0B06-	A9 00	LDA	##00	
0B08-	A6 C0	LDX	##C0	
0B0A-	18	CLC		
0B0B-	69 40	ADC	##40	
0B0D-	CA	DEX		
0B0E-	D0 FA	BNE	##0B0A	
0B10-	18	CLC		
0B11-	E5 A5	SBC	##A5	
0B13-	85 A5	STA	##A5	
0B15-	A0 20	LDY	##20	read data from PIA and STORE IN BLOCK STORAGE (N.B. A/D conv makes anything >255 equal zero - thus this adjustment)
0B17-	AD 00 C7	LDA	##C700	
0B1A-	D0 02	BNE	##0B1E	
0B1C-	A9 FF	LDA	##FF	
0B1E-	81 A5	STA	##(A5,X)	
0B20-	C6 A6	DEC	##A6	
0B22-	88	DEY		
0B23-	D0 F2	BNE	##0B17	
0B25-	C6 C0	DEC	##C0	
0B27-	D0 04	BNE	##0B2D	
0B29-	A5 C1	LDA	##C1	select next relay
0B2B-	85 C0	STA	##C0	
0B2D-	A9 05	LDA	##05	
0B2F-	38	SEC		
0B30-	E5 C0	SBC	##C0	
0B32-	AA	TAX		
0B33-	A9 00	LDA	##00	
0B35-	38	SEC		
0B36-	2A	ROL		
0B37-	CA	DEX		
0B38-	D0 FC	BNE	##0B36	
0B3A-	8D 02 C5	STA	##C502	Relay #1 2 3 4 Loaded#1,2,4,8
0B3D-	A5 BE	LDA	##BE	
0B3F-	C5 BD	CMP	##BD	
0B41-	D0 06	BNE	##0B49	choose subroutine depending on Block #
0B43-	20 70 0B	JSR	##0B70	
0B46-	18	CLC		
0B47-	90 03	BCC	##0B4C	
0B49-	20 2C 0C	JSR	##0C2C	
0B4C-	E6 BF	INC	##BF	
0B4E-	D0 07	BNE	##0B57	
0B50-	C6 BD	DEC	##BD	
0B52-	D0 03	BNE	##0B57	
0B54-	20 75 0C	JSR	##0C75	
0B57-	20 20 09	JSR	##0920	UPDATE DISPLAY
0B5A-	20 5D 12	JSR	##125D	
0B5D-	AD 02 C7	LDA	##C702	RESET INTERRUPT enable on PIA
0B60-	68	PLA		
0B61-	AA	TAX		
0B62-	68	PLA		
0B63-	40	RTI		
0B64-	00	BRK		
0B65-	00	BRK		
0B66-	00	BRK		
0B67-	00	BRK		
0B68-	00	BRK		
0B69-	00	BRK		
0B6B-	00	BRK		
0B6C-	00	BRK		
0B6D-	00	BRK		
0B6E-	00	BRK		
0B6F-	00	BRK		

OPER 1 (done during 1st block in record)

```

0B70- D8 CLD
0B71- A5 BF LDA $BF
0B73- D0 24 BNE $0B99 } check pulse
0B75- A0 10 LDY #10 } count
0B77- B9 EF 5F LDA $5FEF,Y } shift store 1
0B7A- 99 DF 43 STA $43DF,Y } date/time
0B7D- A9 00 LDA #00 } to store 2 & zero store 1
0B7F- 99 EF 5F STA $5FEF,Y
0B82- 88 DEY
0B83- D0 F2 BNE $0B77
0B85- A9 5E LDA #$5E
0B87- 85 A2 STA $A2
0B89- A9 F0 LDA #$F0 } shift first 14 pages store 1 to store 2
0B8B- 85 A1 STA $A1 } (auto & cross) and zero store 1
0B8D- A9 42 LDA #$42
0B8F- 85 A4 STA $A4
0B91- A9 E0 LDA #$E0
0B93- 85 A3 STA $A3
0B95- 20 14 0C JSR $0C14
0B98- 60 RTS
0B99- C9 01 CMP #01
0B9B- D0 16 BNE $0BB3 } shift and zero last 14 pages
0B9D- 20 14 0C JSR $0C14
0BA0- A5 C4 LDA $C4
0BA2- F0 06 BEQ $0BAA } check record count
0BA4- C6 C4 DEC $C4
0BA6- C9 03 CMP #03
0BA8- F0 08 BEQ $0BB2
0BAA- A9 02 LDA #02 } gains on tape flag
0BAC- 85 C6 STA $C6
0BAE- A9 20 LDA #20
0BB0- 85 C2 STA $C2
0BB2- 60 RTS
0BB3- C9 02 CMP #02
0BB5- D0 42 BNE $0BF9
0BB7- A0 40 LDY #40
0BB9- A9 00 LDA #00
0BBB- 99 9F 1F STA $1F9F,Y } zero sig store
0BBE- 88 DEY
0BBF- D0 FA BNE $0BBB
0BC1- A9 E0 LDA #$E0
0BC3- 85 A1 STA $A1
0BC5- A9 26 LDA #$26 } zero binary store 1 & 2
0BC7- 85 A2 STA $A2
0BC9- A2 08 LDX #08
0BCB- A9 00 LDA #00
0BCD- A0 00 LDY #00
0BCF- 91 A1 STA ($A1),Y
0BD1- 88 DEY
0BD2- D0 FB BNE $0BCF
0BD4- C6 A2 DEC $A2
0BD6- CA DEX
0BD7- D0 F4 BNE $0BCD
0BD9- A2 07 LDX #07
0BDB- BD 78 1F LDA $1F78,X } load present time (& other parameters)
0BDE- 9D F8 5F STA $5FF8,X } into date/time store 1
0BE1- CA DEX
0BE2- D0 F7 BNE $0BDB
0BE4- A5 BE LDA $BE
0BE6- 8D F8 5F STA $5FF8
0BE9- A5 B8 LDA $B8
0BEB- 8D F7 5F STA $5FF7
0BEE- A5 B9 LDA $B9
0BF0- 8D F6 5F STA $5FF6
0BF3- A5 BA LDA $BA
0BF5- 8D F5 5F STA $5FF5
0BF8- 60 RTS
0BF9- C9 FC CMP #FC
0BFB- B0 04 BCS $0C01
0BFD- 20 9E 10 JSR $109E CHECK-PBT
0C00- 60 RTS
0C01- 38 SEC
0C02- E9 FC SBC #FC
0C04- 85 A0 STA $A0 } ANT #
0C06- A9 00 LDA #00 } HT #
0C08- 85 A1 STA $A1 } (Convert decimal to binary as each
0C0A- 20 C8 0C JSR $0CC8 } ADR-BLN antenna is finished in the block
0C0D- 20 78 0D JSR $0D78 } DEC-BLN
0C10- 60 RTS

```

0C11-	00	BRK		
0C12-	00	BRK		
0C13-	00	BRK		
0C14-	A2 0E	LDX	##0E	
0C16-	A0 00	LDY	##00	<u>MOVE 14</u>
0C18-	B1 A1	LDA	(&A1),Y	Move 14 pages store 1 auto & cross
0C1A-	91 A3	STA	(&A3),Y	correlts to store 2 and zero store 1
0C1C-	A9 00	LDA	##00	
0C1E-	91 A1	STA	(&A1),Y	
0C20-	88	DEY		
0C21-	D0 F5	BNE	\$0C18	
0C23-	C6 A2	DEC	\$A2	
0C25-	C6 A4	DEC	\$A4	
0C27-	CA	DEX		
0C28-	D0 EC	BNE	\$0C16	
0C2A-	60	RTS		
0C2B-	00	BRK		
0C2C-	A5 BF	LDA	\$BF	
0C2E-	C9 60	CMP	##60	<u>OPER 2</u>
0C30-	B0 28	BCS	\$0C5A	
0C32-	C6 C8	DEC	\$C8	
0C34-	D0 04	BNE	\$0C3A	
0C36-	A9 03	LDA	##03	
0C38-	85 C8	STA	\$C8	
0C3A-	A5 C8	LDA	\$C8	
0C3C-	C9 02	CMP	##02	
0C3E-	F0 0F	BEQ	\$0C4F	
0C40-	90 1C	BCC	\$0C5E	
0C42-	C6 9E	DEC	\$9E	
0C44-	A5 9E	LDA	\$9E	
0C46-	85 A1	STA	\$A1	
0C48-	20 C8 0C	JSR	\$0CC8	ADR-BIN
0C4B-	20 27 0E	JSR	\$0E27	XCOR
0C4E-	60	RTS		
0C4F-	A5 9E	LDA	\$9E	
0C51-	85 A1	STA	\$A1	
0C53-	20 C8 0C	JSR	\$0CC8	ADR-BIN
0C56-	20 38 0F	JSR	\$0F38	ACOR
0C59-	60	RTS		
0C5A-	C9 FC	CMP	##FC	
0C5C-	B0 04	BCS	\$0C62	
0C5E-	20 9E 10	JSR	\$109E	CHECK-PET
0C61-	60	RTS		
0C62-	38	SEC		
0C63-	E9 FC	SBC	##FC	
0C65-	85 A0	STA	\$A0	convert decimal to binary
0C67-	A9 00	LDA	##00	
0C69-	85 A1	STA	\$A1	
0C6B-	20 C8 0C	JSR	\$0CC8	
0C6E-	20 78 0D	JSR	\$0D78	
0C71-	60	RTS		
0C72-	00	BRK		
0C73-	00	BRK		
0C74-	00	BRK		
0C75-	A9 02	LDA	##02	
0C77-	85 9E	STA	\$9E	
0C79-	A9 1F	LDA	##1F	<u>OPER 3</u>
0C7B-	48	PHA		Run over XCOR & ACOR twice to finish up
0C7C-	85 A1	STA	\$A1	HT# counter
0C7E-	20 C8 0C	JSR	\$0CC8	
0C81-	EA	NOP		
0C82-	EA	NOP		
0C83-	EA	NOP		
0C84-	20 27 0E	JSR	\$0E27	
0C87-	68	PLA		
0C88-	85 A1	STA	\$A1	
0C8A-	48	PHA		
0C8B-	20 C8 0C	JSR	\$0CC8	
0C8E-	EA	NOP		
0C8F-	EA	NOP		
0C90-	EA	NOP		
0C91-	20 38 0F	JSR	\$0F38	
0C94-	68	PLA		
0C95-	38	SEC		
0C96-	E9 01	SBC	##01	
0C98-	10 E1	BPL	\$0C7B	


```

0C9A-  C6 9E      DEC  $9E
0C9C-  D0 DB      BNE  $0C79
0C9E-  20 2C 10   JSR  $102C      RESET GAINS
0CA1-  A0 02      LDY  #$02
0CA3-  A2 20      LDX  #$20
0CA5-  BD 7F 1F   LDA  $1F7F,X
0CA8-  09 C0      ORA  #$C0
0CAA-  8D 02 C7   STA  $C702      LOAD P.I.A. with NEW GAINS
0CAD-  CA        DEX
0CAE-  D0 F5      BNE  $0CA5
0CB0-  88        DEY
0CB1-  D0 F0      BNE  $0CA3
0CB3-  A9 00      LDA  #$00
0CB5-  85 BB      STA  $BB
0CB7-  A5 BE      LDA  $BE
0CB9-  85 BD      STA  $BD
0CBB-  A9 34      LDA  #$34
0CBD-  8D 01 C5   STA  $C501      TURN OFF Tx
0CC0-  60        RTS
0CC1-  00        BRK
0CC2-  00        BRK
0CC3-  00        BRK
0CC4-  00        BRK
0CC5-  00        BRK
0CC6-  00        BRK
0CC7-  00        BRK

```

```

0CC8-  A5 A1      LDA  $A1 ← HT#      ADR-BIN
0CCA-  4A        LSR
0CCB-  4A        LSR
0CCC-  85 AD      STA  $AD
0CCE-  0A        ASL
0CCF-  0A        ASL
0CD0-  85 A2      STA  $A2
0CD2-  A9 27      LDA  #$27
0CD4-  38        SEC
0CD5-  E5 AD      SBC  $AD
0CD7-  85 AD      STA  $AD
0CD9-  A9 A7      LDA  #$A7
0CDB-  85 AC      STA  $AC
0CDD-  A5 A1      LDA  $A1
0CDF-  38        SEC
0CE0-  E5 A2      SBC  $A2
0CE2-  F0 0F      BEQ  $0CF3
0CE4-  AA        TAX
0CE5-  A5 AC      LDA  $AC
0CE7-  38        SEC
0CE8-  E9 40      SBC  #$40
0CEA-  85 AC      STA  $AC
0CEC-  B0 02      BCS  $0CF0
0CEE-  C6 AD      DEC  $AD
0CF0-  CA        DEX
0CF1-  D0 F4      BNE  $0CE7
0CF3-  60        RTS
0CF4-  00        BRK
0CF5-  00        BRK
0CF6-  00        BRK
0CF7-  00        BRK
0CF8-  00        BRK
0CF9-  00        BRK
0CFA-  00        BRK
0CFB-  00        BRK
0CFC-  00        BRK
0CFD-  00        BRK

```

```

0CFE-  A5 A1      LDA  $A1 ← HT#      ADR-XCOR
0D00-  4A        LSR
0D01-  4A        LSR
0D02-  4A        LSR
0D03-  0A        ASL
0D04-  0A        ASL
0D05-  85 AF      STA  $AF
0D07-  0A        ASL
0D08-  85 A2      STA  $A2
0D0A-  A5 AF      LDA  $AF
0D0C-  4A        LSR
0D0D-  18        CLC
0D0E-  65 AF      ADC  $AF

```

Calculate address for binary storage, given HT#

Calculate address for accumulating Xcorrels given HT#

0010-	85 AF	STA	\$AF
0012-	A9 5F	LDA	#\$5F
0014-	38	SEC	
0015-	E5 AF	SBC	\$AF
0017-	85 AF	STA	\$AF
0019-	A9 AF	LDA	#\$AF
001B-	85 AE	STA	\$AE
001D-	A5 A1	LDA	\$A1
001F-	38	SEC	
0020-	E5 A2	SBC	\$A2
0022-	F0 0F	BEQ	\$0033
0024-	AA	TAX	
0025-	A5 AE	LDA	\$AE
0027-	38	SEC	
0028-	E9 C0	SBC	#\$C0
002A-	85 AE	STA	\$AE
002C-	B0 02	BCS	\$0030
002E-	C6 AF	DEC	\$AF
0030-	CA	DEX	
0031-	D0 F4	BNE	\$0027
0033-	60	RTS	
0034-	00	BRK	
0035-	00	BRK	
0036-	00	BRK	
0037-	00	BRK	
0038-	00	BRK	
0039-	00	BRK	
003A-	00	BRK	
003B-	00	BRK	
003C-	00	BRK	
003D-	00	BRK	
003E-	00	BRK	
003F-	00	BRK	
0040-	A5 A1	LDA	\$A1
0042-	4A	LSR	
0043-	4A	LSR	
0044-	4A	LSR	
0045-	85 AB	STA	\$AB
0047-	0A	ASL	
0048-	0A	ASL	
0049-	0A	ASL	
004A-	85 A2	STA	\$A2
004C-	A9 47	LDA	#\$47
004E-	38	SEC	
004F-	E5 AB	SBC	\$AB
0051-	85 AB	STA	\$AB
0053-	A9 CF	LDA	#\$CF
0055-	85 AA	STA	\$AA
0057-	A5 A1	LDA	\$A1
0059-	38	SEC	
005A-	E5 A2	SBC	\$A2
005C-	F0 0F	BEQ	\$006D
005E-	AA	TAX	
005F-	A5 AA	LDA	\$AA
0061-	38	SEC	
0062-	E9 20	SBC	#\$20
0064-	85 AA	STA	\$AA
0066-	B0 02	BCS	\$006A
0068-	C6 AB	DEC	\$AB
006A-	CA	DEX	
006B-	D0 F4	BNE	\$0061
006D-	60	RTS	
006E-	00	BRK	
006F-	00	BRK	
0070-	00	BRK	
0071-	00	BRK	
0072-	00	BRK	
0073-	00	BRK	
0074-	00	BRK	
0075-	00	BRK	
0076-	00	BRK	
0077-	00	BRK	
0078-	A5 AC	LDA	\$AC
007A-	38	SEC	
007B-	E9 08	SBC	#\$08
007D-	85 AC	STA	\$AC
007F-	B0 02	BCS	\$0083

ADR-ACOR

Calculate address for
accumulating auto correl
given ht#

DEC-BIN (convert 1 Ant to
binary)

re-adjust address to binary

0D81-	C6 AD	DEC	\$AD		
0D83-	A9 7F	LDA	##7F		
0D85-	85 A2	STA	\$A2		
0D87-	A9 BF	LDA	##BF		
0D89-	85 A1	STA	\$A1	} calculate data address	
0D8B-	A5 A0	LDA	\$A0		
0D8D-	F0 0F	BEQ	\$0D9E		
0D8F-	AA	TAX			
0D90-	A5 A1	LDA	\$A1		
0D92-	38	SEC			
0D93-	E9 40	SBC	##40		
0D95-	85 A1	STA	\$A1		
0D97-	B0 02	BCS	\$0D9B		
0D99-	C6 A2	DEC	\$A2		
0D9B-	CA	DEX			
0D9C-	D0 F4	BNE	\$0D92	} calculate binary address(ht#0) for given ant#	
0D9E-	A9 03	LDA	##03		
0DA0-	38	SEC			
0DA1-	E5 A0	SBC	\$A0		
0DA3-	F0 0F	BEQ	\$0DB4		
0DA5-	A8	TRX			
0DA6-	A5 AC	LDA	\$AC		
0DA8-	18	CLC			
0DA9-	69 10	ADC	##10		
0DAB-	85 AC	STA	\$AC		
0DAD-	90 02	BCC	\$0DB1		
0DAF-	E6 AD	INC	\$AD		
0DB1-	88	DEY			
0DB2-	D0 F4	BNE	\$0DAB	} ht#	
0DB4-	A9 20	LDA	##20		
0DB6-	85 A3	STA	\$A3		
0DB8-	A0 40	LDY	##40	} add 64 amplitudes	
0DBA-	A2 00	LDX	##00		
0DBC-	A9 00	LDA	##00		
0DBE-	18	CLC			
0DBF-	71 A1	ADC	(<A1>,Y		
0DC1-	90 01	BCC	\$0DC4		
0DC3-	E8	INX			
0DC4-	88	DEY			
0DC5-	D0 F7	BNE	\$0DBE		
0DC7-	4A	LSR			} ÷ by 64 = sig
0DC8-	4A	LSR			
0DC9-	4A	LSR			
0DCA-	4A	LSR			
0DCB-	4A	LSR			
0DCC-	4A	LSR			
0DCD-	85 A7	STA	\$A7		
0DCF-	8A	TXA			
0DD0-	0A	ASL			
0DD1-	0A	ASL			
0DD2-	38	SEC		← intended	
0DD3-	65 A7	ADC	\$A7	} calculate index for $\frac{1}{2}$ sig store	
0DD5-	85 A7	STA	\$A7		
0DD7-	A5 A3	LDA	\$A3		
0DD9-	EA	NOP			
0DDA-	EA	NOP			
0DDB-	EA	NOP			
0DDC-	0A	ASL			
0DDD-	AA	TAX			
0DDE-	A5 A7	LDA	\$A7		
0DE0-	18	CLC			
0DE1-	7D 9F 1F	ADC	\$1F9F,X	} add to $\frac{1}{2}$ sig store	
0DE4-	9D 9F 1F	STA	\$1F9F,X		
0DE7-	90 04	BCC	\$0DED		
0DE9-	CA	DEX			
0DEA-	FE 9F 1F	INC	\$1F9F,X		
0DED-	A0 40	LDY	##40		} convert 64 bytes to binary
0DEF-	A9 08	LDA	##08		
0DF1-	85 A6	STA	\$A6		
0DF3-	A9 00	LDA	##00		
0DF5-	85 A4	STA	\$A4		
0DF7-	A2 08	LDX	##08		
0DF9-	B1 A1	LDA	(<A1>,Y		
0DFB-	C5 A7	CMP	\$A7		
0DFD-	66 A4	ROR	\$A4		
0DFF-	88	DEY		} convert 8 data bytes to one "binary byte"	

0E00-	CA	DEX		
0E01-	D0 F6	BNE	\$0DF9	
0E03-	84 A5	STY	\$A5	
0E05-	A4 A6	LDY	\$A6	
0E07-	A5 A4	LDA	\$A4	
0E09-	91 AC	STA	(\$AC),Y	
0E0B-	A4 A5	LDY	\$A5	
0E0D-	C6 A6	DEC	\$A6	
0E0F-	D0 E2	BNE	\$0DF3	
0E11-	A5 AC	LDA	\$AC	
0E13-	38	SEC		
0E14-	E9 40	SBC	#\$40	
0E16-	85 AC	STA	\$AC	
0E18-	B0 02	BCS	\$0E1C	
0E1A-	C6 AD	DEC	\$AD	
0E1C-	C6 A2	DEC	\$A2	
0E1E-	C6 A3	DEC	\$A3	
0E20-	D0 96	BNE	\$0DB8	
0E22-	A9 20	LDA	#\$20	
0E24-	85 9E	STA	\$9E	
0E26-	60	RTS		
<hr/>				
0E27-	20 FE 0C	JSR	\$0CFE	<u>XCOR</u>
0E2A-	A9 A1	LDA	#\$A1	
0E2C-	85 B0	STA	\$B0	
0E2E-	A9 00	LDA	#\$00	
0E30-	85 B1	STA	\$B1	
0E32-	A9 08	LDA	#\$08	
0E34-	85 9F	STA	\$9F	
0E36-	A5 AD	LDA	\$AD	
0E38-	85 B3	STA	\$B3	
0E3A-	A5 AC	LDA	\$AC	
0E3C-	38	SEC		
0E3D-	E9 02	SBC	#\$02	
0E3F-	EA	NOP		
0E40-	EA	NOP		
0E41-	85 B2	STA	\$B2	
0E43-	B0 02	BCS	\$0E47	
0E45-	C6 B3	DEC	\$B3	
0E47-	A0 0A	LDY	#\$0A	
0E49-	B1 B2	LDA	(\$B2),Y	
0E4B-	91 B0	STA	(\$B0),Y	
0E4D-	88	DEY		
0E4E-	D0 F9	BNE	\$0E49	
0E50-	A9 00	LDA	#\$00	
0E52-	85 A0	STA	\$A0	
0E54-	85 A1	STA	\$A1	
0E56-	A9 40	LDA	#\$40	
0E58-	85 B7	STA	\$B7	
0E5A-	A9 03	LDA	#\$03	
0E5C-	85 B6	STA	\$B6	
0E5E-	A5 AE	LDA	\$AE	
0E60-	85 B2	STA	\$B2	
0E62-	A5 AF	LDA	\$AF	
0E64-	85 B3	STA	\$B3	
0E66-	A5 AD	LDA	\$AD	
0E68-	85 B5	STA	\$B5	
0E6A-	A5 AC	LDA	\$AC	
0E6C-	18	CLC		
0E6D-	69 10	ADC	#\$10	
0E6F-	85 B4	STA	\$B4	
0E71-	90 02	BCC	\$0E75	
0E73-	E6 B5	INC	\$B5	
0E75-	A4 9F	LDY	\$9F4	
0E77-	A2 00	LDX	#\$00	
0E79-	B1 B0	LDA	(\$B0),Y	
0E7B-	F0 22	BEQ	\$0E9F	
0E7D-	31 B4	AND	(\$B4),Y	
0E7F-	4A	LSR		
0E80-	90 01	BCC	\$0E83	
0E82-	E8	INX		
0E83-	4A	LSR		
0E84-	90 01	BCC	\$0E87	
0E86-	E3	INX		
0E87-	4A	LSR		
0E88-	90 01	BCC	\$0E8B	

adjust addresses for next ht#

used in OPER 2 for selecting ht# for XCOR & ACOR

load Ant# 4 binary (10bytes) into zero page for "comparing" and rolling

zero 2 bytes below this (not required)

lag x 2 #

Xcorrel #

= eight for -ve lags
= ten for +ve lags

add the #1's in the result of the "AND", LINEAR program to save time.

0E8A-	E8	INX	
0E8B-	4A	LSR	
0E8C-	90 01	BCC	\$0E8F
0E8E-	E8	INX	
0E8F-	4A	LSR	
0E90-	90 01	BCC	\$0E93
0E92-	E8	INX	
0E93-	4A	LSR	
0E94-	90 01	BCC	\$0E97
0E96-	E8	INX	
0E97-	4A	LSR	
0E98-	90 01	BCC	\$0E9B
0E9A-	E8	INX	
0E9B-	4A	LSR	
0E9C-	90 01	BCC	\$0E9F
0E9E-	E8	INX	
0E9F-	88	DEY	
0EA0-	D0 D7	BNE	\$0E79
0EA2-	A4 B7	LDY	\$B7
0EA4-	8A	TXA	
0EA5-	18	CLC	
0EA6-	71 B2	ADC	(\$B2),Y
0EA8-	91 B2	STA	(\$B2),Y
0EAA-	90 08	BCC	\$0EB4
0EAC-	88	DEY	
0EAD-	18	CLC	
0EAE-	B1 B2	LDA	(\$B2),Y
0EB0-	69 01	ADC	#\$01
0EB2-	91 B2	STA	(\$B2),Y
0EB4-	A5 B4	LDA	\$B4
0EB6-	18	CLC	
0EB7-	69 10	ADC	#\$10
0EB9-	85 B4	STA	\$B4
0EBB-	90 02	BCC	\$0EBF
0EBD-	E6 B5	INC	\$B5
0EBF-	A5 B2	LDA	\$B2
0EC1-	38	SEC	
0EC2-	E9 40	SBC	#\$40
0EC4-	85 B2	STA	\$B2
0EC6-	B0 02	BCS	\$0ECA
0EC8-	C6 B3	DEC	\$B3
0ECA-	C6 B6	DEC	\$B6
0ECC-	D0 A7	BNE	\$0E75
0ECE-	C6 B7	DEC	\$B7
0ED0-	C6 B7	DEC	\$B7
0ED2-	F0 38	BEQ	\$0F0F
0ED4-	18	CLC	
0ED5-	26 AB	ROL	\$AB
0ED7-	26 AA	ROL	\$AA
0ED9-	26 A9	ROL	\$A9
0EDB-	26 A8	ROL	\$A8
0EDD-	26 A7	ROL	\$A7
0EDF-	26 A6	ROL	\$A6
0EE1-	26 A5	ROL	\$A5
0EE3-	26 A4	ROL	\$A4
0EE5-	26 A3	ROL	\$A3
0EE7-	26 A2	ROL	\$A2
0EE9-	26 A1	ROL	\$A1
0EEB-	26 A0	ROL	\$A0
0EED-	A5 B7	LDA	\$B7
0EEF-	C9 20	CMP	#\$20
0EF1-	D0 19	BNE	\$0F0C
0EF3-	A9 0A	LDA	#\$0A
0EF5-	85 9F	STA	\$9F
0EF7-	A9 9F	LDA	#\$9F
0EF9-	85 B0	STA	\$B0
0EFB-	A9 00	LDA	#\$00
0EFD-	85 A0	STA	\$A0
0EFF-	85 A1	STA	\$A1
0F01-	A5 AC	LDA	\$AC
0F03-	38	SEC	
0F04-	E9 02	SBC	#\$02
0F06-	85 AC	STA	\$AC
0F08-	B0 02	BCS	\$0F0C

add #1's to Xcorrel store 1 at proper lag

shift ANT#4 zero page sequence to the left by 1 bit

adjust addresses etc for +ve lags

0F0A-	C6 AD	DEC	\$AD	
0F0C-	4C 5A 0E	JMP	\$0E5A	
0F0F-	A5 AC	LDA	\$AC	
0F11-	18	CLC		
0F12-	69 02	ADC	#\$02	Calculate addressing for following operation
0F14-	85 AC	STA	\$AC	
0F16-	90 02	BCC	\$0F1A	
0F18-	E6 AD	INC	\$AD	
0F1A-	A5 AD	LDA	\$AD	
0F1C-	85 B5	STA	\$B5	
0F1E-	A5 AC	LDA	\$AC	
0F20-	38	SEC		
0F21-	E9 08	SBC	#\$08	
0F23-	85 B4	STA	\$B4	
0F25-	B0 02	BCS	\$0F29	
0F27-	C6 B5	DEC	\$B5	
0F29-	A0 08	LDY	#\$08	
0F2B-	B1 B4	LDA	(\$B4),Y	Transfer ANT#4 store 1 binary to store 2 & zero store 1
0F2D-	91 AC	STA	(\$AC),Y	
0F2F-	A9 00	LDA	#\$00	
0F31-	91 B4	STA	(\$B4),Y	
0F33-	88	DEY		
0F34-	D0 F5	BNE	\$0F2B	
0F36-	60	RTS		
0F37-	00	BRK		
0F38-	20 40 0D	JSR	\$0D40	<u>ACOR</u> (calculate Auto correl for ANT# 1,2,3)
0F3B-	A5 AD	LDA	\$AD	Calculate addresses to binary store
0F3D-	85 AF	STA	\$AF	
0F3F-	A5 AC	LDA	\$AC	
0F41-	18	CLC		
0F42-	69 0E	ADC	#\$0E	
0F44-	85 AE	STA	\$AE	
0F46-	90 02	BCC	\$0F4A	
0F48-	E6 AF	INC	\$AF	
0F4A-	A5 AF	LDA	\$AF	
0F4C-	85 B3	STA	\$B3	
0F4E-	A5 AE	LDA	\$AE	
0F50-	18	CLC		
0F51-	69 02	ADC	#\$02	auto correl # (ANT#)
0F53-	85 B2	STA	\$B2	
0F55-	90 02	BCC	\$0F59	
0F57-	E6 B3	INC	\$B3	
0F59-	A9 03	LDA	#\$03	
0F5B-	85 B4	STA	\$B4	
0F5D-	A9 A1	LDA	#\$A1	transfer sequence to zero page
0F5F-	85 B0	STA	\$B0	
0F61-	A9 00	LDA	#\$00	
0F63-	85 B1	STA	\$B1	
0F65-	A0 08	LDY	#\$08	
0F67-	B1 B2	LDA	(\$B2),Y	modify zero page address and zero two bytes below
0F69-	91 B0	STA	(\$B0),Y	
0F6B-	88	DEY		
0F6C-	D0 F9	BNE	\$0F67	
0F6E-	A9 9F	LDA	#\$9F	
0F70-	85 B0	STA	\$B0	
0F72-	A9 00	LDA	#\$00	
0F74-	85 A0	STA	\$A0	
0F76-	85 A1	STA	\$A1	
0F78-	A9 20	LDA	#\$20	"AND" ten bytes
0F7A-	85 B5	STA	\$B5	
0F7C-	A9 00	LDA	#\$00	
0F7E-	85 B6	STA	\$B6	
0F80-	A0 0A	LDY	#\$0A	
0F82-	B1 B0	LDA	(\$B0),Y	add #1's in result of "AND" operation
0F84-	F0 0C	BEQ	\$0F92	
0F86-	31 AE	AND	(\$AE),Y	
0F88-	A2 08	LDR	#\$08	
0F8A-	4A	LSR		
0F8B-	90 02	BCC	\$0F8F	
0F8D-	E6 B6	INC	\$B6	
0F8F-	CA	DEX		
0F90-	D0 F8	BNE	\$0F8A	
0F92-	88	DEY		
0F93-	D0 ED	BNE	\$0F82	
0F95-	A4 B5	LDY	\$B5	
0F97-	B1 AA	LDA	(\$AA),Y	
0F99-	18	CLC		

0F9A-	65 B6	ADC	\$B6	
0F9C-	91 AA	STA	(&AA),Y	
0F9E-	90 08	BCC	\$0F98	
0FA0-	88	DEY		
0FA1-	B1 AA	LDA	(&AA),Y	
0FA3-	18	CLC		
0FA4-	69 01	ADC	##01	
0FA6-	91 AA	STA	(&AA),Y	
0FA8-	18	CLC		
0FA9-	26 A9	ROL	\$A9	} shift left by 1 bit
0FAB-	26 A8	ROL	\$A8	
0FAD-	26 A7	ROL	\$A7	
0FAF-	26 A6	ROL	\$A6	
0FB1-	26 A5	ROL	\$A5	
0FB3-	26 A4	ROL	\$A4	
0FB5-	EA	NOP		
0FB6-	EA	NOP		
0FB7-	26 A3	ROL	\$A3	
0FB9-	26 A2	ROL	\$A2	
0FBB-	26 A1	ROL	\$A1	} reset addresses for next ant#
0FBD-	26 A0	ROL	\$A0	
0FBF-	C6 B5	DEC	\$B5	
0FC1-	C6 B5	DEC	\$B5	
0FC3-	D0 B7	BNE	\$0F7C	
0FC5-	A5 AE	LDA	\$AE	
0FC7-	18	CLC		
0FC8-	69 10	ADC	##10	
0FCA-	85 AE	STA	\$AE	
0FCC-	90 02	BCC	\$0FD0	
0FCE-	E6 AF	INC	\$AF	} modify addresses for following operation
0FD0-	A5 B2	LDA	\$B2	
0FD2-	18	CLC		
0FD3-	69 10	ADC	##10	
0FD5-	85 B2	STA	\$B2	
0FD7-	90 02	BCC	\$0FDB	
0FD9-	E6 B3	INC	\$B3	
0FDB-	C6 B4	DEC	\$B4	
0FDD-	F0 03	BEQ	\$0FE2	
0FDF-	4C 5D 0F	JMP	\$0F5D	
0FE2-	A5 B2	LDA	\$B2	} move store 2 binary to store 1 binary and zero store 2 binary for ANT#1,2,3
0FE4-	38	SEC		
0FE5-	E9 10	SBC	##10	
0FE7-	85 B2	STA	\$B2	
0FE9-	B0 02	BCS	\$0FED	
0FEB-	C6 B3	DEC	\$B3	
0FED-	A5 B3	LDA	\$B3	
0FEF-	85 AF	STA	\$AF	
0FF1-	A5 B2	LDA	\$B2	
0FF3-	38	SEC		
0FF4-	E9 08	SBC	##08	
0FF6-	85 AE	STA	\$AE	
0FF8-	B0 02	BCS	\$0FFC	
0FFA-	C6 AF	DEC	\$AF	
0FFC-	A9 03	LDA	##03	
0FFE-	85 B4	STA	\$B4	
1000-	A0 08	LDY	##08	
1002-	B1 AE	LDA	(&AE),Y	
1004-	91 B2	STA	(&B2),Y	
1006-	A9 00	LDA	##00	
1008-	91 AE	STA	(&AE),Y	
100A-	88	DEY		
100B-	D0 F5	BNE	\$1002	
100D-	38	SEC		
100E-	A5 B2	LDA	\$B2	
1010-	E9 10	SBC	##10	
1012-	85 B2	STA	\$B2	
1014-	B0 02	BCS	\$1018	
1016-	C6 B3	DEC	\$B3	
1018-	38	SEC		
1019-	A5 AE	LDA	\$AE	
101B-	E9 10	SBC	##10	
101D-	85 AE	STA	\$AE	
101F-	B0 02	BCS	\$1023	
1021-	C6 AF	DEC	\$AF	
1023-	C6 B4	DEC	\$B4	
1025-	D0 D9	BNE	\$1000	
1027-	60	RTS		
1028-	00	BRK		
1029-	00	BRK		
102A-	00	BRK		
102B-	00	BRK		

102C-	A9 20	LDA	#\$20	} ht#	<u>RESET-GAINS</u>	
102E-	85 A3	STA	#\$A3			
1030-	A9 00	LDA	#\$00			
1032-	85 A1	STA	#\$A1			
1034-	85 A0	STA	#\$A0			
1036-	A5 BE	LDA	#\$BE			
1038-	0A	ASL				
1039-	0A	ASL				
103A-	85 A2	STA	#\$A2			← # Blocks X4
103C-	A2 08	LDX	#\$08			
103E-	18	CLC				
103F-	66 A2	ROR	#\$A2			
1041-	66 A1	ROR	#\$A1			
1043-	A5 A3	LDA	#\$A3			
1045-	0A	ASL				
1046-	A8	TAY				
1047-	38	SEC				
1048-	B9 9F IF	LDA	#\$1F9F,Y			
104B-	E5 A1	SBC	#\$A1			
104D-	85 A4	STA	#\$A4			
104F-	88	DEY				
1050-	B9 9F IF	LDA	#\$1F9F,Y			
1053-	E5 A2	SBC	#\$A2			
1055-	90 09	BCC	#\$1060			
1057-	99 9F IF	STA	#\$1F9F,Y			
105A-	C8	INY				
105B-	A5 A4	LDA	#\$A4			
105D-	99 9F IF	STA	#\$1F9F,Y			
1060-	26 A0	ROL	#\$A0			
1062-	CA	DEX				
1063-	D0 D9	BNE	#\$103E			
1065-	A6 A3	LDX	#\$A3			
1067-	BD 7F IF	LDA	#\$1F7F,X	} save old gains for O/P to tape		
106A-	9D 36 IF	STA	#\$1F36,X			
106D-	A5 A0	LDA	#\$A0			
106F-	C9 50	CMP	#\$50			
1071-	B0 0D	BCS	#\$1080			
1073-	BD 7F IF	LDA	#\$1F7F,X	} Reset-gains if $80 \leq \text{sig} \leq 160$ leave gain as is, otherwise try to move it up or down by one step		
1076-	C9 07	CMP	#\$07			
1078-	F0 12	BEQ	#\$108C			
107A-	FE 7F IF	INC	#\$1F7F,X			
107D-	18	CLC				
107E-	90 0C	BCC	#\$108C			
1080-	C9 A0	CMP	#\$A0			
1082-	90 08	BCC	#\$108C			
1084-	BD 7F IF	LDA	#\$1F7F,X			
1087-	F0 03	BEQ	#\$108C			
1089-	DE 7F IF	DEC	#\$1F7F,X			
108C-	A6 A3	LDX	#\$A3			
108E-	A5 A0	LDA	#\$A0			
1090-	9D 56 IF	STA	#\$1F56,X	} save sig for O/P to tape		
1093-	C6 A3	DEC	#\$A3			
1095-	D0 99	BNE	#\$1030			
1097-	60	RTS				
1098-	00	BRK				
1099-	A5 C4	LDA	#\$C4			
109B-	C9 02	CMP	#\$02			
109D-	D0 01	BNE	#\$10A0			
109F-	60	RTS				
10A0-	A5 C2	LDA	#\$C2			
10A2-	D0 01	BNE	#\$10A5			
10A4-	60	RTS				
10A5-	A5 C3	LDA	#\$C3			
10A7-	F0 14	BEQ	#\$10BD			
10A9-	A9 11	LDA	#\$11			
10AB-	8D 04 C5	STA	#\$C504			
10AE-	A9 00	LDA	#\$00			
10B0-	8D 04 C5	STA	#\$C504			
10B3-	AD 00 C5	LDA	#\$C500			
10B6-	C9 F0	CMP	#\$F0			
10B8-	F0 F9	BEQ	#\$10B3			
10BA-	C6 C3	DEC	#\$C3			
10BC-	60	RTS				
10BD-	A5 C2	LDA	#\$C2			
10BF-	C9 1F	CMP	#\$1F			

CHECK-PET

(see flow chart)
controls data O/P to and I/P
from PET, and tape recorder
(data, block gaps, file gaps)

10C1-	D0 08	BNE	\$10CB
10C3-	A5 C6	LDA	#C6
10C5-	F0 04	BEQ	\$10CB
10C7-	20 3A 11	JSR	\$113A
10CA-	60	RTS	
10CB-	AD 01 C3	LDA	#C301
10CE-	29 80	AND	##80
10D0-	D0 01	BNE	\$10D3
10D2-	60	RTS	
10D3-	A9 00	LDA	##00
10D5-	8D 01 C3	STA	#C301
10D8-	8D 00 C3	STA	#C300
10DB-	A9 3C	LDA	##3C
10DD-	8D 01 C3	STA	#C301
10E0-	20 A0 11	JSR	\$11A0
10E3-	A5 C2	LDA	#C2
10E5-	C9 20	CMP	##20
10E7-	D0 27	BNE	\$1110
10E9-	A5 C5	LDA	#C5
10EB-	F0 05	BEQ	\$10F2
10ED-	C6 C5	DEC	#C5
10EF-	18	CLC	
10F0-	90 21	BCC	\$1113
10F2-	A5 C4	LDA	#C4
10F4-	D0 1A	BNE	\$1110
10F6-	A5 C7	LDA	#C7
10F8-	F0 14	BEQ	\$110E
10FA-	AD EB 43	LDA	\$43EB
10FD-	D0 0F	BNE	\$110E
10FF-	C6 C7	DEC	#C7
1101-	20 85 11	JSR	\$1185
1104-	20 1A 11	JSR	\$111A
1107-	EA	NOP	
1108-	EA	NOP	
1109-	EA	NOP	
110A-	EA	NOP	
110B-	18	CLC	
110C-	90 05	BCC	\$1113
110E-	E6 C3	INC	#C3
1110-	20 85 11	JSR	\$1185
1113-	20 CA 11	JSR	\$11CA
1116-	C6 C2	DEC	#C2
1118-	60	RTS	
1119-	00	BRK	
111A-	A9 12	LDA	##12
111C-	8D 04 C5	STA	#C504
111F-	A9 00	LDA	##00
1121-	8D 04 C5	STA	#C504
1124-	AD 00 C5	LDA	#C500
1127-	C9 F0	CMP	##F0
1129-	F0 F9	BEQ	\$1124
112B-	A2 03	LDX	##03
112D-	BD 7C 1F	LDA	\$1F7C,X
1130-	95 00	STA	#00,X
1132-	CA	DEX	
1133-	D0 F8	BNE	\$112D
1135-	60	RTS	
1136-	00	BRK	
1137-	00	BRK	
1138-	00	BRK	
1139-	00	BRK	
113A-	A5 C6	LDA	#C6
113C-	C9 01	CMP	##01
113E-	F0 27	BEQ	\$1167
1140-	A2 10	LDX	##10
1142-	BD DF 43	LDA	\$43DF,X
1145-	8D 06 C5	STA	#C506
1148-	AD 00 C5	LDA	#C500
114B-	C9 E0	CMP	##E0
114D-	F0 F9	BEQ	\$1148
114F-	CA	DEX	
1150-	D0 F0	BNE	\$1142
1152-	A2 20	LDX	##20
1154-	BD 56 1F	LDA	\$1F56,X
1157-	8D 06 C5	STA	#C506
115A-	AD 00 C5	LDA	#C500
115D-	C9 E0	CMP	##E0
115F-	F0 F9	BEQ	\$115A

FILE-GAP

"update" initial day #

GAINS-TO-TAPE

put date/time store 2 on tape

Batch 1 O/P

put sig on tape

1161-	CA	DEX		
1162-	D0 F0	BNE	\$1154	} "Batch 2" O/P put gain settings on tape
1164-	C6 C6	DEC	\$C6	
1166-	60	RTS		
1167-	A2 20	LDX	##20	
1169-	B0 36 1F	LDA	\$1F36,X	
116C-	8D 06 C5	STA	\$C506	
116F-	AD 00 C5	LDA	\$C500	
1172-	C9 E0	CMP	##E0	
1174-	F0 F9	BEQ	.\$116F	
1176-	CA	DEX		
1177-	D0 F0	BNE	\$1169	
1179-	C6 C6	DEC	\$C6	
117B-	60	RTS		
117C-	00	BRK		
117D-	00	BRK		
117E-	00	BRK		
117F-	00	BRK		
1180-	00	BRK		
1181-	00	BRK		
1182-	00	BRK		
1183-	00	BRK		
1184-	00	BRK		
1185-	A2 10	LDX	##10	<u>TAPE 16</u>
1187-	B0 26 1F	LDA	\$1F26,X	Put 16 bytes (pet O/P) on tape
118A-	8D 06 C5	STA	\$C506	
118D-	AD 00 C5	LDA	\$C500	
1190-	C9 E0	CMP	##E0	
1192-	F0 F9	BEQ	\$118D	
1194-	CA	DEX		
1195-	D0 F0	BNE	\$1187	
1197-	60	RTS		
1198-	00	BRK		
1199-	00	BRK		
119A-	00	BRK		
119B-	00	BRK		
119C-	00	BRK		
119D-	00	BRK		
119E-	00	BRK		
119F-	00	BRK		
11A0-	A2 10	LDX	##10	<u>READ 16</u>
11A2-	AD 01 C3	LDA	\$C301	Accept 16 bytes from Pet
11A5-	29 80	AND	##80	
11A7-	F0 F9	BEQ	\$11A2	
11A9-	AD 00 C3	LDA	\$C300	
11AC-	9D 26 1F	STA	\$1F26,X	
11AF-	A9 3C	LDA	##3C	
11B1-	8D 01 C3	STA	\$C301	
11B4-	A9 34	LDA	##34	
11B6-	8D 01 C3	STA	\$C301	
11B9-	CA	DEX		
11BA-	D0 E6	BNE	\$11A2	
11BC-	AD 00 C3	LDA	\$C300	
11BF-	60	RTS		
11C0-	00	BRK		
11C1-	00	BRK		
11C2-	00	BRK		
11C3-	00	BRK		
11C4-	00	BRK		
11C5-	00	BRK		
11C6-	00	BRK		
11C7-	00	BRK		
11C8-	00	BRK		
11C9-	00	BRK		
11CA-	A9 20	LDA	##20	<u>WRITE 240</u>
11CC-	38	SEC		Transmit 240 bytes to PET
11CD-	E5 C2	SBC	\$C2	
11CF-	85 A1	STA	\$A1 4 — HT#	
11D1-	8D E4 43	STA	\$43E4 ← put ht# in date/time store 2 as data is O/P	
11D4-	A9 00	LDA	##00	
11D6-	8D 01 C3	STA	\$C301	
11D9-	A9 FF	LDA	##FF	
11DB-	8D 00 C3	STA	\$C300	
11DE-	A9 3C	LDA	##3C	

11E0-	8D 01 C3	STA	#C301	
11E3-	A2 10	LDX	##10	
11E5-	BD DF 43	LDA	#43DF,X	date/time store 2 O/P
11E8-	8D 00 C3	STA	#C300	
11EB-	20 3D 12	JSR	#123D	
11EE-	CA	DEX		
11EF-	D0 F4	BNE	\$11E5	
11F1-	20 FE 0C	JSR	#0CFE	
11F4-	A5 AE	LDA	\$AE	
11F6-	38	SEC		
11F7-	E9 90	SBC	##90	
11F9-	85 AE	STA	\$AE	
11FB-	B0 02	BCS	\$11FF	
11FD-	C6 AF	DEC	\$AF	
11FF-	A5 AF	LDA	\$AF	
1201-	38	SEC		Xcorrel O/P
1202-	E9 1C	SBC	##1C	(calculate address from ADR XCOR and adjust for store 2 (Auto & Cross))
1204-	85 AF	STA	\$AF	
1206-	A0 C0	LDY	##C0	
1208-	B1 AE	LDA	(<\$AE>,Y	
120A-	8D 00 C3	STA	#C300	
120D-	20 3D 12	JSR	#123D	
1210-	88	DEY		
1211-	D0 F5	BNE	\$1208	
1213-	20 40 0D	JSR	#0D40	
1216-	A5 AA	LDA	\$AA	
1218-	38	SEC		
1219-	E9 10	SBC	##10	
121B-	85 AA	STA	\$AA	
121D-	B0 02	BCS	\$1221	
121F-	C6 AB	DEC	\$AB	
1221-	A5 AB	LDA	\$AB	Auto correl O/P
1223-	38	SEC		
1224-	E9 1C	SBC	##1C	
1226-	85 AB	STA	\$AB	
1228-	A0 20	LDY	##20	
122A-	B1 AA	LDA	(<\$AA>,Y	
122C-	8D 00 C3	STA	#C300	
122F-	20 3D 12	JSR	#123D	
1232-	88	DEY		
1233-	D0 F5	BNE	\$122A	
1235-	60	RTS		
1236-	00	BRK		
1237-	00	BRK		
1238-	00	BRK		
1239-	00	BRK		
123A-	00	BRK		
123B-	00	BRK		
123C-	00	BRK		
123D-	A9 3C	LDA	##3C	<u>PIA-INIT</u>
123F-	8D 01 C3	STA	#C301	
1242-	A9 34	LDA	##34	
1244-	8D 01 C3	STA	#C301	
1247-	AD 01 C3	LDA	#C301	
124A-	29 80	AND	##80	
124C-	F0 F9	BEQ	\$1247	
124E-	AD 00 C3	LDA	#C300	
1251-	60	RTS		
1252-	00	BRK		
1253-	00	BRK		
1254-	00	BRK		
1255-	00	BRK		
1256-	00	BRK		
1257-	00	BRK		
1258-	00	BRK		
1259-	00	BRK		
125A-	00	BRK		
125B-	00	BRK		
125C-	00	BRK		
125D-	A9 0C	LDA	##0C	<u>DISPLAY</u>
125F-	85 0C	STA	#0C	
1261-	A9 05	LDA	##05	PUT 17 "bytes" ON SCREEN
1263-	85 0D	STA	#0D	address 0A,0B must be set manually when program is loaded
1265-	A0 11	LDY	##11	
1267-	B1 0A	LDA	(<\$0A>,Y	
1269-	29 0F	AND	##0F	

126B-	09 B0	ORA	##B0	
126D-	91 0C	STA	(#0C),Y	
126F-	88	DEY		
1270-	D0 F5	BNE	\$1267	
1272-	60	RTS		
1273-	00	BRK		
1274-	00	BRK		
1275-	20 BA 12	JSR	\$12BA	WAIT (NOT A SUBROUTINE)
1278-	D0 03	BNE	\$127D	
127A-	4C 02 09	JMP	\$0902	For manual operations
127D-	C9 D7	CMP	##D7	while system is running
127F-	F0 03	BEQ	\$1284	
1281-	4C 02 09	JMP	\$0902	
1284-	8D 0C 05	STA	\$050C	store "W" on screen
1287-	20 BA 12	JSR	\$12BA	
128A-	C9 C5	CMP	##C5	"E"
128C-	D0 06	BNE	\$1294	
128E-	20 D7 12	JSR	\$12D7	W= WAIT
1291-	4C 87 12	JMP	\$1287	E=END OF TAPE
1294-	20 BA 12	JSR	\$12BA	S= START
1297-	C9 D3	CMP	##D3	(ie return from "W")
1299-	D0 08	BNE	\$12A3	M= initiate manual record
129B-	A9 A0	LDA	##A0	
129D-	8D 0C 05	STA	\$050C	"6"
12A0-	4C 02 09	JMP	\$0902	
12A3-	C9 CD	CMP	##CD	"M"
12A5-	D0 E0	BNE	\$1287	
12A7-	A9 A0	LDA	##A0	"6"
12A9-	8D 0C 05	STA	\$050C	
12AC-	4C 0F 09	JMP	\$090F	
12AF-	00	BRK		
12B0-	00	BRK		
12B1-	00	BRK		
12B2-	00	BRK		
12B3-	00	BRK		
12B4-	00	BRK		
12B5-	00	BRK		
12B6-	00	BRK		
12B7-	00	BRK		
12B8-	00	BRK		
12B9-	00	BRK		
12BA-	20 20 09	JSR	\$0920	"update"
12BD-	A9 00	LDA	##00	KEY
12BF-	2C 00 C0	BIT	\$C000	if no key pressed then
12C2-	10 08	BPL	\$12CC	accum = 0
12C4-	AD 00 C0	LDA	\$C000	
12C7-	2C 10 C0	BIT	\$C010	
12CA-	09 80	ORA	##80	
12CC-	60	RTS		
12CD-	00	BRK		
12CE-	00	BRK		
12CF-	00	BRK		
12D0-	00	BRK		
12D1-	00	BRK		
12D2-	00	BRK		
12D3-	00	BRK		
12D4-	00	BRK		
12D5-	00	BRK		
12D6-	00	BRK		
12D7-	A2 10	LDX	##10	END OF TAPE
12D9-	A9 00	LDA	##00	
12DB-	8D 06 C5	STA	\$C506	put 16 bytes (0's) and a
12DE-	AD 00 C5	LDA	\$C500	file gap on tape
12E1-	C9 E0	CMP	##E0	
12E3-	F0 F9	BEQ	\$12DE	
12E5-	CA	DEX		
12E6-	D0 F3	BNE	\$12DB	
12E8-	20 1A 11	JSR	\$111A	
12EB-	A9 02	LDA	##02	
12ED-	85 C4	STA	\$C4	
12EF-	A9 01	LDA	##01	
12F1-	85 C5	STA	\$C5	
12F3-	60	RTS		
12F4-	00	BRK		
12F5-	00	BRK		

APPENDIX B

Theory of binary correlation

1. Calculation of correlation coefficient from number of 1-matches
2. Correction to binary correlation (Gaussian amplitude distribution)
3. Effect of overloading on correction function (Gaussian)
4. Test of correction function on Rayleigh distributed data
5. Further comments on the binary method

1. Calculation of correlation from # 1-matches

The correlation coefficient is defined by:

$$\rho = \frac{\overline{xy} - \bar{x}\bar{y}}{\sqrt{(\overline{x^2} - \bar{x}^2)(\overline{y^2} - \bar{y}^2)}}$$

where x and y are elements of the same (auto correlation) or different (cross correlation) fading sequences, and there may be some lag between x and y measurements.

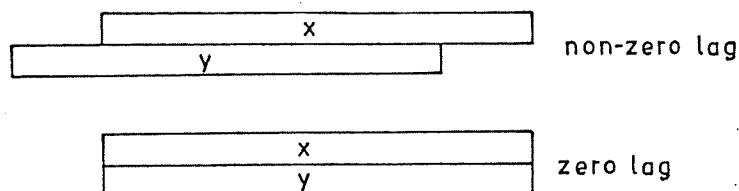


Fig. 1

If the elements x and y are binary (i.e. 0 or 1), then $\sum xy$ is given directly by the results of the 'AND' operation; and so $\overline{xy} = (\# \text{ 1-matches}) / (N - \ell)$, where N is the total number of samples in a sequence and ℓ is the lag. Also $\bar{x} = \sum x / N$, where $\sum x$ is the number of ones in the sequence; and $\overline{x^2} = \sum x^2 / N$, where $\sum x^2$ is again the number of ones (since x can only be 0 or 1). The number of 1's could be counted in each sequence separately, however, since the fading should be statistically the same at each antenna, the sum of the #1-matches from the zero lag auto correlation calculations can be used to estimate the 'average' number of 1's in a sequence. For various reasons this sum has only be calculated over the three outer antennas (#1,2,3). Since the amplitude distribution is usually skewed towards low amplitudes (Rayleigh type), the number of 1's is usually $< \frac{1}{2}N$. Thus, if a_0 is the number of 1-matches ($\sum xy$), and a_1 is the number of 1's in any sequence ($\sum x$), the correlation coefficient is:

$$\rho' = \frac{\frac{a_0}{N - \ell} - \frac{a_1^2}{N^2}}{\frac{a_1}{N} - \frac{a_1^2}{N^2}}$$

This is the correlation between binary sequences, and must be corrected as explained below, to estimate the correlation which would have been obtained with the original sequences.

2. Correction to binary correlation

As will be shown below, the expectation value of correlation between two amplitude sequences which have been converted to binary with respect to their means is lower than that which would have been found between the original sequences. Thus some sort of correction is required. The derivation of this correction function closely follows that of Weinreb(1963). A basic assumption is that the amplitudes have an approximately Gaussian (i.e. Normal) distribution. This is not generally true in practice - usually the distribution is closer to a Rayleigh - however, it makes the problem solvable. The binary correlation, as employed in the present analysis, counts the number of times that both X and Y amplitudes are greater than their means. The probability of this event will be derived next.

Assume that X and Y are taken from a Normal distribution with zero mean and unit s.d. Then the joint probability density function for X and Y is (e.g. Hald, 1952):

$$f(x,y) = \frac{1}{2\pi\sqrt{1-\rho^2}} \exp \left\{ -\frac{1}{2(1-\rho^2)}(x^2 - 2\rho xy + y^2) \right\}$$

where ρ is the correlation coefficient between X and Y. Thus the probability of finding X between x and x+dx and Y between y and y+dy simultaneously is $f(x,y)dx dy$. The probability of finding X and Y both greater than their means is then:

$$P(X>0, Y>0) = \int_{x=0}^{\infty} \int_{y=0}^{\infty} f(x,y) dy dx$$

If polar coordinates are used ($x=r \cos \theta$, $y=r \sin \theta$) the integral is reduced to

$$P(X>0, Y>0) = \frac{1}{2\pi\sqrt{1-\rho^2}} \int_{\theta=0}^{\frac{\pi}{2}} \int_{r=0}^{\infty} \exp\left\{\frac{-r^2(1-\rho \sin 2\theta)}{2(1-\rho^2)}\right\} r \, dr \, d\theta$$

A further change of variable $\xi = r^2$, $d\xi = 2rdr$, and integration of the inner integral leaves:

$$P(X>0, Y>0) = \frac{\sqrt{1-\rho^2}}{2\pi} \int_{\theta=0}^{\frac{\pi}{2}} \frac{1}{(1-\rho \sin 2\theta)} \, d\theta$$

Then let $\alpha = 2\theta$:

$$\begin{aligned} P(X>0, Y>0) &= \frac{\sqrt{1-\rho^2}}{2} \int_{\alpha=0}^{\frac{\pi}{2}} \frac{1}{1-\rho \sin \alpha} \, d\alpha \\ &= \frac{1}{\pi} \tan^{-1} \left\{ \frac{\tan \frac{\alpha}{2} - \rho}{\sqrt{1-\rho^2}} \right\} \Bigg|_{\alpha=0}^{\frac{\pi}{2}} \end{aligned}$$

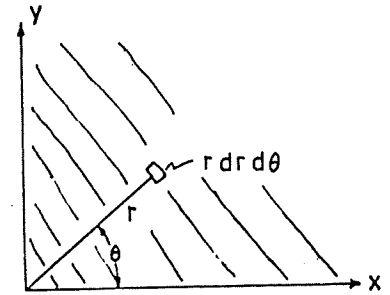


Fig. 1

(This last integral is taken from math tables). The identity

$$\tan^{-1}(a) - \tan^{-1}(b) = \tan^{-1}\left(\frac{a-b}{1-ab}\right)$$

is used to obtain:

$$P(X>0, Y>0) = \frac{1}{\pi} \tan^{-1} \sqrt{\frac{1+\rho}{1-\rho}}$$

Now $P(X>0, Y>0)$ must be related to the correlation of the binary sequences X' and Y' . These are defined by:

$$\begin{aligned} X' &= 1 \quad \text{if } X > 0 & Y' &= 1 \quad \text{if } Y > 0 \\ X' &= 0 \quad \text{if } X \leq 0 & Y' &= 0 \quad \text{if } Y \leq 0 \end{aligned}$$

Figure 2 shows the joint probability function of X', Y' .

Here $d = P(X>0, Y>0)$. Since a Gaussian distribution

has been assumed, the marginal probabilities are

$P(X>0) = P(X \leq 0) = P(Y>0) = P(Y \leq 0) = \frac{1}{2}$. (It is easy to show that $a=d$ and $b=c$).

The cross correlation between X' and Y' is then:

$$\rho' = \frac{\overline{X' Y'} - \overline{X'} \overline{Y'}}{\sqrt{\overline{X'^2} - \overline{X'}^2} \sqrt{\overline{Y'^2} - \overline{Y'}^2}} = \frac{d - (\frac{1}{2})^2}{\sqrt{(\frac{1}{2}) - (\frac{1}{2})^2} \sqrt{(\frac{1}{2}) - (\frac{1}{2})^2}} = 4d - 1$$

		$X' =$		tot.
		0	1	
$Y' = 0$	0	a	c	$\frac{1}{2}$
	1	b	d	$\frac{1}{2}$
tot.		$\frac{1}{2}$	$\frac{1}{2}$	

Fig. 2

This:

$$\rho' = \frac{4}{\pi} \tan^{-1} \sqrt{\frac{1+\rho}{1-\rho}} - 1$$

This can be re-arranged to give

$$\rho = \sin \frac{\pi}{2} \rho'$$

This **then is the** correction function which must be applied to the result of correlating the binary representations of sequences in order to estimate the actual correlation between the original sequences.

3. Effect of overloading on the correlation correction

It is of interest to inquire whether receiver overloading will affect the correction function. This may also indicate whether less major receiver non-linearity will affect the results.

The receiver output for an input signal causing overloading will be assumed to be the maximum output value. It can be seen (Fig. 1) that \bar{A} , the mean amplitude, is less than the actual mean of the Gaussian. Thus the binary sequence will have more ones than zeroes, and thus a mean $> \frac{1}{2}$. Alternatively non-linearity of the type shown in Fig. 2 will skew the distribution towards low amplitudes (assuming no overloading occurs) resulting in a mean $< \frac{1}{2}$. Such skewing occurs naturally in a Rayleigh distribution, and a later section will try to answer the present question in this case.

Here, the procedure is to choose a value, ϵ , and find $P(X > \epsilon, Y > \epsilon)$ for various values of ρ (the actual correlation). ϵ is negative in the case of overloading. Fig. 3 shows contours of $f(x, y)$ for $\rho = +ve, 0$, and $-ve$; and the region (shaded) which must be integrated over

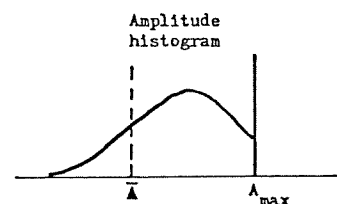


Fig. 1

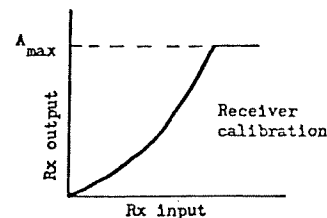


Fig. 2

to find $P(X > \epsilon, Y > \epsilon)$.

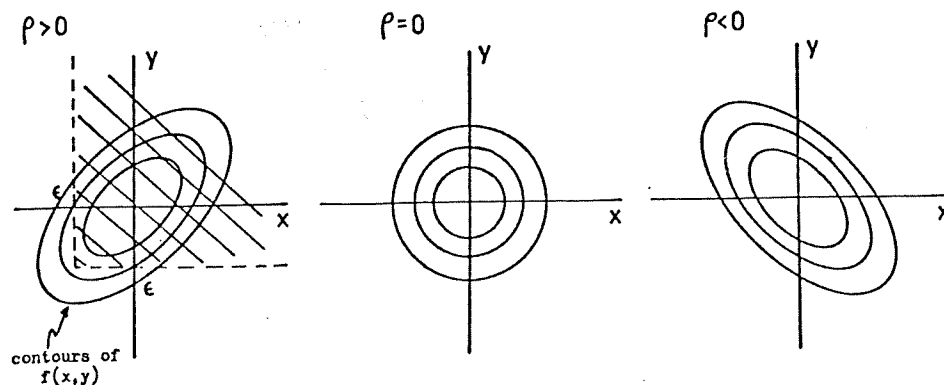


Fig. 3

It is assumed that the Gaussians (for X and Y) have zero mean and unit s.d. as before (there is no loss of generality in this assumption). Thus the mean of the binary sequences, m , is the area of the normal curve (from statistical tables) above ϵ (see Fig. 4), and the binary correlation is:

$$\rho' = \frac{P(X > \epsilon, Y > \epsilon) - m^2}{m(1-m)}$$

where: $P = \int_{x=\epsilon}^{\infty} \int_{y=\epsilon}^{\infty} f(x,y) dy dx$. Unfortunately

when $\epsilon \neq 0$ the integral does not reduce to a simple solution, and has been

performed numerically here (The upper limits used were $x=y=8$, so the P values will be slightly low). Values of ϵ between ± 1 were used; however, instead of quoting ϵ , m is given since it is directly available in the experimental situation. Fig. 5 shows a plot of ρ vs. $\sin \frac{\pi}{2} \rho'$ for +ve ρ and several values of m . The difference between the ideal and overloaded data

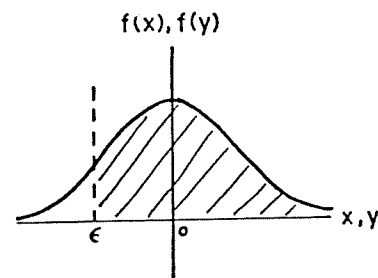


Fig. 4

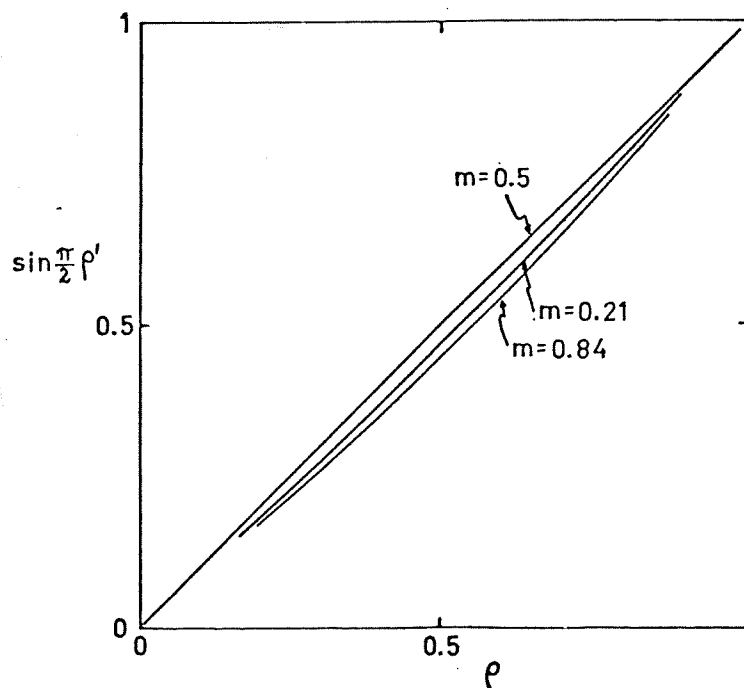


Fig. 5

($m=0.84$) is seen to be almost negligible, and not worth correcting for.

In practice, however, overloading may have serious effects on the correlation if, as is done here, the sequences are assumed to be statistically equivalent (i.e. equal m). This problem will be discussed in Section 5.

4. Test with a Rayleigh distribution

In order to see whether the correction function works when the data are Rayleigh distributed, pairs of sequences X and Y were constructed as follows:

$$X_i = \sqrt{(a_i + K e_i)^2 + (b_i + K f_i)^2}$$

$$Y_i = \sqrt{(c_i + K e_i)^2 + (d_i + K f_i)^2}$$

where a_i , b_i , etc. are independent random values taken from a Gaussian distribution. This definition results in a non-zero expected correlation between X and Y while leaving them Rayleigh distributed. The expectation value of correlation depends on K, which has been chosen to be 2 for this test. 200 pairs of sequences (100 points each) were set up and correlated before and after conversion to binary (with respect to the mean). Histograms of the ratio of the actual to corrected binary correlations, and the actual correlations are shown in Fig. 1. Although there is some spread in the corrected values, there is no bias away from a ratio of 1. Thus the

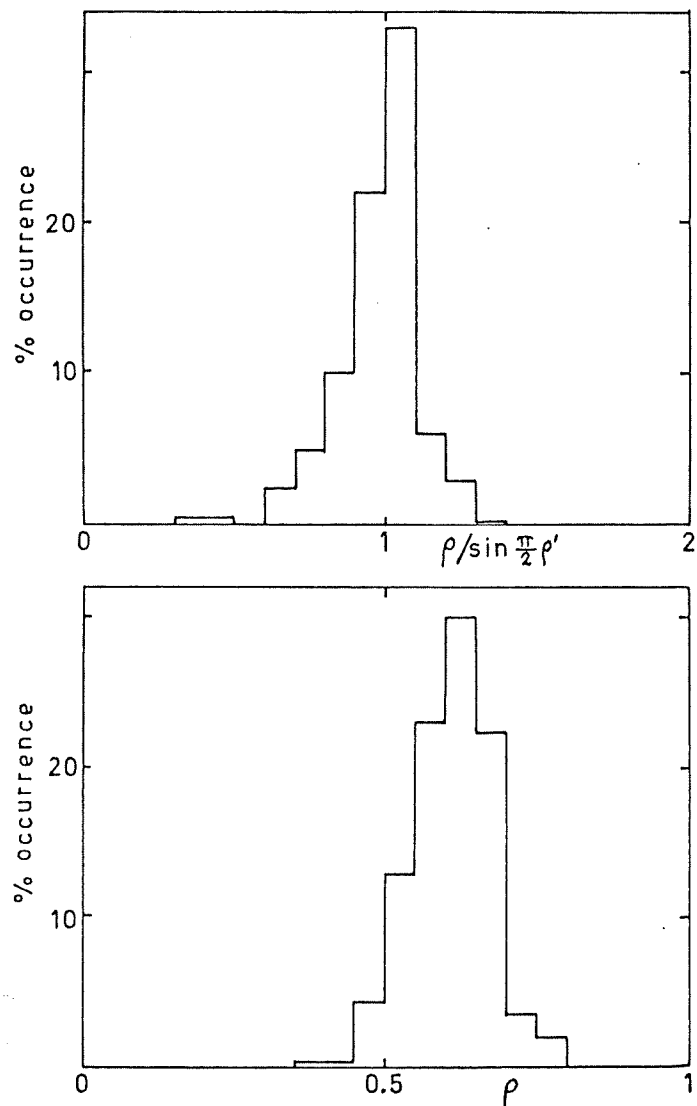


Fig. 1

same correction function should work for Rayleigh distributed data as well.

5. Further comments on binary correlation

a) The binary correlation, ρ' , has been defined as:

$$\rho'_{ij} = \frac{P_{ij} - m^2}{m(1-m)}$$

where m is the 'average' mean of the binary sequences over all antennas (in practice just #1,2,3). This is not a strictly correct definition; it should be

$$\rho'_{ij} = \frac{P_{ij} - m_i m_j}{\sqrt{m_i(1-m_i)m_j(1-m_j)}}$$

where ρ'_{ij} is the correlation between binary sequences from antenna i and j ; however, if ρ' is not very sensitive to changes in m , the first definition should be sufficient since all antennas should see similar amplitude distributions (and therefore similar values of m). A plot of ρ' (and $\rho = \sin \frac{\pi}{2} \rho'$) versus P and m is shown in Fig. 1.

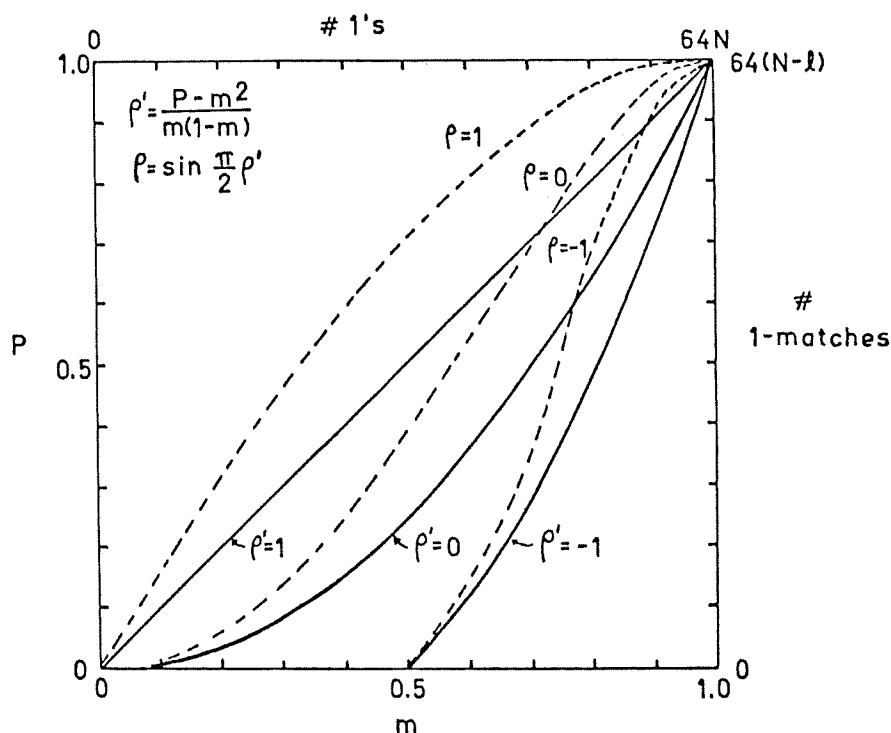


Fig. 1

Note that P cannot be $> m$ since the # 1-matches cannot be greater than the number of 1's in any sequence; also, if $m > 0.5$, P must be $> 2m^2 - m$, because if both sequences are more than half 1's, there must be some 1-matches.

For reasonably Gaussian data (or any data with a symmetrical distribution) $m \cong 0.5$, and ρ' is relatively insensitive to small errors in m . However, if there is serious overloading, m will generally increase and Fig. 1 shows that ρ' becomes more sensitive to errors in m (and to errors in P for that matter, although P is a strictly correct value and will only have errors in a statistical sense). If, in addition, the amplitude distributions at different antennas are not strictly the same then the small error caused by the use of m rather than m_i, m_j may move the whole correlation curve up or down (with only small changes in shape) depending on whether m is less or greater than $\sqrt{m_i m_j}$. This shifting is seen quite often in badly overloaded data (Fig. 2) although a direct connection to differences in m_i, m_j etc. has not been proven as yet.

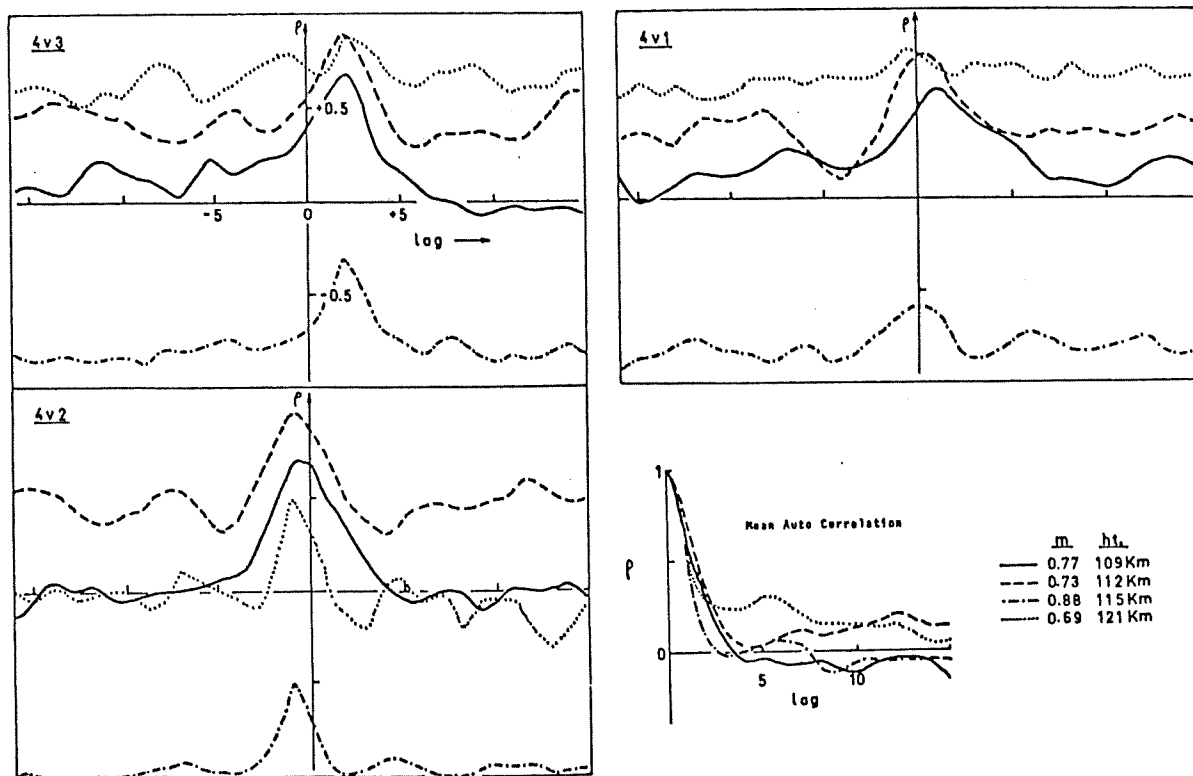


Fig. 2

In a computer simulation of the binary correlation process on real data which used separate sequence means, this problem didn't occur at all. On a separate point, if m_i and m_j are different but m is used to find ρ' , then values > 1 or < -1 are possible; however, the correction through the sin function will reduce the limits to ± 1 .

Other information which can be deduced from Fig. 1 is the expected resolution in ρ' (or ρ); how likely are exactly equal adjacent values of ρ (see the discussion in Appendix C, Section 1, pg.69 on locating peaks in cross correlations). The number of 1-matches is given by $P(N-l)$, and the number of 1's by mN (where N is the number of points in a sequence and l is the lag - since $l \ll N$, it may be ignored). For example if $N=512$ (8 blocks/record) and $m=0.5$, there are 256 possible values of correlation which gives an average resolution of ~ 0.008 in ρ' . If $m=0.8$ (i.e. bad overloading), there are 154 values, giving an average resolution of 0.14.

The conclusion is that overloading (or any other receiver non-linearity which causes amplitude distributions to be skewed towards high values) is quite serious, and should be avoided if at all possible. On the other hand skewing the distribution in the other direction (i.e. low m) is quite tolerable, and perhaps desirable, since ρ' becomes less sensitive to m here; and it has already been shown that the $\sin \frac{\pi}{2} \rho'$ correction works on at least one distribution of this type (i.e. Rayleigh).

If, as has been postulated, the sensitivity of ρ' to m is the cause of the trouble shown in Fig. 2, then a solution which tolerates overloading would be to use individual means m_i instead of the average mean m in the calculation of ρ' . From the point of view of the Apple program, the auto correlation routine (ACOR) would have to be modified to operate on all four antennas, and the zero lag values would have to be accumulated separately for each antenna. This appears to be feasible from an execution time stand-

point and would not require major modifications to the program.

b) It has been noticed by simulating the binary correlation on raw data from the previous wind system, that the correlation correction seems to work well below ~ 80 Km, but results in an average reduction in peak values of $\sim 1.5-2$ below the normal (non-binary calculation) correlation peaks above ~ 80 Km. Presumably this has something to do with a change in amplitude distribution - possibly related to the effect of meteor trails at these heights. Investigation proceeds...

APPENDIX C

Pet program details

1. Description of Pet analysis
2. Correction to time delays for antenna cycling
3. Weighted least squares fit to time delays
4. Poor man's full correlation analysis (PMFCA)
5. Machine language subroutine
6. Input/output memory locations
7. Listing of BASIC program

1. Pet analysis

The input to the Pet computer is the number of 1-matches at each of 32 lags (-16 to +15) for the three cross correlations (c_{ij} ; $i=1,32$, $j=1,3$) and the sum (over three antennas) of the number of 1-matches for 16 lags (0 to 15) for the auto correlation (A_i ; $i=0,15$). The conversion of these values to correct correlation values has been described in Appendix B (pg.58) and the relevant equations are repeated below.

$$\rho = \sin \frac{\pi}{2} \rho'$$

where:

$$\rho' = \frac{\frac{c_{ij}}{N-l} - m^2}{m(1-m)} \quad (\text{cross correlation})$$

or:

$$\rho' = \frac{\frac{a_i}{N-l} - m^2}{m(1-m)} \quad (\text{auto correlation})$$

where $a_i = A_i/3$, $m = a_i/N$, $N = \#$ points in each fading sequence (i.e. $N =$ number of blocks per record times 64), and l is the absolute value of the lag. Here ρ' is the actual correlation value for the binary sequences and ρ is the corrected value. In the Pet analysis usually ρ' is used in place of ρ until an accurate value is required, since time is at a premium.

The program (see listing on pg.79) falls into nine sections, which are described below.

a) Initialization

D is the antenna spacing (assumed equal for all antenna pairs); $P1, P2, P3$ are the directions of the antenna pairs $4v3$ (in the sense 4 to 3 since t_{43} is being used), $4v2$, and $4v1$. TS is the sample spacing at any antenna (= the lag step), CC is the rejection limit on the auto correlation (see (c)); LX is the rejection limit for peaks in the cross correlation, and CY is used in the cycling correction (see section 2, pg.72). Since the Pet is a BASIC machine, the machine language subroutine used to communicate with the Apple must be loaded

from BASIC. It is placed in Tape Buffer #2, and a listing is given in Section 5 (pg. 77).

b) Data input

The locations of the incoming and outgoing data are shown in Section 6 (pg. 78). The data (which apply to a single height) are read from these locations and put in the appropriate matrices.

c) Fast fading test and definition of one point in the mean auto correlation

ρ is calculated for increasing lag (1 to 5) until it is less than CC. If $\rho < CC$ at the first lag, the record is rejected for fast fading (i.e. noise). If $\rho \not< CC$, its value at the last tested lag is saved for future use in the PMFCA. Otherwise linear interpolation is used to find the exact time for ρ to fall to CC. Under the Gaussian assumption this point defines the auto correlation function.

d) Limited lag

This section tries to define a more limited lag (< 16) for the peak search in the cross correlations as described by Meek (1978). Successful definition requires that all cross correlations at zero lag be greater than zero. This is checked by examining just the numerator of ρ' . At present the nominal zero lag values (not corrected for cycling) are used, otherwise interpolation to find the correlation value at true zero lag would be required.

The main purpose of this attempt is to avoid spurious peaks in the cross correlations, not to save time.

e) Peak location

A search is made for the two greatest local maxima in each cross correlation, and their index positions saved. Because of the binary nature of the correlated sequences, the possibility of exactly equal adjacent values at a peak cannot be ignored, especially if the number of blocks/record is small. The present analysis identifies the first such value, from -ve lag, as the peak. Fig. 1 illustrates

two undesirable situations which could occur. The present resolution (8 blocks/record) is about 0.008 (see Appendix B, Section 5, pg. 63 for further comments) and so exactly equal values are not very likely.

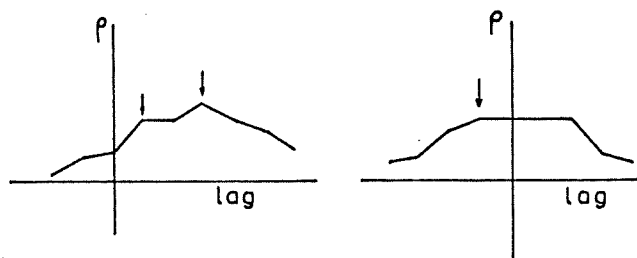


Fig. 1

f) Peak selection

Here accurate values of ρ'_{\max} and t_{\max} for each peak are found by a parabolic fit around the index positions found in e). Then ρ'_{\max} is corrected to get ρ_{\max} and the t_{\max} are corrected for cycling (see section 2, pg. 71). If there is a resulting peak value less than LX it is discarded. If after this process there is a sequence without peaks, the record is rejected. The number of peaks is further reduced by comparing primary and secondary (if any) peaks in each sequence. If $\rho_{\text{pri}}/\rho_{\text{sec}} > 1.8$, then ρ_{sec} is discarded.

At this point there are only one or two peaks left in each sequence. If there is only one significant peak left in each cross correlation sequence, the NTD must be less than 0.3 for the calculation to continue; if two significant peaks in one cross correlation and one in each of the others, then one of the two combinations of peaks must have an NTD < 0.2 or the data are rejected; if two significant peaks in two or more cross correlations, then the calculation continues only if the NTD of the greatest peaks is less than 0.1. If the data have not been rejected, the result is three time delays and three maximum correlation values.

g) Least squares fit to times

Since seven measured values are available to determine six FCA parameters, there is one degree of freedom. The derivation of the PMFCA (Section 4, pg. 73) has assumed that the NTD=0 (as it must be if the assumptions for FCA are

satisfied). To make $NTD=0$ a weighted least squares fit (see Section 3, pg.72) of an apparent velocity vector to the time delays is made, which is then translated back into time delays. Since the peaks were chosen on the basis of low NTD, the resulting time delays will be shifted only slightly from the originals. The original ρ_{max} values are retained.

h) PMFCA

The PMFCA method is based on the Gaussian correlation assumption (e.g. Fedor,1967). Under this assumption it can be shown that the widths of the auto and all cross correlation peaks (e.g. at half the peak value) are equal (Meek,1978). Thus the width of the auto correlation theoretically defines all the widths - and is used for this purpose since it is better defined under experimental conditions than the widths of the cross correlation peaks. The remaining unknowns, for a complete definition of the correlation function (and thus the pattern parameters, velocity etc.), are the magnitudes of the three cross correlation peaks, and two of the time delays (the third being defined by $NTD=0$). The solution for the FCA parameters in terms of these measured values is derived in Section 4 (pg.73).

i) Output

Here the output variables are put into the 16 specified memory locations which the machine language routine sends back to the Apple. If there is no solution, the bytes reserved for FCA parameters are zeroed.

2. Cycling correction

Since all antennas are not sampled simultaneously, "0 lag" is only a nominal term in the Apple program, and a correction to the time lag (e.g. for maximum cross correlation) is required. Fig. 1 (below) shows the transfer of a particular correlation value to the Pet. The antenna cycling is in the order #1,2,3,4 at a rate of 7.5 Hz (.133 sec between antennas). Now suppose there is

← time

a peak in the 4v3 cross correlation exactly at zero lag (nominal). Since antenna #3 is sampled 0.133 sec before #4, this means that the 'event' must have occurred in #3 and moved to #4 in 0.133 sec (although nominally they occur simultaneously)- i.e. the pattern moves from #3 to #4. Thus if

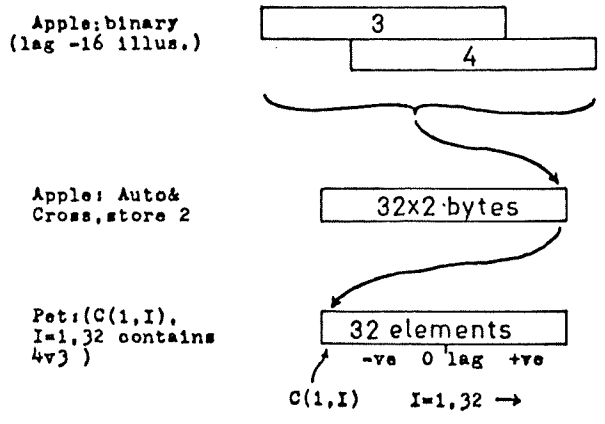


Fig. 1

t_{34}^i (the time taken for the pattern to move from #3 to #4) is the nominal value, the corrected value is:

$$t_{34} = t_{34}^i + 0.133 \text{ sec} ,$$

similarly: $t_{24} = t_{24}^i + 2 \times 0.133 \text{ sec} ,$

and: $t_{14} = t_{14}^i + 3 \times 0.133 \text{ sec} .$

But, by the definition of negative lag in the sketch, we are actually measuring t_{43}^i etc. (not t_{34}^i) - since a peak at negative lag indicates that the 'event' happened in #3 before #4. Therefore, since $t_{43}^i = - t_{34}^i$, the required correction is:

$$t_{43} = t_{43}^i - 0.133 \text{ sec} \text{ etc.}$$

3. Weighted least squares fit to times

If the antenna pairs are defined by (d_i, ψ_i) , where the time delay t_i (given by the position of the maximum in the cross correlation) is taken to be positive if it indicates pattern motion in the direction ψ_i ; then the delays for a particular apparent velocity (V_a, ϕ_a) are given by:

$$t_i = \frac{d_i}{V_a} \cos(\psi_i - \phi_a)$$

$$= \frac{\cos \phi_a}{V_a} (d_i \cos \psi_i) + \frac{\sin \phi_a}{V_a} (d_i \sin \psi_i) .$$

Let

$$\beta = \begin{bmatrix} \frac{\cos \phi_a}{V_a} \\ \frac{\sin \phi_a}{V_a} \end{bmatrix}, \quad Y = \begin{bmatrix} t_1 \\ \cdot \\ \cdot \\ \cdot \\ t_n \end{bmatrix}, \quad X = \begin{bmatrix} d_1 \cos \psi_1 & d_1 \sin \psi_1 \\ \cdot & \cdot \\ \cdot & \cdot \\ \cdot & \cdot \\ d_n \cos \psi_n & d_n \sin \psi_n \end{bmatrix}.$$

and let ρ_i be the peak values of cross correlation (at delay t_i). (Usually there will be only $n=3$ pairs of antennas involved). Then the solution for a least squares fit of apparent velocity to time delays is (e.g. Hoel et al., 1971 Ch. 4):

$$\beta = (X^T X)^{-1} X^T Y$$

This solution minimizes the squared error in the time delays, which is

$$\sum e^2 = (X\beta - Y)^T (X\beta - Y).$$

If X and Y are modified according to:

$$x_{ij} = x_{ij} \rho_i$$

$$y_i = y_i \rho_i$$

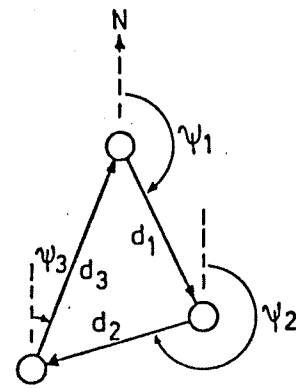
before solving for β , the squared error in time delays is effectively weighted by ρ_i^2 . Thus the final apparent velocity vector will agree best with the time delay corresponding to the highest cross correlation value, and this delay will be modified the least when the apparent velocity is translated back into time delays.

4. Poor man's FCA

This section derives a set of equations for determining FCA parameters from just the positions, t_{\max_i} , and the values of the maxima, ρ_{\max_i} , in the three cross correlations, and one point, (t_a, ρ_a) , from the mean auto correlation.

The assumptions are that the correlation function is Gaussian, and the NTD=0.

Fig. 1 describes the general antenna arrangement. The direction given to each pair defines the sign of t_{\max_i} ; i.e. a pattern moving in the direction of the arrow gives a positive delay.



Antenna array

Fig. 1

The auto correlation is given

by:

$$\rho(0,0,t) = \exp\left(-\frac{1}{2}\left[V^2\left(\frac{\cos^2(\phi-\theta)}{a^2} + \frac{\sin^2(\phi-\theta)}{b^2}\right) + \frac{1}{c^2}\right]t^2\right) \quad \dots(1)$$

Let

$$Q = V^2\left(\frac{\cos^2(\phi-\theta)}{a^2} + \frac{\sin^2(\phi-\theta)}{b^2}\right) + \frac{1}{c^2} \quad \dots(2)$$

$$\text{Then } \rho(0,0,t) = \exp(-\frac{1}{2}Qt^2), \quad \dots(3)$$

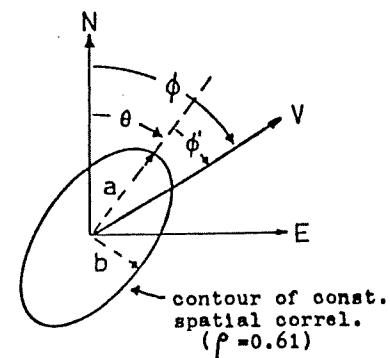
and Q can be found from one point on the auto correlation:

$$Q = \frac{-2\ln \rho_a}{t_a^2} \quad \dots(4)$$

The cross correlations are given by:

$$\rho(d_i, \psi_i, t) = \exp\left[-\frac{1}{2}\left(\frac{(d_i \cos(\psi_i - \theta) - V \cos(\phi - \theta)t)^2}{a^2} + \frac{(d_i \sin(\psi_i - \theta) - V \sin(\phi - \theta)t)^2}{b^2} + \frac{t^2}{c^2}\right)\right] \quad \dots(5)$$

where the antenna pair and direction are defined by d_i and ψ_i as shown in Fig. 1. The other parameters are the axes of the characteristic ellipse (a, b), the characteristic time (c), the true velocity (V), the tilt angle (θ) of the ellipse, and the drift direction

FCA parameters
Fig. 2

(ϕ). These are shown in Fig. 2.

Then for the i^{th} antenna pair, t_{max_i} is found by setting the derivative of (5) equal to zero.

$$t_{\text{max}_i} = \frac{Vd_i}{Q} \left(\frac{\cos(\phi - \theta) \cos(\psi_i - \theta)}{a^2} + \frac{\sin(\phi - \theta) \sin(\psi_i - \theta)}{b^2} \right) \quad \dots(6)$$

This is substituted into (4) to get ρ_{max_i} :

$$-2 \ln \rho_{\text{max}_i} = d_i^2 \left(\frac{\cos^2(\psi_i - \theta)}{a^2} + \frac{\sin^2(\psi_i - \theta)}{b^2} \right) - Qt_{\text{max}_i}^2 \quad \dots(7)$$

Now define

$$m_i = \frac{-2 \ln \rho_{\text{max}_i} + Qt_{\text{max}_i}^2}{d_i^2} \\ = \frac{1}{r^2 b^2} \left(\cos^2(\psi_i - \theta) + \sin^2(\psi_i - \theta) r^2 \right) \quad \dots(8)$$

where $r = a/b$, and let $m_{i,j} = m_i - m_j$.

Then the solution for the pattern parameters is:

$$\tan(2\theta) = \frac{m_{32} \cos^2 \psi_1 + m_{13} \cos^2 \psi_2 + m_{21} \cos^2 \psi_3}{-(m_{32} \sin^2 \psi_1 + m_{13} \sin^2 \psi_2 + m_{21} \sin^2 \psi_3)} \\ \frac{r^2}{1-r^2} = \frac{m_1 \cos^2(\psi_2 - \theta) - m_2 \cos^2(\psi_1 - \theta)}{m_{21}} \quad \dots(9) \\ b^2 = \frac{\frac{3}{2} \left[\frac{1}{r^2} + 1 \right]}{m_1 + m_2 + m_3}, \quad a^2 = b^2 r^2$$

The drift velocity is found from equation (6) for $i=1,2$.

$$\tan(\phi - \theta) = \frac{t_1 d_2 \cos(\psi_2 - \theta) - t_2 d_1 \cos(\psi_1 - \theta)}{t_2 d_1 \sin(\psi_1 - \theta) - t_1 d_2 \sin(\psi_2 - \theta)} \left(\frac{b^2}{a^2} \right) \quad \dots(10)$$

$$V = \frac{Q t_1}{d_1 \left[\frac{\cos(\phi - \theta) \cos(\psi_1 - \theta)}{a^2} + \frac{\sin(\phi - \theta) \sin(\psi_1 - \theta)}{b^2} \right]} \quad \dots(11)$$

where $t_i = t_{\max_i}$. Then c is found from equation (1):

$$c = \sqrt{\frac{1}{Q - V^2 \left[\frac{\cos^2(\phi - \theta)}{a^2} + \frac{\sin^2(\phi - \theta)}{b^2} \right]}} \quad \dots(12)$$

If a^2 , b^2 , or c^2 is negative, the FCA solution must be rejected as physically impossible.

Table 1 gives an example for testing the programmed solution. Note that the quoted V (drift) is the pattern velocity (from eqn. 11) divided by 2.

Table 1.

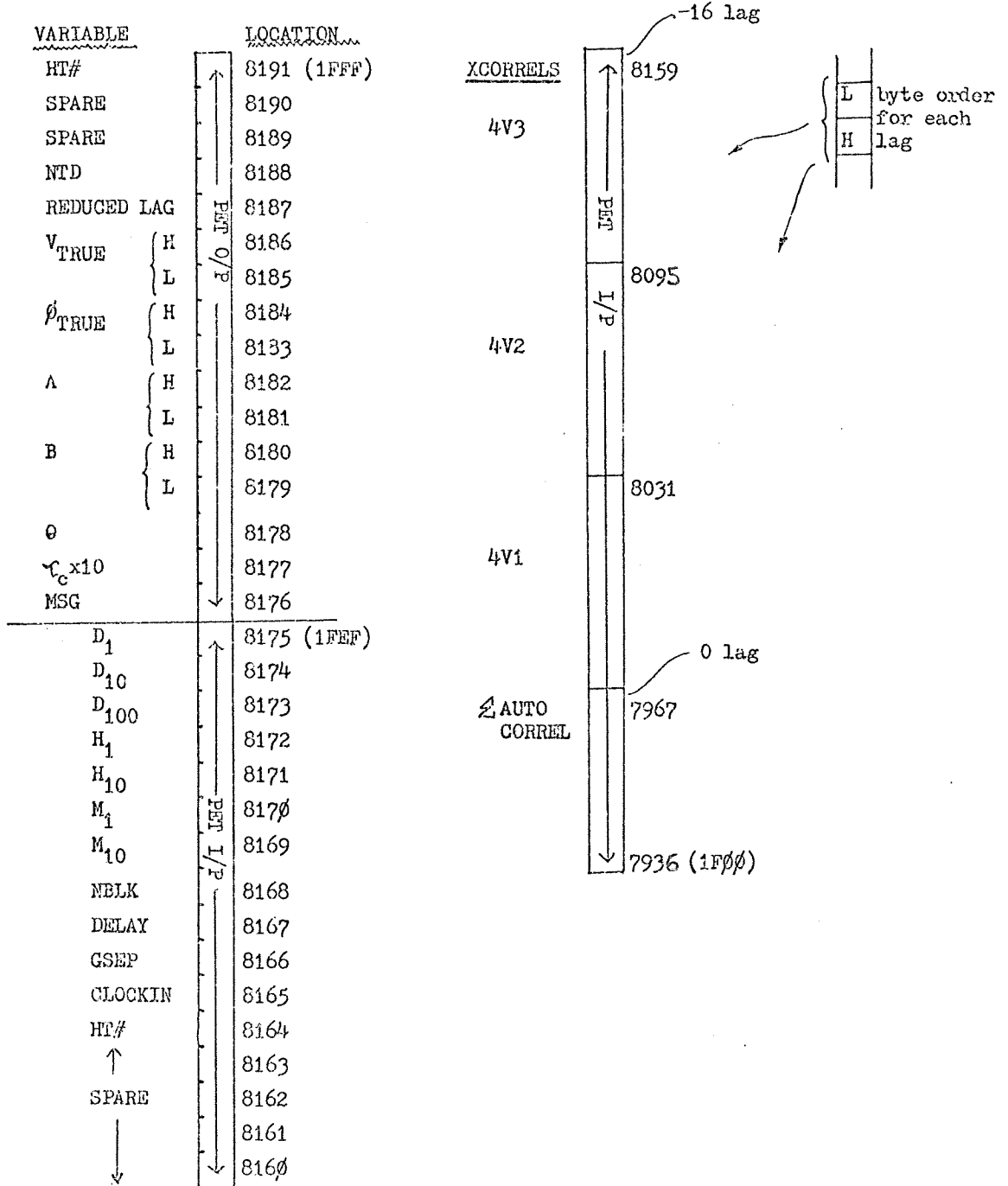
antenna parameters: $d_1 = d_2 = d_3 = 156\text{m}$; $\psi_1 = 210^\circ$, $\psi_2 = 90^\circ$, $\psi_3 = 330^\circ$

<u>input data</u>	<u>solution</u>
$t_a = 1.066 \text{ sec}$	$Q = 0.53$
$\rho_a = 0.74$	$a = 97.4\text{m}$
$t_{\max_i} = -2.05\text{s}, 2.84\text{s}, -.79\text{s}$	$b = 56.0\text{m}$
$\rho_{\max_i} = 0.17, 0.25, 0.31$	$\theta = 158^\circ$
	$V = 16.6\text{m/s (drift)}$
	$\phi = 90^\circ$
	$c = 2.2\text{s}$

5. Pet Machine Language routine

ADDRESS		PROGRAM			ASSEM.	
DEC	HEX	HEX	DEC			
826	\$033A	A9 FF	169 255		LDA #\$FF	} OUTPUT \$10bytes
	033C	8D 43 E8	141 67 232		STA \$E843	
	033F	A2 10	162 16		LDX #\$10	
	0341	BD EF 1F	189 239 31		LDA \$1FEF, X	
	0344	8D 41 E8	141 65 232		STA \$E841	
	0347	A9 EA	169 234		LDA #\$EA	
	0349	8D 4C E8	141 76 232		STA \$E84C	
	034C	A9 DA	169 218		LDA #\$DA	
	034E	8D 4C E8	141 76 232		STA \$E84C	
	0351	AD 4D E8	173 77 232		LDA \$E84D	
	0354	29 02	41 2		AND #\$02	
	0356	F0 F9	240 249		BEQ \$0351	
	0358	CA	202		DEX	
	0359	D0 E6	208 230		BNE \$0341	
	035B	AD 41 E8	173 65 232		LDA \$E841	
	035E	A9 00	169 0		LDA #\$00	
	0360	8D 43 E8	141 67 232		STA \$E843	
	0363	A2 F0	162 240		LDX #\$F0	
	0365	A9 EA	169 234		LDA #\$EA	
	0367	8D 4C E8	141 76 232		STA \$E84C	
	036A	AD 4D E8	173 77 232		LDA \$E84D	
	036D	29 02	41 2		AND #\$02	
	036F	F0 F9	240 249		BEQ \$036A	
	0371	AD 41 E8	173 65 232		LDA \$E841	
	0374	9D FF 1E	157 255 30		STA \$1EFF, X	
	0377	A9 EA	169 234		LDA #\$EA	
	0379	9D 4C E8	141 76 232		STA \$E84C	
	037C	A9 DA	169 218		LDA #\$DA	
	037E	8D 4C E8	141 76 232		STA \$E84C	
	0381	CA	202		DEX	
	0382	D0 E6	208 230		BNE \$036A	
	0384	AD 41 E8	173 65 232		LDA \$E841	
903	0387	60	96		RTS	} INPUT \$F0bytes

6. Input/output memory locations



7. Pet Program: version 6d

```

10 REM-PMFCA FOR REALTIME WINDS
20 DIMC(3,32),M(3),MX(3,2),T(3,2),R(3,2),A(6),DX(16)
30 D=156:P=57.29578:P1=90/P:P2=210/P:P3=330/P
40 D2=D*D:X1=D*COS(P1):X2=D*SIN(P1):X3=D*COS(P2)
50 X4=D*SIN(P2):X5=D*COS(P3):X6=D*SIN(P3)
60 U1=COS(2*P1):U2=COS(2*P2):U3=COS(2*P3)
70 V1=SIN(2*P1):V2=SIN(2*P2):V3=SIN(2*P3)
80 TS=.533:CC=.4:LX=.1:CY=.133
90 REM N=#SAMPLES,TS=TSTEP,CC=AUTO LMT,LX=CROSS LMT
110 DATA169,255,141,67,232,162,16,189,239,31,141,65,232,169,234,141,76
111 DATA232,169,218,141,76,232,173,77,232,41,2,240,249,202,208,230,173,65
112 DATA232,169,0,141,67,232,162,240,169,234,141,76,232,173,77,232,41,2,240
113 DATA249,173,65,232,157,255,30,169,234,141,76,232,169,218,141,76,232,202
114 DATA208,230,173,65,232,96
120 POKE1,58:POKE2,3
121 FORI=825TO903
122 READXX
123 POKE1,XX
124 NEXTI
130 XX=USR(XX)
131 K=8159
132 FORI=1TO3
133 FORJ=1TO32
134 C(I,J)=PEEK(K-1)*256+PEEK(K)
135 K=K-2
136 NEXTJ
137 NEXTI
138 FORI=1TO6:A(I)=(PEEK(K)+256*PEEK(K-1))/3.:K=K-2:NEXTI
140 FORI=1TO16
141 DX(I)=PEEK(8159+I)
142 NEXTI
150 DYX=DX(14)*100+DX(15)*10+DX(16)
151 HRX=DX(13)+DX(12)*10:MNX=DX(11)+DX(10)*10:HTX=49+3*DX(5):N=DX(9)*64
152 IFHTX=49THENPRINT"D/T="DYX:HRX:MNX:"N=":N
153 PRINT"HT=":HTX:"SIG":PEEK(8163):"GN=":PEEK(8162)
160 IX=0:IR=0:IS=0:NT=255:RL=15
161 IFA(I)=NTHENNT=254:GOTO2000
160 A1=A(I)*A(I)/N/N:A2=A(I)/N*(1-A(I)/N)
510 S=2
520 FORI=1TO3
530 X0=(C(I,17)/N-A1)/A2
540 IFX0<STHENS=X0
550 NEXTI
551 X0=S
600 RA=CC:S0=1
610 FORI=2TO6
630 S=SIN(1.5708*((A(I)/(N-I+1)-A1)/A2))
640 IFS<C0THEN690
650 S0=S
660 NEXTI
670 TA=(I-2)*TS:RA=S0:GOTO720
690 IFI>2THEN710
700 PRINT"FF":IX=7:GOTO2000
710 TA=(I-2+(S0-CC)/(S0-S))*TS
720 IFX0>0THEN740
730 RL=15:GOTO750
740 X0=SIN(1.5708*X0):RL=TA*((LOG(X0)-2)/LOG(RA))/TS+3
741 RL=INT(RL+0.5)
745 IFRL>15THENRL=15

```

```

750 REM-FIND PEAKS
760 M1=17-RL
770 M2=16+RL
780 FORI=1TO3
790 M=0:R(I,2)=0
800 FORJ=M1TOM2
805 S=C(I,J):IF S<N*A1 THEN840
810 IF S<C(I,J-1) THEN840
820 IF S<C(I,J+1) THEN840
830 M=M+1:IF M=1 THEN836
832 IF S<R(I,1) THEN834
833 MX(I,2)=MX(I,1):R(I,2)=R(I,1):MX(I,1)=J:R(I,1)=S:GOTO840
834 IF S>R(I,2) THENMX(I,2)=J:R(I,2)=S:GOTO840
835 GOTO840
836 MX(I,1)=J:R(I,1)=S
840 NEXTJ
850 M(I)=M
960 NEXTI
961 GOSUB1600
965 IFIX=1 THENGOTO970
966 IFIX=2 THENGOTO970
967 PRINT MX(1,1),MX(2,1),MX(3,1);INT(R(1,1)*100);INT(R(2,1)*100);INT(R(3,1)*100)
970 IFIR=4 THENGOTO2000

1000 REM-WEIGHTED LSFIT
1010 S1=R1*R1:S2=R2*R2:S3=R3*R3
1020 Q1=X1*X1*S1+X3*X3*S2+X5*X5*S3
1030 Q=X2*X1*S1+X3*X4*S2+X5*X6*S3
1040 Q2=X2*X2*S1+X4*X4*S2+X6*X6*S3
1050 D1=Q1*Q2-Q*Q:A1=Q2/D1:A=-Q/D1:A2=Q1/D1
1060 X=T1*(X1*A1+A*X2)*S1+T2*(X3*A1+X4*A)*S2
1070 X=X+T3*(X5*A1+X6*A)*S3:Y=T1*(X1*A+X2*A2)*S1
1080 Y=Y+T2*(X3*A+X4*A2)*S2+T3*(X5*A+X6*A2)*S3
1090 Q=X*X+Y*Y
1100 GOSUB1420
1110 VA=1/SQR(Q):PA=A:REM END OF WLSFIT
1130 T1=D*COS(P1-A1)/VA:T2=D*COS(P2-A1)/VA:T3=D*COS(P3-A1)/VA
1140 VA=VA/2
1150 REM-PMFCA
1160 Q=-2*LOG(RA)/TA/TA
1170 MA=(-2*LOG(R1)+Q*T1*T1)/D2
1180 MB=(-2*LOG(R2)+Q*T2*T2)/D2
1190 MC=(-2*LOG(R3)+Q*T3*T3)/D2
1200 N2=MC-MB:N3=MA-MC:N1=MB-MA
1210 Y=N2*U1+N3*U2+N1*U3:X=-N2*V1-N3*V2-N1*V3
1220 GOSUB1420
1230 T0=A1/2:TL=A/2
1240 C1=COS(P1-T0):C2=COS(P2-T0):S1=SIN(P1-T0):S2=SIN(P2-T0)
1250 R2=(MA+C2*C2-MB*C1*C1)/N1:R2=R2/(1+R2)
1260 B2=(C1+C1/R2+S1*S1)/MA:A2=R2*B2
1265 IFA2<0ORB2<0 THEN PRINT"IMAG AXIS":IX=5:GOTO2000
1270 AX=SQR(A2):BX=SQR(B2)
1280 X=A2*(T2*S1-T1*S2):Y=B2*(T1*C2-T2*C1)
1290 GOSUB1420
1300 PT=A+TL:PT=PT-180*(1+SGN(P1-360))
1310 Q1=COS(A1):Q2=SIN(A1)
1320 VT=Q*T1/D/(Q1*C1/R2+Q2*S1/B2)/2
1330 IFVT>0 THEN1350
1340 PT=PT-180*SGN(P1-180):VT=-VT
1350 C2=1/(Q-4*VT*VT*(Q1*Q1/R2+Q2*Q2/B2))
1355 IFC2<0 THENPRINT"IMAG TC":IX=6:GOTO2000
1360 TC=SQR(C2)
1370 IFA2>0 THEN1390
1380 A1=AX:AX=BX:BX=A1:TL=TL-90*SGN(TL-90)
1390 PRINT"FCA":INT(VT);INT(PT);INT(AX);INT(BX);INT(TL);INT(TC*10)/10
1400 REM END OF FCA
1410 GOTO2100
1420 REM ARCTAN (Y/X), 0 TO 360
1430 IFX<>0 THEN1450
1440 A=90*(2-SGN(Y)):GOTO1500
1450 A=ATN(Y/X)*P+90*(2-SGN(Y))*(1+SGN(X))
1500 A1=A/P
1510 RETURN

```

```

1600 REM SUBROUTINE WEED FOR PMFCA
1610 FORK=1T03
1615 IFM(K)<=0 THEN PRINT "NLP";K:IR=4:IX=1:RETURN
1618 IFM(K)>2 THEN M(K)=2
1619 MM=M(K)
1620 FORJ=1TMM:MX=MX(K,J)
1630 Y1=C(K,MX-1):Y2=R(K,J):Y3=C(K,MX+1)
1640 A5=(Y1+Y3)/2-Y2:B5=(Y3-Y1)/2:X=-B5/A5/2
1650 S=A5*X*X+B5*X+Y2
1660 T(K,J)=(X+MX-17)*TS-K*CY
1670 R(K,J)=SIN(1.5708*((S/(N-ABS(MX-17))-A1)/A2))
1680 IFR(K,J)<LX THEN M(K)=M(K)-1
1685 IFM(K)=0 THEN PRINT "EXIT WEED";K:IR=4:IX=2:RETURN
1690 NEXTJ
1700 IFM(K)>2 THEN 1720
1710 IFR(K,1)/R(K,2)>1.8 THEN M(K)=M(K)-1
1720 NEXTK
1730 REM HERE M(K)=1 OR 2
1740 IS=M(1)+M(2)+M(3)-3
1750 ON IS+1 GOTO 1760,1800,1900,1900
1750 S1=0:S2=0:FORK=1T03:S1=S1+T(K,1):S2=S2+ABS(T(K,1)):NEXTK
1761 NT=INT(ABS(S1)/S2*200)
1770 IFABS(S1)/S2>.3 THEN IR=4:IX=3:RETURN
1780 GOTO 1940
1800 FORK=1T03:IFM(K)=2 THEN KI=K
1805 NEXTK:IR=KI
1810 K2=3-INT(KI/3):K3=2-INT(KI/2):S1=T(K2,1)+T(K3,1)
1815 S2=ABS(T(K2,1))+ABS(T(K3,1)):NT=INT(ABS(S1+T(K1,1))/(S2+ABS(T(K1,1)))+200)
1820 FORK=1T02:S=T(K1,K):IFABS(S1+S)/(S2+ABS(S))<.2 THEN 1840:NEXTK
1830 IR=4:IX=4:RETURN
1840 T(K1,1)=T(K1,K):R(K1,1)=R(K1,K):GOTO 1940
1900 NT=ABS(T(1,1)+T(2,1)+T(3,1))/(ABS(T(1,1))+ABS(T(2,1))+ABS(T(3,1)))
1910 IFNT>.1 THEN IR=4:NT=INT(NT*200):RETURN
1920 IR=5:NT=INT(NT*200)
1940 T1=T(1,1):T2=T(2,1):T3=T(3,1):R1=R(1,1):R2=R(2,1):R3=R(3,1)
1950 RETURN
1990 REM-O/P
2000 VT=0:PT=0:AX=0:BX=0:TL=0:TC=0
2001 PRINT "ERR";NT:RL:IS:IR:IX
2100 IFVT>65280 THEN VT=65280
2101 XX=INT(VT/256):POKE8186,XX
2105 ERX=IS+IR*4+IX*32
2115 XX=INT(VT-XX*256):POKE8185,XX
2120 XX=INT(PT/256):POKE8184,XX:XX=INT(PT-XX*256):POKE8183,XX
2125 IFAX>65280 THEN AX=65280
2126 XX=INT(AX/256):POKE8182,XX
2130 XX=INT(AX-XX*256):POKE8181,XX
2135 IFBX>65280 THEN BX=65280
2136 XX=INT(BX/256):POKE8180,XX
2140 XX=INT(BX-XX*256):POKE8179,XX
2145 XX=INT(TL):POKE8178,XX:XX=TC*10
2146 IFXX>255 THEN XX=255
2147 POKE8177,XX:POKE8176,ERX
2155 POKE8188,NT:POKE8187,RL
2160 POKE8190,HRX:POKE8189,MNX:XX=DX*5:POKE8191,XX
2170 GOTO 130

```

(f)

(i)

APPENDIX D

Tape output format

One tape block (592 bytes) constitutes the results of a full wind record (32 heights). It is read in 16 byte "logical records" (LR).

LR#1 : gives, in order, day#(units), day#(tens), day#(hundreds), hours(units), hours(tens), minutes(units), minutes(tens) in GMT; the number of blocks per wind record (1 block = 34 sec at present; i.e. 256 pulses at 7.5 Hz.); and three parameters which define the heights recorded. The first (DLY) gives the nominal delay before the first height gate according to:

$$44+5xDLY\pm 1 = \text{program delay in } \mu\text{sec (assuming that the branches in the interrupt routine do not cross page boundaries)- the uncertainty is in the 'reaction time' of the Apple to an interrupt. Because of equipment design (see Chapter 3, pg. 10) the actual delay before the first height gate is } D = \text{INT}((44+5xDLY\pm 1-10)/20) \times 20 + 30 \mu\text{sec (where INT means 'the integer part of')},$$

and there is no uncertainty in the position of the first height gate. Thus the first height gate (measured with respect to the leading edge of the transmitter pulse) is at $D \times (3/20) - 2$ Km (the 2 Km accounts for the travel time to the transmitter). The second, GSEP, defines the gate separation; 08 gives 3 Km (see Chapter 3, pg. 10). The third, CLKIN, defines a delay which must allow the clocking in of exactly 64 height gates (i.e. it depends on the gate separation): delay = $24 + \text{CLKIN} \times 63 \mu\text{sec}$. This is best set by counting the gates on an oscilloscope. The last five bytes are free, but they are only accessed from the Apple (before any winds are calculated) and in this sense are not very useful. In future they will be used to feed more information to the Pet.

LR#2,3 : gives the average of the average signal in each block for the lowest to highest height gate.

LR#4,5 : gives the gain setting (0=lowest gain, to 7= highest gain) for each height gate (lowest to highest).

LR#6-37 : gives the wind results from each height (#0 to #31; referred to in some places as #1 to #32). These are the direct output of the Pet. The present output in each LR is as follows:

Byte#1 : height number (0-31). This is essentially a free byte since the heights are always in the same order, however it has been used to advantage in reconstructing data recorded on a less than perfect tape machine.

Byte#2,3 : hour and minute. These are also redundant, and may not be relied upon in future modifications.

Byte#4 : If this is 0-200 it is the value of the NTDx200 (for the largest peaks in the cross correlations). If it is 254, the record was rejected due to constant input signal. If it is 255, the NTD was not calculated either because the record was rejected for fast fading, or at least one cross correlation did not have peaks of sufficient size (i.e. > 0.1) to define a time delay.

Byte#5 : is the reduced lag used for finding peaks- a value of 16 (now 15) means that no such reduction was possible.

Byte#6,7 : gives the true drift velocity ($B_6 \times 256 + B_7$) in m/s. Since a true velocity > 256m/s is very unlikely (and probably spurious if it does occur), #6 could be regarded as a free byte, or could be used to increase the accuracy of the output value.

Byte#8,9 : gives the true drift direction in degrees (0-360) East of North;
 $\phi = B_8 \times 256 + B_9$ degrees.

Byte#10,11 : gives the major axis of the characteristic ellipse (at $\rho=0.61$)
on the ground ; $A = B_{10} \times 256 + B_{11}$ metres.

Byte#12,13 : gives the minor axis of the characteristic ellipse; $B = B_{12} \times 256 + B_{13}$.

Byte#14 : gives the tilt angle of the major axis of the characteristic ellipse in degrees East of North (0-180).

Byte#15 : gives the characteristic time (at $\rho=0.61$) in seconds times 10.

(a value ≥ 256 is set equal to 256).

Byte#16 : is a message which indicates reasons for rejection of data - or the process by which data was accepted. It is the sum of three message indicators at present; $IER=IS+IR \times 4 + IX \times 32$ ($IS=0-3, IR=0-5, IX=0-7$), and there is some redundancy which may be cleared up a a later date. The important configurations are listed:

IX=7: data rejected for fast fading

IX=1,2 : No local peaks > 0.1 in at least one cross correlation

IX=3,4 or IX=0 and IR=4 : NTD too high so data rejected

IR=5 : two or more peaks in two or more cross correlations but NTD < 0.1 so FCA attempted.

IX=5,6 : FCA attempted, but no acceptable solution found.

IX=0 and NTD $\neq 254$: FCA solution found.

Note that if no acceptable FCA solution has been found, bytes#6-15 are zero.

Pg.85 shows a dump of a tape block.

Dump of one tape block(HEX)

LR#	day#(104)	hr(22)	min(50)	DLY	GSEP	CLKIN	#blocks/rec (is now 08)	
1	04000102	02000510	39081400	00000000				
2	05040404	04082129	1F353737	31231D18				Avg. sig. (ht#0-31)
3	15191111	28888C41	15100A04	03030304				
4	07070707	07070707	07070706	02020204				
5	05060603	01010000	01010406	06060606				Gain setting (ht# 0-31)
6 ht#	001632FF	10000000	00000000	000000E0				
7	011632FF	10000000	00000000	000000E0				
8	021632FF	10000000	00000000	000000E0				
9 hr	031632FF	10000000	00000000	000000E0				
10	041632FF	10000000	00000000	000000E0				
11	051632FF	10000000	00000000	000000E0				
12 min	06163208	10000A00	5F007600	4DB12A17				MSG
13	07163217	10000A00	51006500	30042B00				$\tau_c \times 10$
14	0816326B	10000000	00000000	00000091				tilt of A
15	091632FF	10000000	00000000	00000050				
16	0A163208	10000000	00000000	00000012				
17	0B163272	10000000	00000000	00000013				
18	0C16321D	10000000	00000000	00000012				
19	0D163202	10002001	13008200	40030D00				
20	0E16320C	10002001	23006900	38B20B05				
21	0F163279	10000000	00000000	00000091				B
22	101632FF	10000000	00000000	000000E0				
23	111632FF	10000000	00000000	000000E0				
24	121632FF	10000000	00000000	000000E0				
25	131632FF	10000000	00000000	000000E0				
26	1416320C	10001800	62009F00	70081B09				A
27	15163200	10001100	55008800	6E0F1900				
28	16163207	10001000	62008C00	5E0B1700				
29	17163210	10001800	5900A500	510D1400				
30	1816321F	10001700	4F005E00	37181200				
31	19163289	00000000	00000000	00000091				
32	1A1632FF	10000000	00000000	00000050				
33	1B1632FF	10000000	00000000	000000E0				
34	1C1632FF	10000000	00000000	000000E0				
35	1D1632FF	10000000	00000000	000000E0				
36	1E1632FF	10000000	00000000	000000E0				
37	1F1632FF	10000000	00000000	000000E0				

NTDx200

Reduced lag

V(true)
(10 m/s)

ϕ (true)
(95° E of N)

MSG

$\tau_c \times 10$

tilt of A

B

A

APPENDIX E

Machine language reference

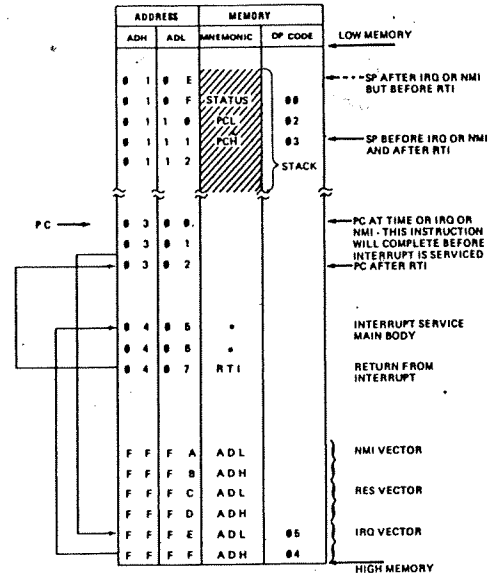


FIG. 1 IRQ, NMI, RTI, BRK OPERATION

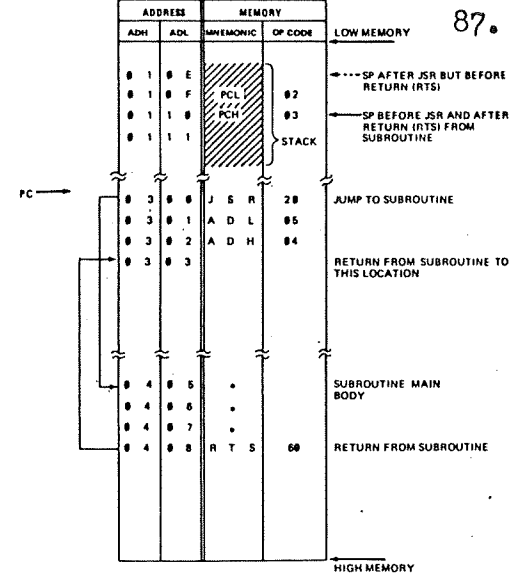


FIG. 2 JSR, RTS OPERATION

ASSEMBLER DIRECTIVES

- OPT - IF USED MUST BE THE FIRST EXECUTABLE STATEMENT IN THE PROGRAM.
- OPTIONS ARE - (OPTIONS LISTED ARE THE DEFAULT VALUE, TURNED OFF BY (NO) PREFIX.)
- COUNT (COU OR CNT) - LIST ALL INSTRUCTIONS AND THEIR USAGE.
- NOGENERATE (NOG) - DO NOT GENERATE MORE THAN ONE LINE OF CODE FOR ASCII STRINGS.
- XREF (XRE) - PRODUCE A CROSS-REFERENCE LIST IN THE SYMBOL TABLE.
- ERRORS (ERR) - CREATE AN ERROR FILE.
- MEMORY (MEM) - CREATE AN ASSEMBLY OBJECT OUTPUT FILE.
- LIST (LIS) - PRODUCE A FULL ASSEMBLY LISTING.
- BYTE - PRODUCES A SINGLE BYTE IN MEMORY EQUAL TO EACH OPERAND SPECIFIED.
- WORD - PRODUCES TWO BYTES IN MEMORY EQUAL TO EACH OPERAND SPECIFIED.
- * - DEFINES THE BEGINNING OF A NEW PROGRAM COUNTER SEQUENCE.
- PAGE - ADVANCES THE LISTING TO THE TOP OF A NEW PAGE.
- END - DEFINES THE END OF A SOURCE PROGRAM.

LABELS:

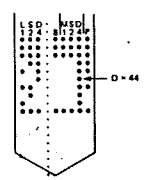
LABELS BEGIN IN COLUMN 1 AND ARE SEPARATED FROM THE INSTRUCTION BY AT LEAST ONE SPACE. LABELS CAN BE UP TO 8 ALPHANUMERIC CHARACTERS LONG AND MUST BEGIN WITH AN ALPHA CHARACTER. A, X, Y, S, AND P ARE RESERVED AND CANNOT BE USED AS LABELS. LABEL - EXPRESSION CAN BE USED TO EQUATE LABELS TO INSTRUCTIONS. LABEL * * * N CAN BE USED TO RESERVE AREAS IN MEMORY

CHARACTERS USED AS SPECIAL PREFIXES:

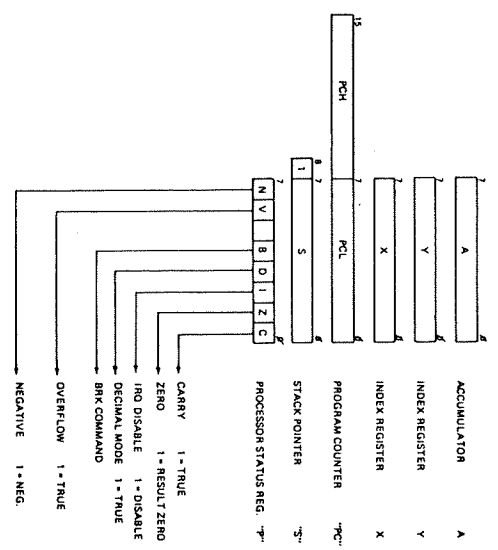
- INDICATES AN ASSEMBLER DIRECTIVE
- # SPECIFIES THE IMMEDIATE MODE OF ADDRESSING.
- \$ SPECIFIES A HEXADECIMAL CHARACTER.
- @ SPECIFIES AN OCTAL NUMBER.
- % SPECIFIES A BINARY NUMBER.
- * SPECIFIES AN ASCII LITERAL CHARACTER.
- { INDICATES INDIRECT ADDRESSING.
- : IN COLUMN 1 INDICATES A COMMENT.

ASCII CHARACTER SET (7-BIT CODE)

MSD	ASCII CHARACTER SET (7-BIT CODE)							
	0	1	2	3	4	5	6	7
LSD	000	001	010	011	100	101	110	111
0	0000 NUL	DLE SP	0	@ P				
1	0001 SOH	DC1 I	1	A Q	a q			
2	0010 STX	DC2 "	2	B R	b r			
3	0011 ETX	DC3 #	3	C S	c s			
4	0100 EOT	DC4 \$	4	D T	d t			
5	0101 ENG	NAK %	5	E U	e u			
6	0110 ACK	SYN &	6	F V	f v			
7	0111 BEL	ETB ' 7	G W	g w				
8	1000 BS	CAN (8	H X	h x			
9	1001 HT	EM]	9	I Y	i y			
A	1010 LF	SUB *	:	J Z	j z			
B	1011 VT	ESC +	...	K [k [
C	1100 FF	FS -	v	L \	l \			
D	1101 CR	GS -	-	M]	m]			
E	1110 SO	RS +	v	N ^	n ^			
F	1111 SI	VS /	/	O _	o _			DEL



PROCESSOR PROGRAMMING MODEL



MCS650X INSTRUCTION SET SUMMARY (6501 Thru 6506)

MOS TECHNOLOGY, INC.
 Valley Forge Corporate Center
 950 Rittenhouse Road
 Norristown, Pa. 19401



MEMORIC	OPERATION	IMMEDIATE		ABSOLUTE		ZERO PAGE		ACCUM.		IMPLIED		(IND,X)		(IND,Y)		Z,PAGE,X		ABS,X		ABS,Y		RELATIVE		INDIRECT		Z,PAGE,Y		CONDITION CODES									
		OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	N	Z	C	I	D	V				
ADC	A+M → A	69	2	2	6D	4	3	65	3	2			61	6	2	71	5	2	75	4	2	7D	4	3	79	4	3										
AND	A&M → A	29	2	2	2D	4	3	25	3	2			21	6	2	31	5	2	35	4	2	3D	4	3	39	4	3										
ASL	A ← A ← 2				BE	6	3	96	5	2	BA	2	1					16	6	2	1E	7	3														
BCC	BRANCH ON C=0																						98	2	2												
BCS	BRANCH ON C=1																						86	2	2												
BEQ	BRANCH ON Z=1																																				
BIT	A&M				2C	4	3	2A	3	2																											
BMI	BRANCH ON N=1																						38	2	2												
BNE	BRANCH ON Z=0																							09	2	2											
BPL	BRANCH ON N=0																							18	2	2											
BRK	(See Fig 11)												80	7	1																						
BVC	BRANCH ON V=0																							58	2	2											
BVS	BRANCH ON V=1																								78	2	2										
CLC	0 ← C												18	2	1																						
CLD	0 ← D												08	2	1																						
CLI	0 ← I												58	2	1																						
CLV	0 ← V												88	2	1																						
CMP	A-M	11	1	C0	2	2	C0	4	3	C5	3	2					C1	6	2	D1	5	2	D5	4	2	D9	4	3									
CPX	X-M			E8	2	2	E8	4	3	E4	3	2																									
CPY	Y-M			C8	2	2	C8	4	3	C4	3	2																									
DEC	M-1 → M				CE	6	3	CE	5	2																											
DEX	X-1 → X												CA	2	1																						
DEY	Y-1 → Y												88	2	1																						
EOR	A ⊕ M → A	11	1	A9	2	2	AD	4	3	A5	3	2					41	6	2	61	5	2	65	4	2	6D	4	3	69	4	3						
INC	M+1 → M				EE	6	3	EE	5	2																											
INX	X+1 → X												E8	2	1																						
INY	Y+1 → Y												C8	2	1																						
JMP	JUMP TO NEW LOC				4C	3	3																														
JSR	(See Fig 7) JUMPSUB				28	6	3																														
LDA	M → A	11	1	A9	2	2	AD	4	3	A5	3	2					A1	6	2	B1	5	2	B5	4	2	B9	4	3									

MEMORIC	OPERATION	IMMEDIATE		ABSOLUTE		ZERO PAGE		ACCUM.		IMPLIED		(IND,X)		(IND,Y)		Z,PAGE,X		ABS,X		ABS,Y		RELATIVE		INDIRECT		Z,PAGE,Y		CONDITION CODES								
		OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	OP	N	N	Z	C	I	D	V			
LDX	M → X	(1)		A2	2	2	AE	4	3	AE	3	2																								
LDY	M → Y	(1)		A8	2	2	AC	4	3	AC	3	2																								
LSR	0 ← C				4E	6	3	4E	5	2	AA	2	1																							
NOP	NO OPERATION												EA	2	1																					
ORA	A ∨ M → A	11	1	A9	2	2	AD	4	3	A5	3	2					81	6	2	11	5	2	15	4	2	1D	4	3	19	4	3					
PHA	A → Ms S → S																																			
PHP	P → Ms S → S																																			
PLA	S → S Ms → A																																			
PLP	S → S Ms → P																																			
ROL	0 ← C				2E	6	3	2E	5	2	2A	2	1																							
RTI	(See Fig 11) RTRN INT																																			
RTS	(See Fig 2) RTRN SUB																																			
SBC	A-M-C → A	(1)		E9	2	2	ED	4	3	E5	3	2					E1	6	2	F1	5	2	F5	4	2	FD	4	3	F9	4	3					
SEC	1 → C																																			
SED	1 → D																																			
SEI	1 → I																																			
STA	A → M				BD	4	3	BD	3	2																										
STX	X → M				BE	4	3	BE	3	2																										
STY	Y → M				BC	4	3	BC	3	2																										
TAX	A → X																																			
TAY	A → Y																																			
TSX	S → X																																			
TXA	X → A																																			
TXS	X → S																																			
TYA	Y → A																																			

(1) ADD 1 TO "N" IF PAGE BOUNDARY IS CROSSED X INDEX X
(2) ADD 1 TO "N" IF BRANCH OCCURS TO SAME PAGE Y INDEX Y
ADD 2 TO "N" IF BRANCH OCCURS TO DIFFERENT PAGE A ACCUMULATOR
(3) CARRY NOT - BORROW M MEMORY PER EFFECTIVE ADDRESS A AND
Ms MEMORY PER STACK POINTER V OR Mz MEMORY BIT 7
Mz MEMORY BIT 6

OP-CODE TABLE

LSD	#	1	2	3	4	5	6	7	8	9	A	B	C	D	E	F	MSD
8	BRK	ORA-IND,X				ORA Z,Page	ASL Z,Page		PHP	ORA-IMM	ASL A						8
1	BPL	ORA-IND,Y				ORA Z,Page,X	ASL Z,Page,X		CLC	ORA-ABS,Y							1
2	JSR	AND-IND,X				AND Z,Page	ROL Z,Page		PLP	AND-IMM	ROL A						2
3	BMI	AND-IND,Y				AND Z,Page,X	ROL Z,Page,X		SEC	AND-ABS,Y							3
4	RTI	EOR-IND,X				EOR Z,Page	LSR Z,Page		PHA	EOR-IMM	LSR A						4
5	BVC	EOR-IND,Y				EOR Z,Page,X	LSR Z,Page,X		CLI	EOR-ABS,Y							5
6	RTS	ADC-IND,X				ADC Z,Page			PLA	ADC-IMM							6
7	BVS	ADC-IND,Y				ADC Z,Page,X			SCD	ADC-ABS,Y							7
8		STA-IND,X				STY Z,Page	STX Z,Page		DEY		TXA						8
9	BCC	STA-IND,Y				STY Z,Page,X	STX Z,Page,X		TYA	STA-ABS,Y							9
A	LDY-IMM	LDA-IND,X	LDX-IMM			LDV Z,Page	LDA Z,Page	LDX Z,Page</									

